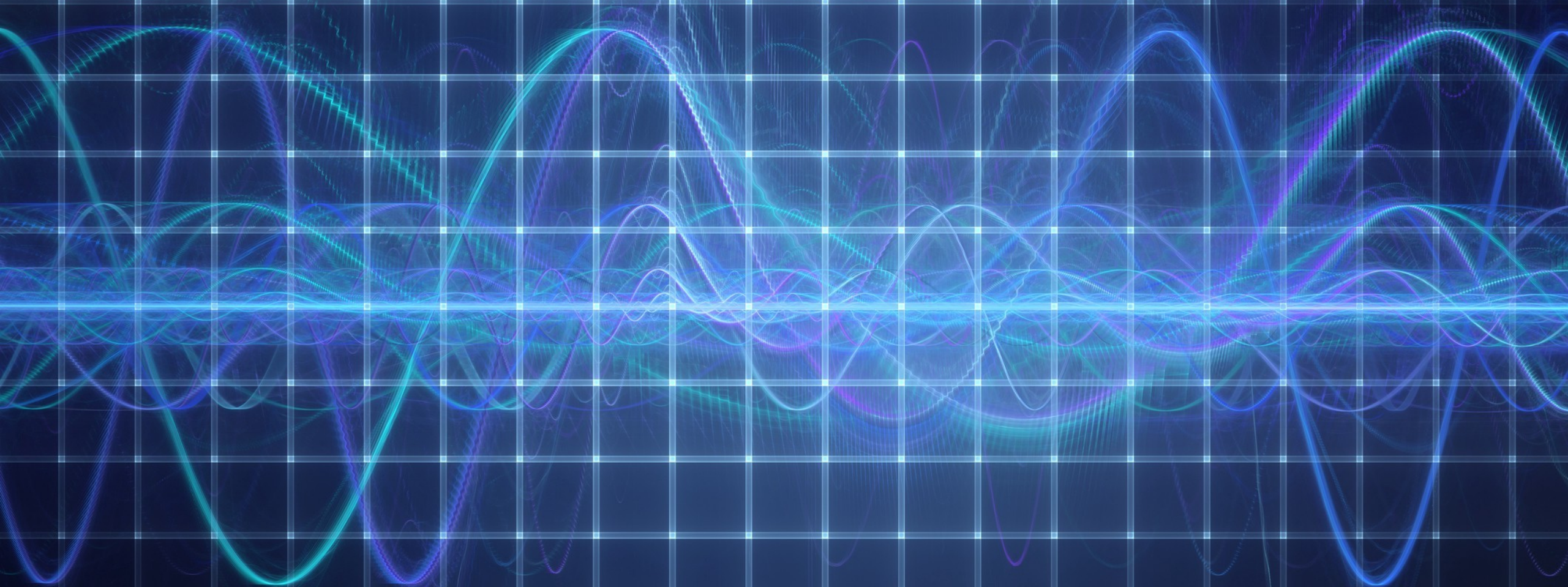


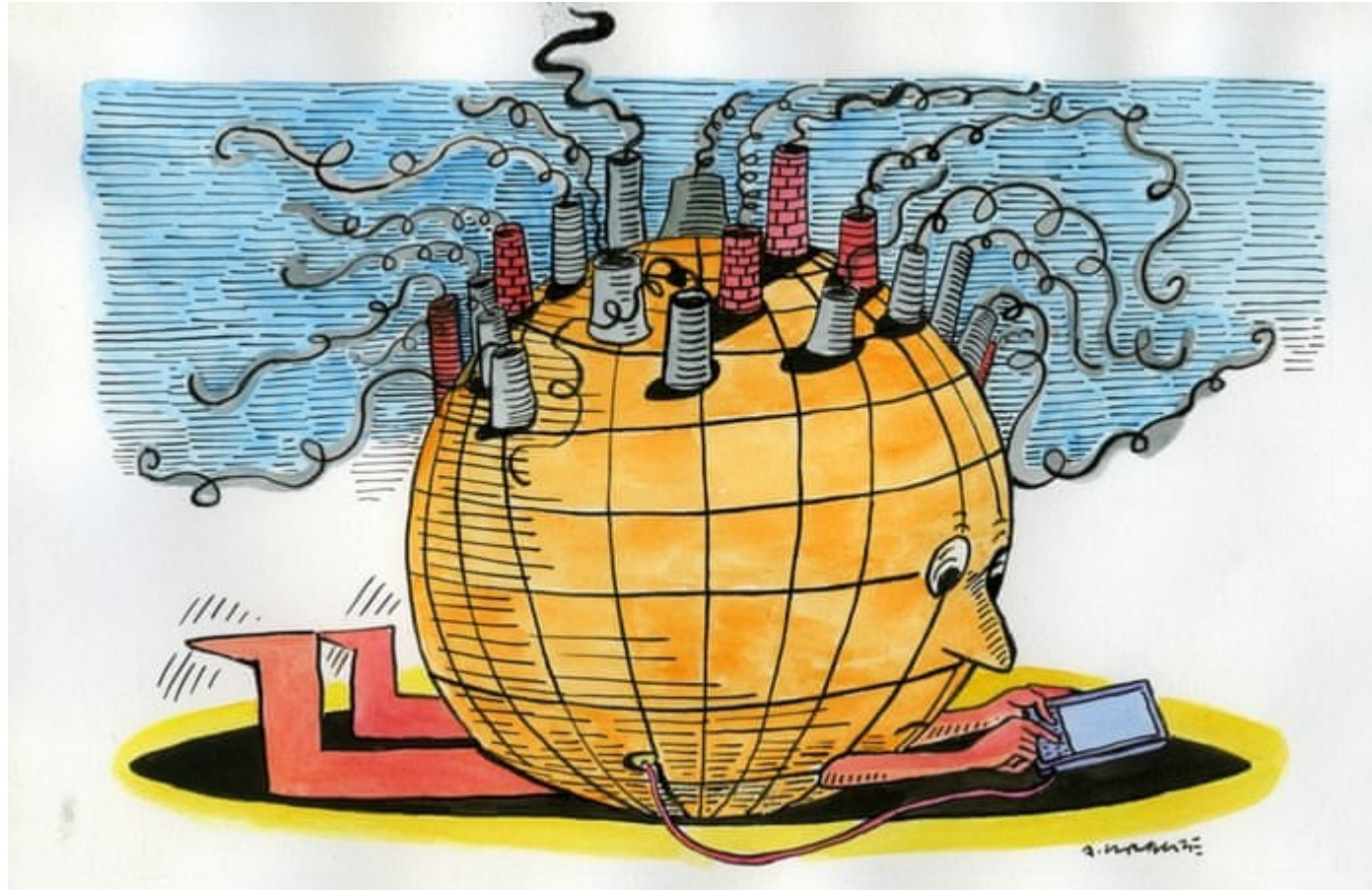


SIMULTANEOUS WIRELESS BACKSCATTER AND POWER TRANSFER

Nuno Borges Carvalho,
Instituto de Telecomunicações
Universidade de Aveiro, Portugal



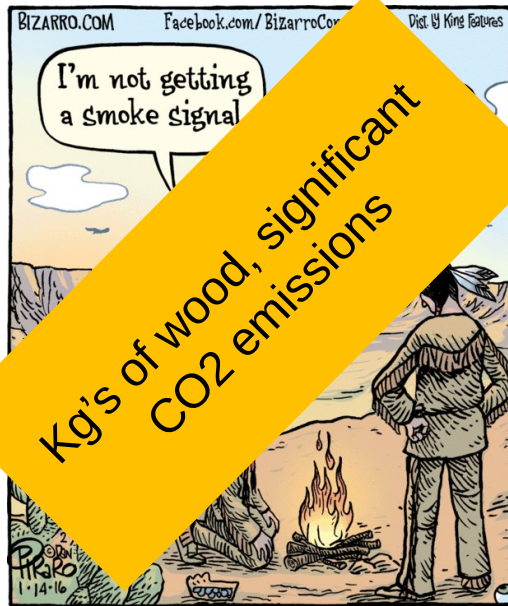
ENERGY IN COMMUNICATIONS



How much Energy I need to comunicate?



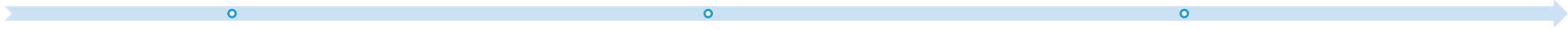
MOTIVATION COMMUNICATION



*Pre historical smoke signals
burning wood*



*Carrier pigeons
corn*



MOTIVATION COMMUNICATIONS



Mobile communications: from 1G to 5G

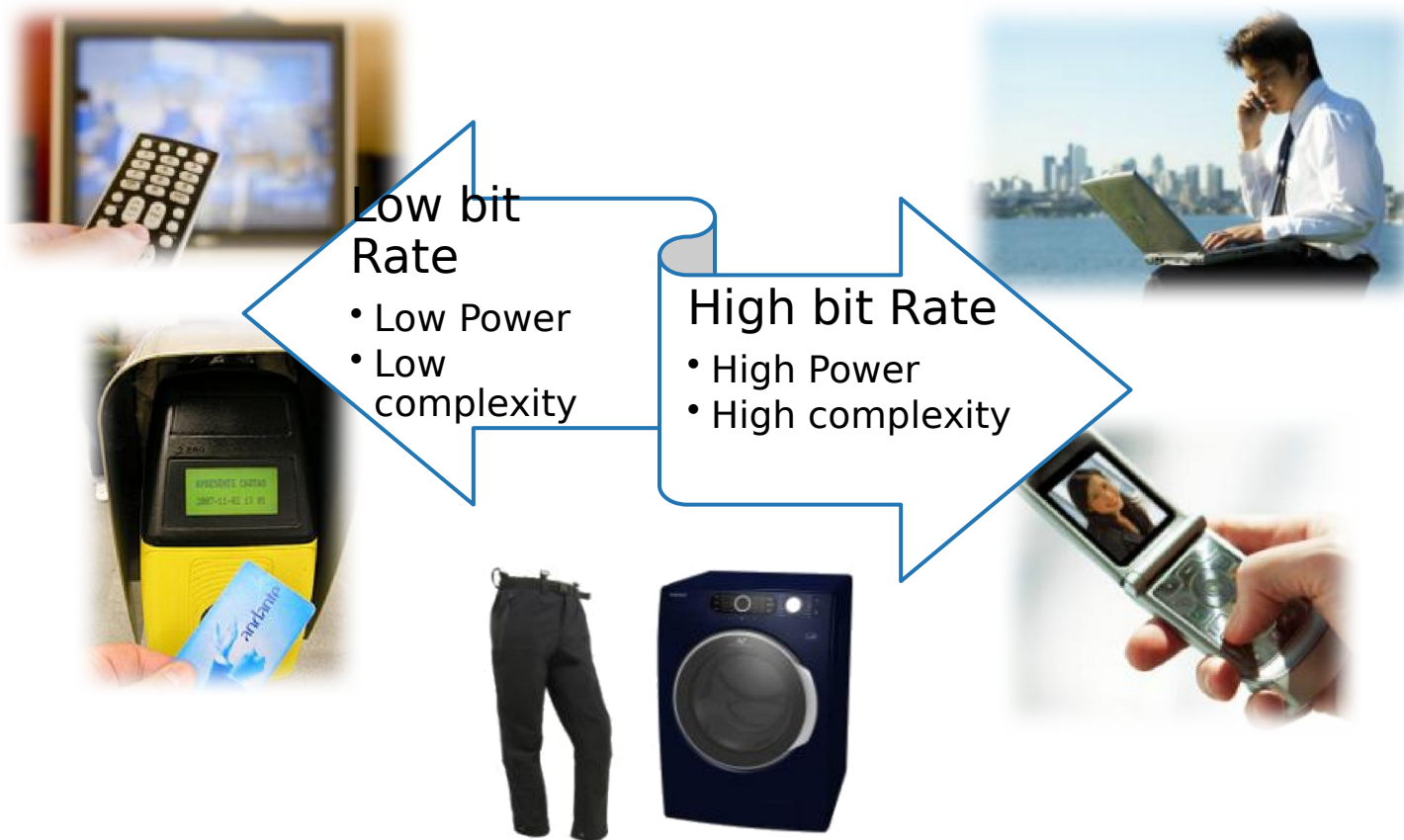
Generation	Device	Specifications
1G		<ul style="list-style-type: none"> Year: 1982 Standards: AMPS, TACS Technology: Analog Bandwidth: 30 kHz Frequency: 900 MHz Services: Voice
2G		<ul style="list-style-type: none"> Year: 1991 Standards: GSM, GPRS, EDGE Technology: Digital Bandwidth: 128 kbps Frequency: 900 MHz Services: Voice, Text, Data
3G		<ul style="list-style-type: none"> Year: 2001 Standards: UTRA, UMB, HSPA Technology: Digital Bandwidth: 3.1 Mbps Frequency: 2.1 GHz Services: Voice, Text, Data, Video
4G		<ul style="list-style-type: none"> Year: 2010 Standards: LTE, WiMAX, 4G-LTE Technology: Digital Bandwidth: 100 Mbps Frequency: 2.6 GHz Services: Voice, Text, Data, Video, Streaming
5G		<ul style="list-style-type: none"> Year: 2020 Standards: 5G NR Technology: Digital Bandwidth: 10 Gbps Frequency: 28 GHz Services: Voice, Text, Data, Video, Streaming, AR/VR, IoT

People

People & Things

5G is about Communication, Storage, Processing...

IOT-IOE WIRELESS THINGS



MOTIVATION ENERGY FOOTPRINT

The bright Smart Future

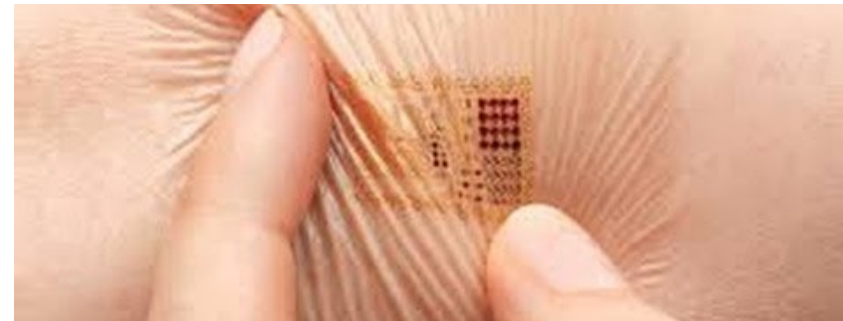


Battery Waste

Huge Amount of Disposals

Large Amount of Energy

IOT WIRELESS THINGS



5G

Mobile communications: from 1G to 5G

Generation	Device	Specifications
1G		1G Year: 1979 Standard: 1G-TACS Technology: FDMA Features: Voice
2G		2G Year: 1991 Standard: GSM, IS-136 Technology: TDMA Features: Voice, SMS
3G		3G Year: 2001 Standard: UTRA, UTRAN Technology: W-CDMA Features: Voice, SMS, Video
4G		4G Year: 2009 Standard: LTE, LTE-Advanced Technology: OFDM Features: Voice, SMS, Video, High Speed
5G		5G Year: 2019 Standard: 5G NR Technology: OFDM, MIMO Features: Voice, SMS, Video, High Speed, Low Latency

People

People & Things

Smart grids, Connected house, eHealth, Entertainment, Traffic priority, Smart Car, Car-to-car communication, Apps beyond imagination

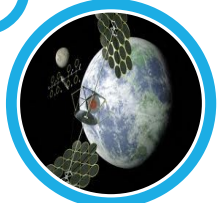
5G is about Communication, Storage, Processing...



Battery-less Sensors for health applications



Car Energy Collector



High Efficient Energy Collection



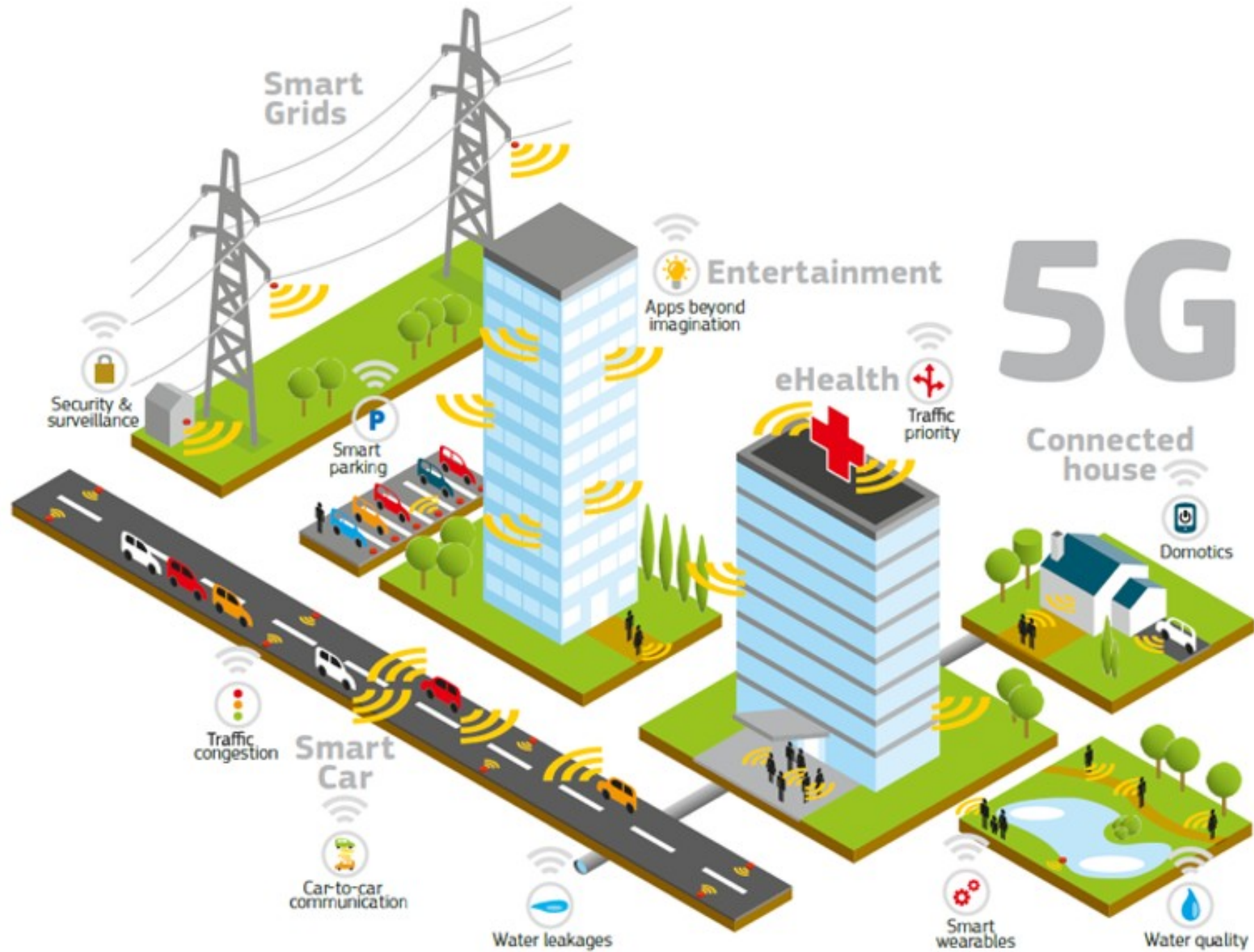
RFID's



Domestic Appliances Wireless Energized

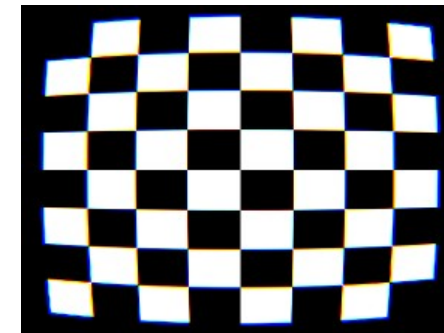
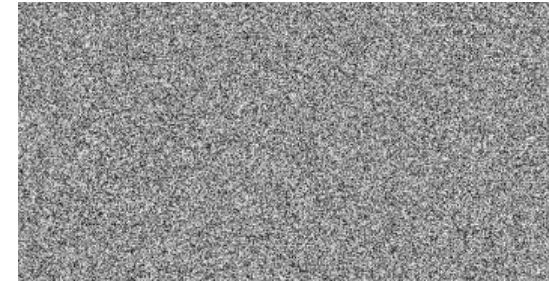
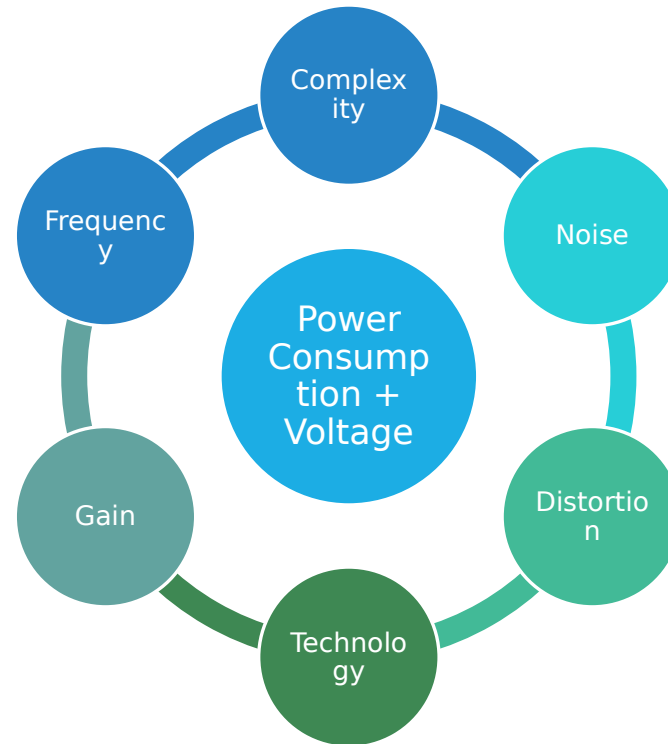
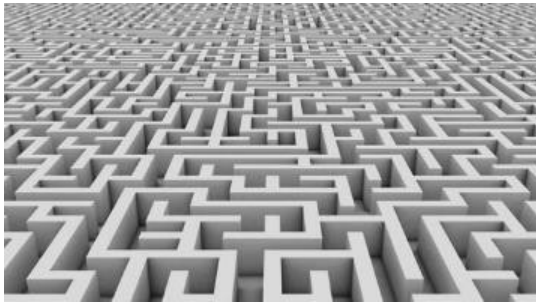


Agriculture passive sensors

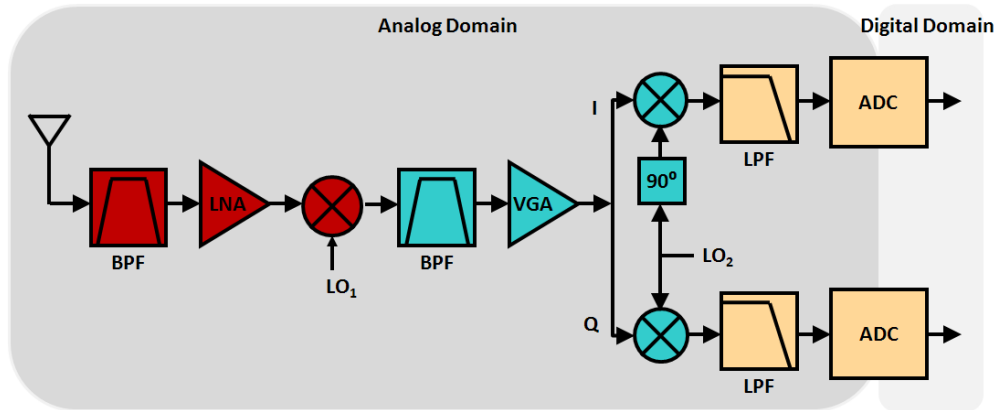


RADIO SYSTEM DRAWBACKS

Limitations



RF RECEIVERS

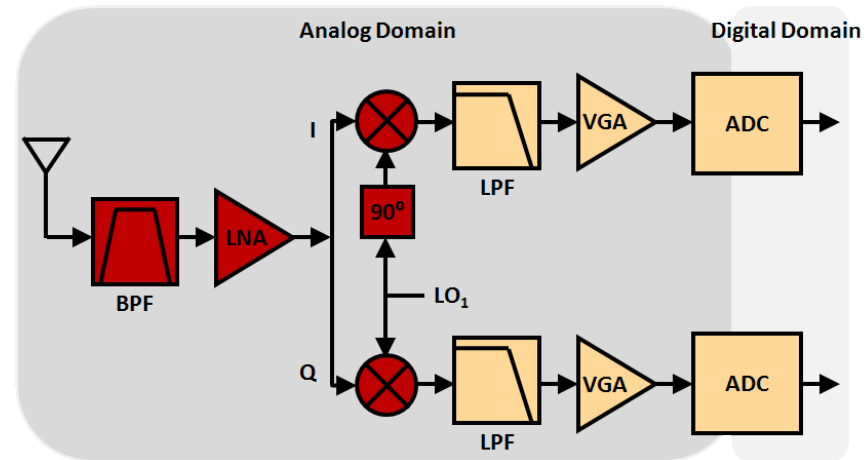


Super-heterodyne

- Conversion to the digital domain at baseband where it can be processed
- Currently adopted in most radio receivers due to low cost components
- Full on-chip integration is concerned and its design to a specific channel → prevents the expansion of receiving band

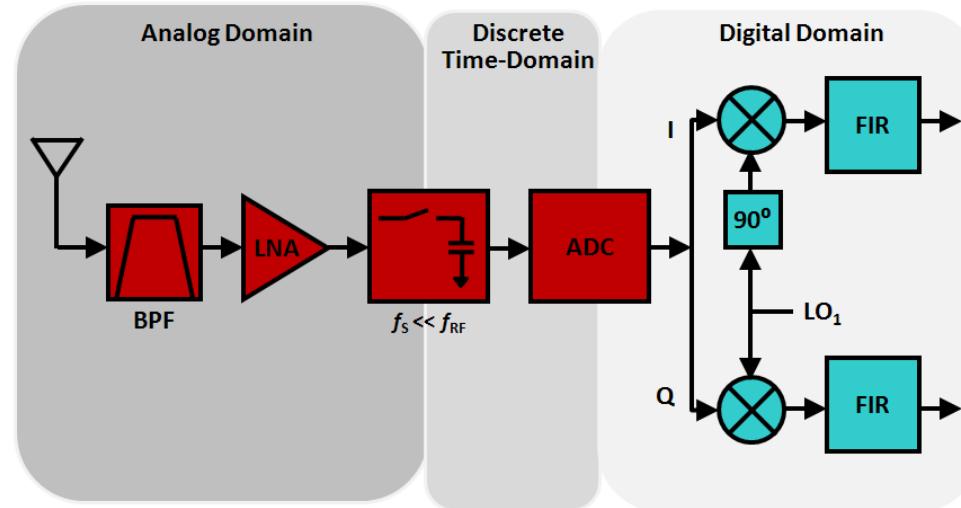
Zero-IF

- Signal is selected at RF by BPF, amplified and directly translated to DC
- Evident reduction in number of components → high level integration
- Components much more difficult to design
DC offset, 2nd order IMD products generated around DC



SOFTWARE DEFINED RADIO

Bandpass Sampling Receiver:

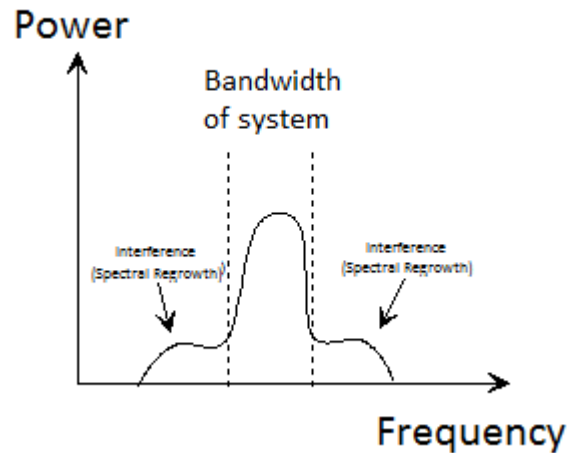


- Takes the fact that S/H circuit translates the signal to 1st Nyquist zone
- Digital processing capabilities exploited → multi-band reception
- Mandatory BPF to avoid overlap of signals → tunable or bank of filters
- Analog BW of ADC must include RF carrier

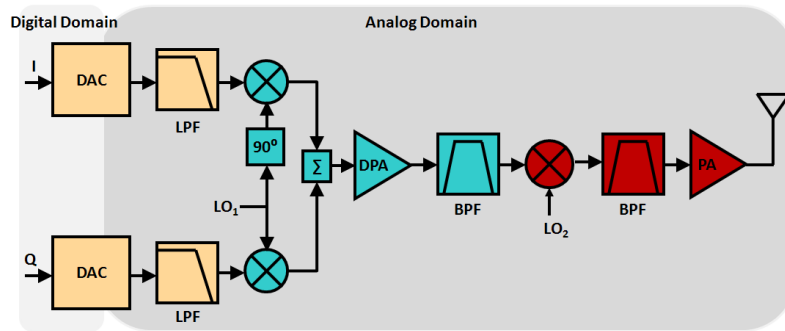
TRANSMITTER

The RF transmitter should also fulfill some requests, for instance:

- Use only the bandwidth that refers to the system standards
- Create low values of harmonic distortion
- Transmit the maximum RF power and simultaneously consume the minimum DC power from the system



TRANSMITTER

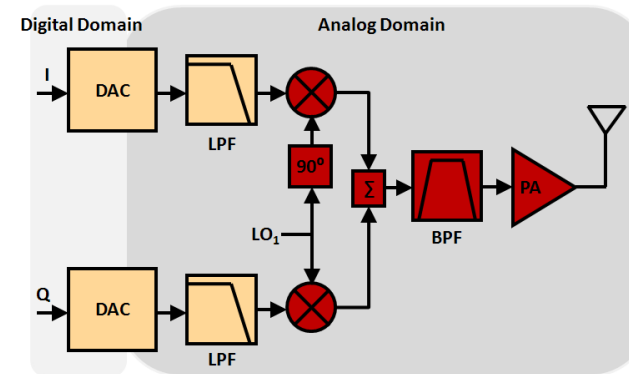


Super-Heterodyne Transmitter

- Digital baseband signals are converted and directly modulated to RF
- Reduced amount of circuitry that allows high level integration
- Carrier leakage, phase gain mismatch, and requires highly linear PA
- With careful design can be employed in SDR TX's

- Signal created in digital domain, modulated at IF, and up-converted
- I/Q modulator working at IF; Output spectrum is far away from LO
- Suffers from similar problems of the receiver case
- Multi-mode implementation is difficult

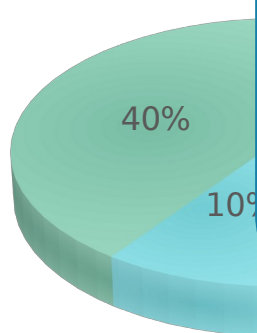
Direct-Conversion Transmitter



If 10% of the world's mobile phone users turned off their chargers after use, the energy saved in one year could power 60,000 European homes.

MOTIVATION NETWORKS ENERGY WASTE II

Base
Consum



5G Communications will have a demand of extra 300% Energy compared with 4G

- Air Conditioning
- Power Suply
- Signal Processing (Analogue-Digital)
- Power amplifier including Feeder

One base station
Network



Source: Nokia calculations based on published operator figures in 2012 UPS ... Uninterruptable Power Supplies

Figure 2: Where energy is consumed in a network - only 15% are used to "transmit bits"

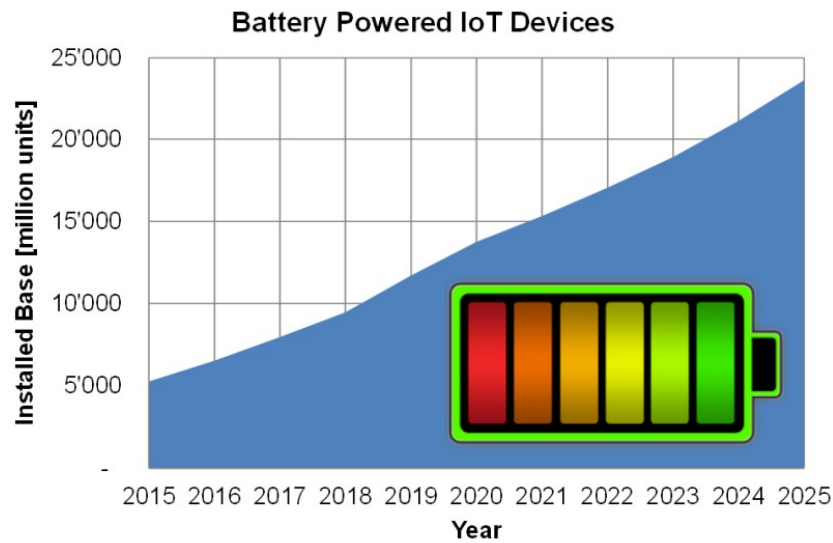
Huawei White Paper

RADIO COMMUNICATIONS

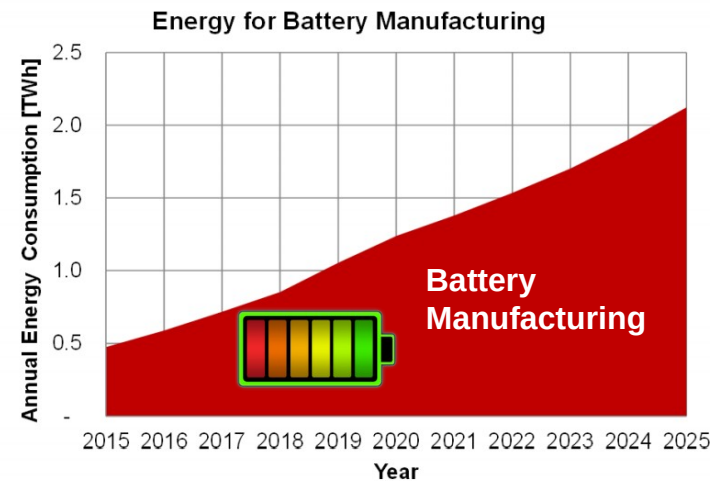
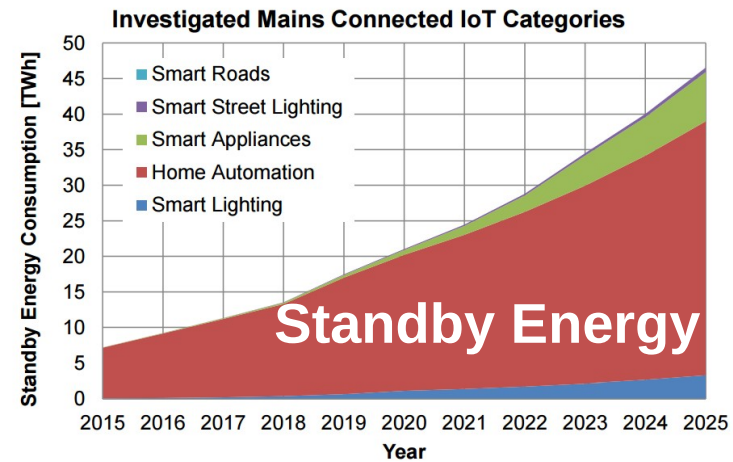
These radio architectures are responsible for a large amount of energy consumption....



MOTIVATION ENERGY W.



Source: 2016 data <http://edna.iea-4e.org>



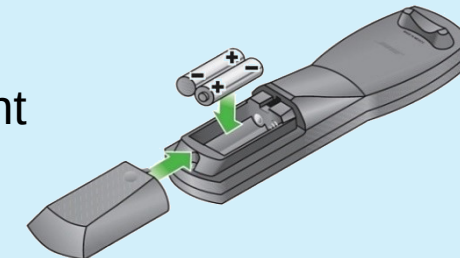
BATTERIES

Batteries take **hundreds of years to decompose**, posing a serious threat to the public health and to the environment.



❖ Considering 4 Million habitual residences in Portugal (INE – Censos 2011) and assuming that:

- ✓ 75% of them have a TV equipment
- ✓ 40% have a cable TV Box
- ✓ 30% have a Sound System



- ❖ We end up with an average of **5.8 Millions of remotes in Portugal**
- ❖ Assuming two batteries per remote and two battery changes per year we have a ...

total of 23.2 Millions batteries being wasted every year !!

Case study



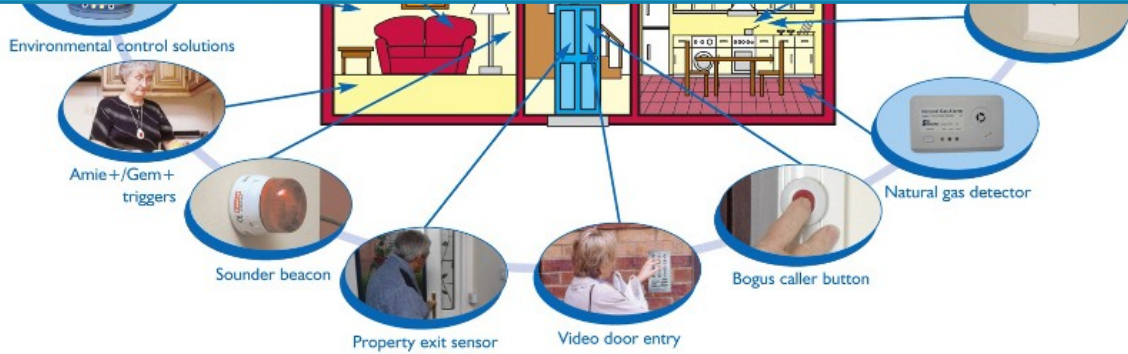
ENERGY CONSUMPTION

APPROACHES SOLUTIONS FOR LOW BIT RATE



Low Cost

How much power is low power?



Power Sensors

MARKET SOLUTIONS

POWER CONSUMPTION

Typical example from one of the top sellers of IoT Sensors

Parameter	Value	Unit
Minimum supply voltage	1.9	V
Temperature range	-40 to +85	°C
Supply current in transmit @ -10dBm output power	9	mA
Supply current in receive mode	12.5	mA
Supply current for μ -controller 4MHz @ 3volt	1	mA
Supply current for ADC	0.9	mA
Maximum transmit output power	10	dBm
Data rate	50	kbps
Sensitivity	-100	dBm
Supply current in power down mode	2.5	μ A

- 2.4 GHz transceiver
 - -96 dBm sensitivity in *Bluetooth*[®] low energy mode
 - 1 Mbps, 2 Mbps supported data rates
 - TX power -20 to +4 dBm in 4 dB steps
 - Single-pin antenna interface
 - 5.3 mA peak current in TX (0 dBm)
 - 5.4 mA peak current in RX
 - RSSI (1 dB resolution)

SoC 1 - Sub <1GHz

Transmit – 17mW
Receive – 24mW

SoC 2 – UltraLow Power Bluetooth

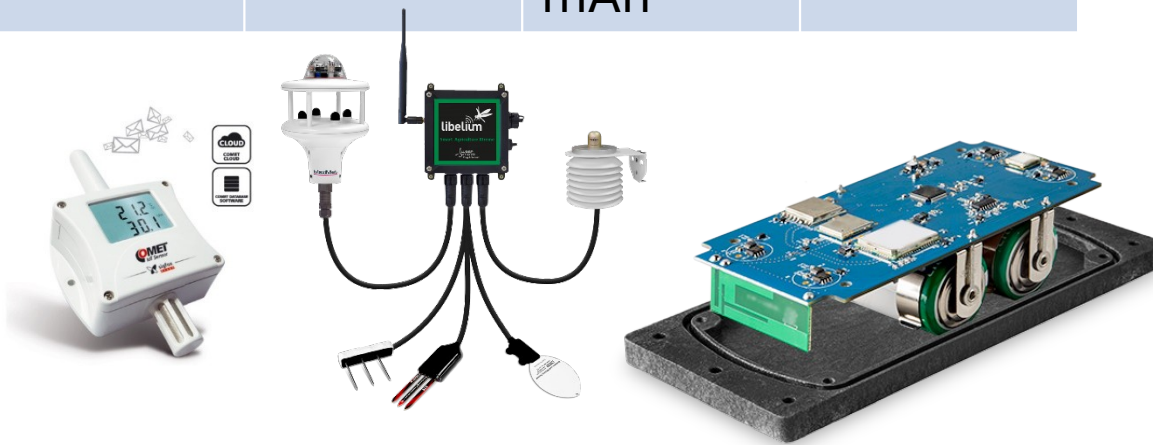
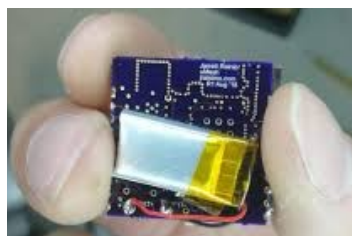
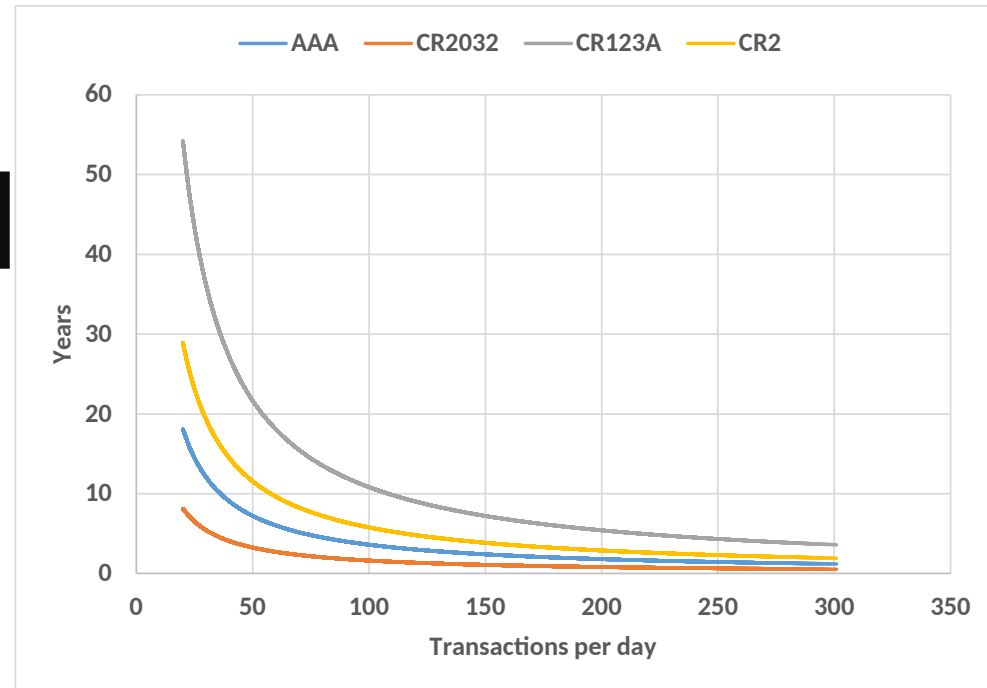
Supply voltage 1.7 to
3,6 V
Transmit/Receive –
9 mW

MARKET SOLUTION

CONSUMPTION

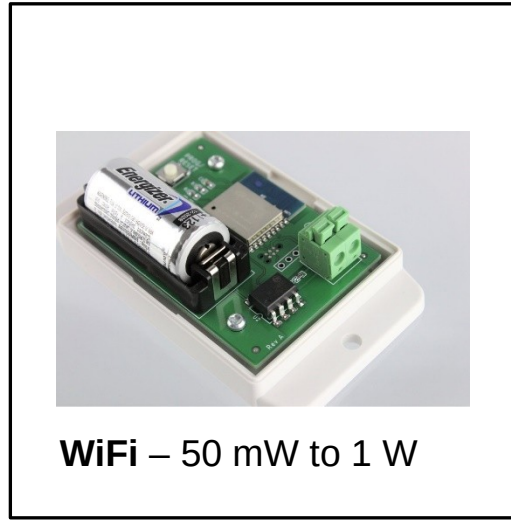
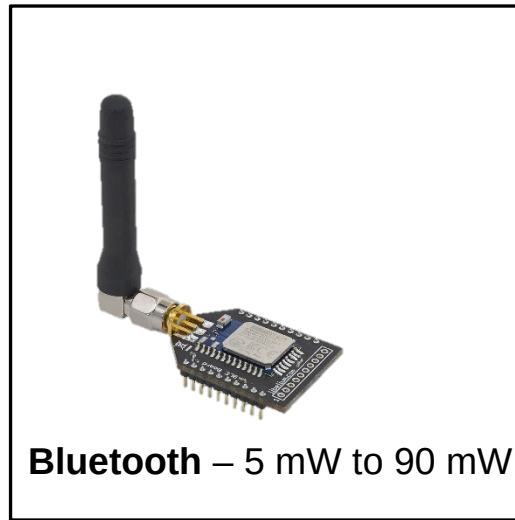
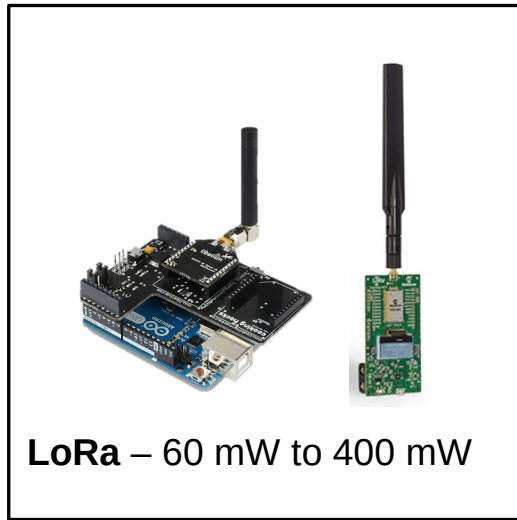


Type	AAA	CR2032	CR123A	CR2
Material	Alkaline	LiMnO ₂ *	Lithium	Lithium
Voltage	3V	3V	3V	3V
Capacity	500 mAh	225 mAh	1500 mAh	800 mAh

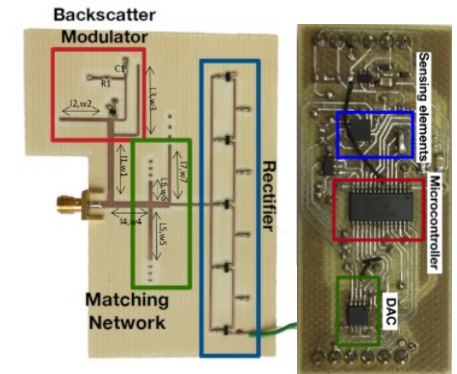


"Guilty as charged."

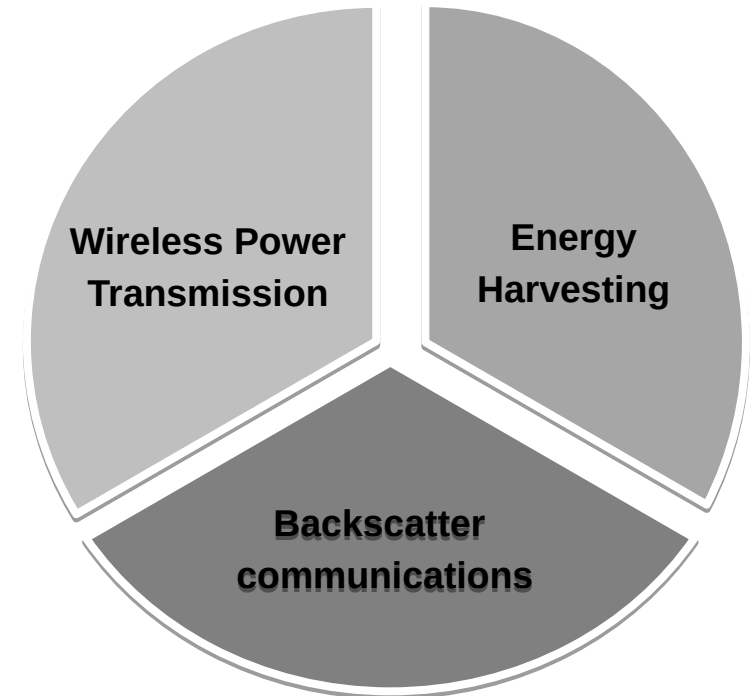
THE SOLUTION



- Remove the need of power for transmit / receive!
- Eliminate batteries!

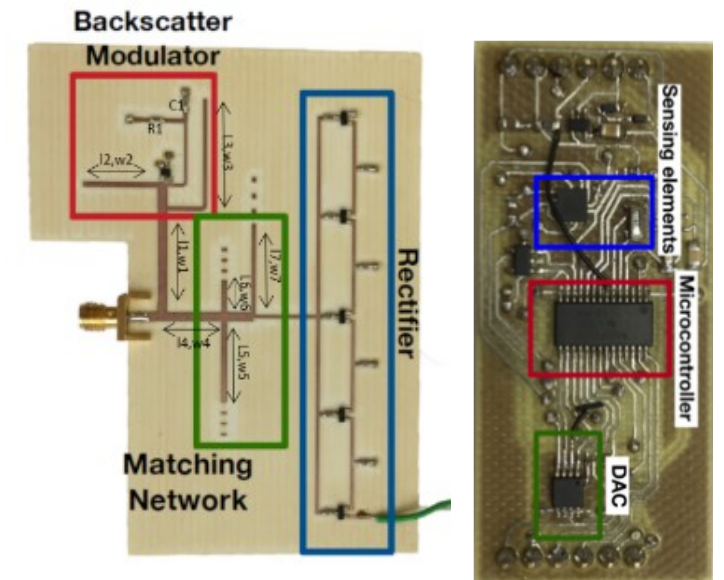
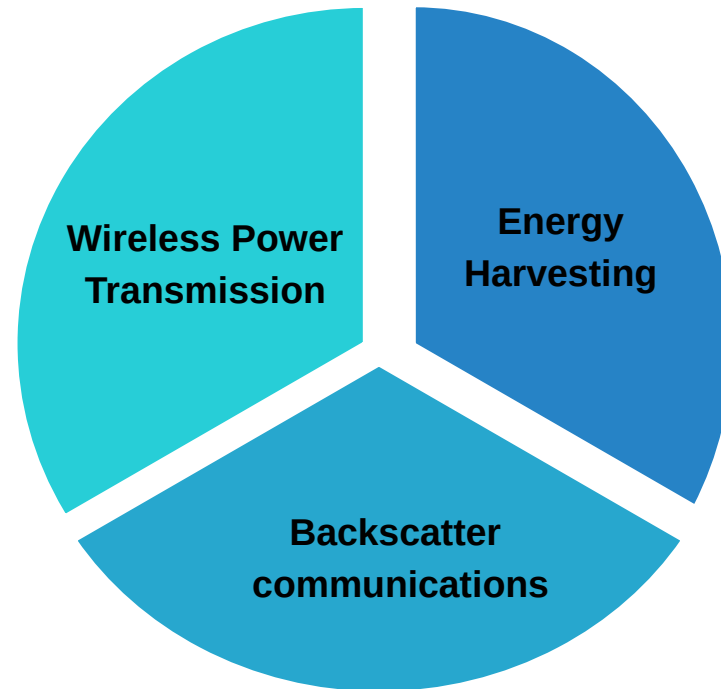


Passive



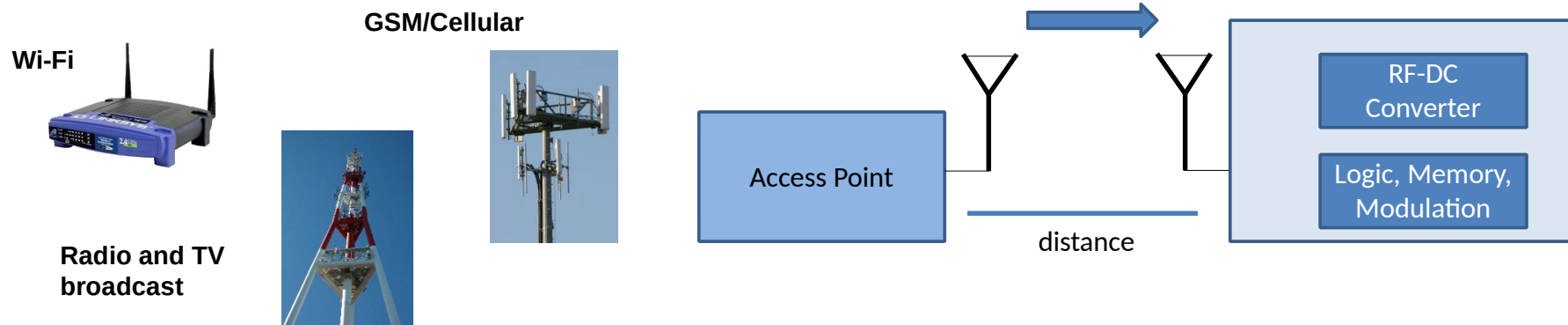
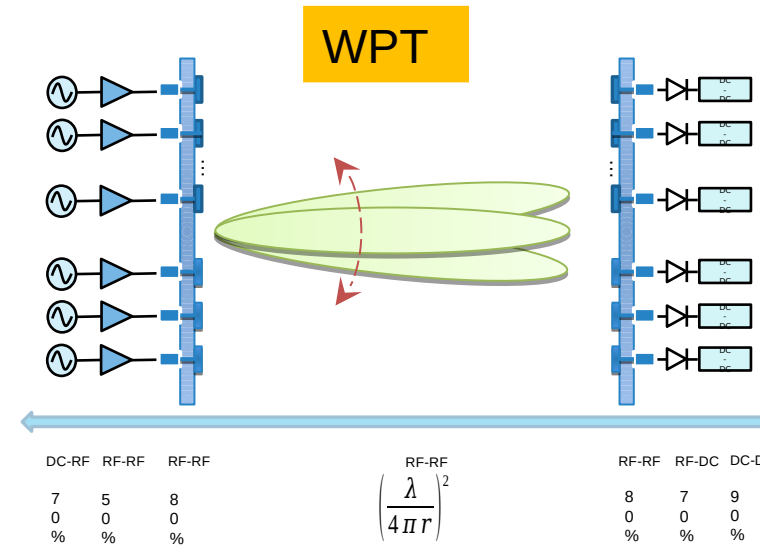
SWIPT

Simultaneous Wireless Information and Power Transfer



ENERGY HARVESTING VS WIRELESS POWER TRANSMISSION

Comparison of Power Density of Energy Harvesting Methods		
Energy Source	Power Density & Performance	Source of Information
Acoustic Noise	0.003 $\mu\text{W}/\text{cm}^3$ @ 75Db 0.96 $\mu\text{W}/\text{cm}^3$ @ 100Db	(Rabaey, Ammer, Da Silva Jr, Patel, & Roundy, 2000)
Temperature Variation	10 $\mu\text{W}/\text{cm}^3$	(Roundy, Steingart, Fr�chet, Wright, Rabaey, 2004)
Ambient RF	1 $\mu\text{W}/\text{cm}^2$	(Yeatman, 2004)
Ambient Light	100 mW/cm^2 (direct sun) 100 W/cm^2 (illuminated office)	Not Cited
Thermoelectric	60 W/cm^2	(Stevens, 1999)
Vibration (micro generator)	4 W/cm^3 (human motion - Hz) 800 W/cm^3 (machines - kHz)	(Mitcheson, Green, Yeatman, & Holmes, 2004)
Vibrations (Piezoelectric)	200 $\mu\text{W}/\text{cm}^3$	(Roundy, Wright, & Pister, 2002)
Airflow	1 $\mu\text{W}/\text{cm}^2$	(Holmes, 2004)
Push Buttons	50 J/N	(Paradiso & Feldmeier, 2001)
Shoe Inserts	330 $\mu\text{W}/\text{cm}^2$	(Shenck & Paradiso, 2001)
Hand Generators	30 W/kg	(Starner & Paradiso, 2004)
Heel Strike	7 W/cm^2	(Yaglioglu, 2002) (Shenck & Paradiso, 2001)

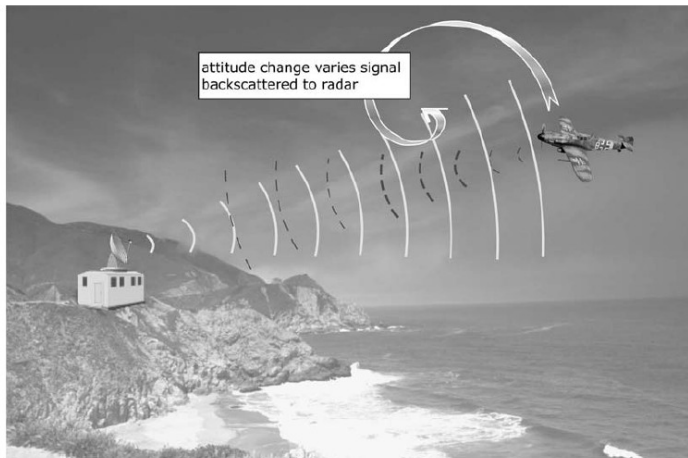




BACKSCATTER COMMUNICATIONS

HISTORY OF BACKSCATTER RADIO

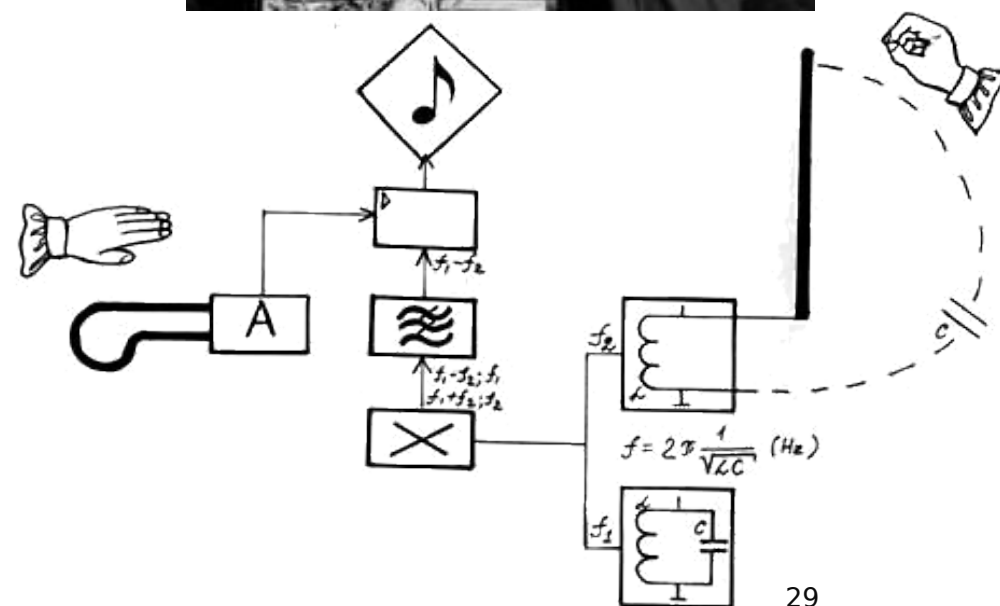
Backscatter RF tags have their origins in radar
- the origin of identification by power reflection



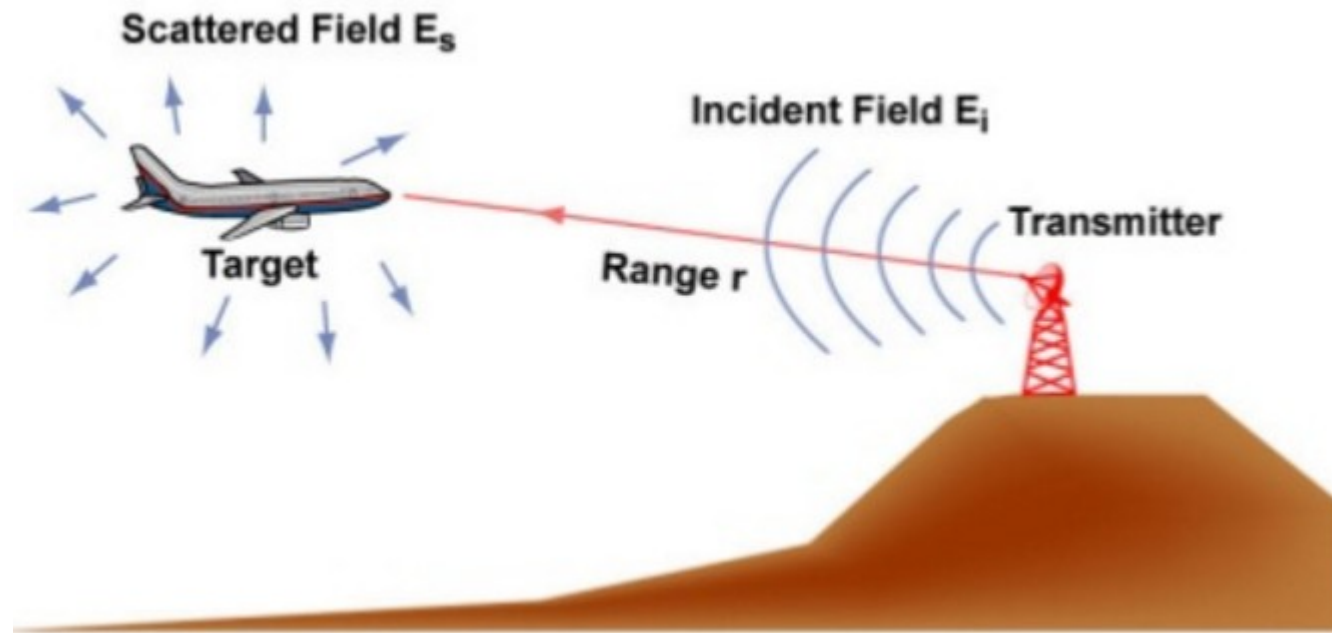
The IFF – Identification Friend or Foe system used during 2nd World War

- The foundation for RFID laid;
- Development of electronic article surveillance (EAS) systems (1-bit tags);
- RFID began finding mainstream commercial applications, in particular electronic toll collection. UHF and microwave tags still using discrete components.
- Development of useful Schottky diodes allow the entire tag to be integrated on a single chip.
- Establishment of widely accepted protocols – e.g. the Electronic Product Code protocols.

THEREMIN



RADAR CROSS SECTION



$$\text{RCS} = \lim_{r \rightarrow \infty} 4 \pi r^2 \frac{|E_s|^2}{|E_i|^2} \quad (\text{Unit: Area})$$

Radar Cross Section (RCS) is the hypothetical area, that would intercept the incident power at the target, which if scattered isotropically, would produce the same echo power at the radar, as the actual target.

FAR FIELD OPERATION

Contrary to inductive coupling, electromagnetic backscatter operates in the far field.

The range can be calculated based on the energy available at the transponder which is calculated using the Friis formula:

$$P_T = A_{e2} \frac{P_{in}}{4\pi r^2} G_1 = \frac{\lambda}{4\pi} G_2 \frac{P_{in}}{4\pi r^2} G_1 \Leftrightarrow P_T = P_{in} \left(\frac{\lambda}{4\pi r} \right)^2 G_t G_r$$

Table 3.7 Free space path loss a_F at different frequencies and distances. The gain of the transponder's antenna was assumed to be 1.64 (dipole), the gain of the reader's antenna was assumed to be 1 (isotropic emitter)

Distance r	868 MHz	915 MHz	2.45 GHz
0.3 m	18.6 dB	19.0 dB	27.6 dB
1 m	29.0 dB	29.5 dB	38.0 dB
3 m	38.6 dB	39.0 dB	47.6 dB
10 m	49.0 dB	49.5 dB	58.0 dB

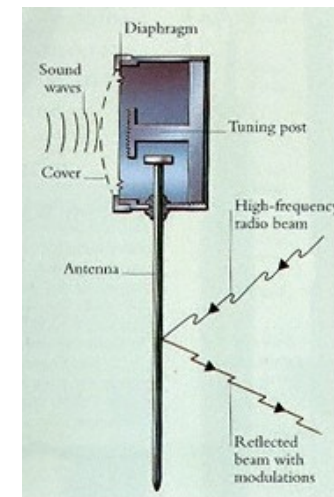
BACKSCATTER BATTERY-LESS PARADIGM



HISTORY OF BACKSCATTER RADIO

The Great Embassy Seal Bug

- Given as “gift” to US by USSR in 1946;
- Passive transduction of sound, interrogated from across the street in the Soviet Embassy;
- Undiscovered until 1952;
- Invented by Leon Theremin;
- Vibrating diaphragm changes capacitive load seen by antenna;
- Analog speech modulates the backscattered information;
- Reflected signal looks like small-carrier AM;

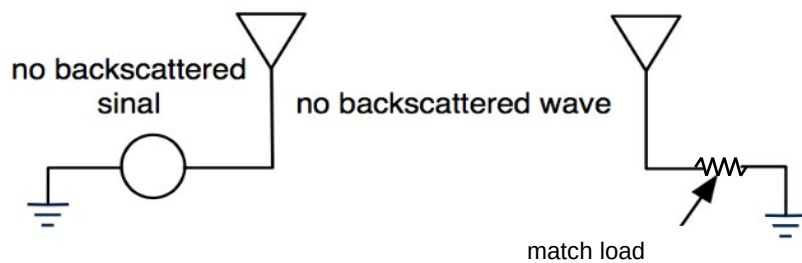
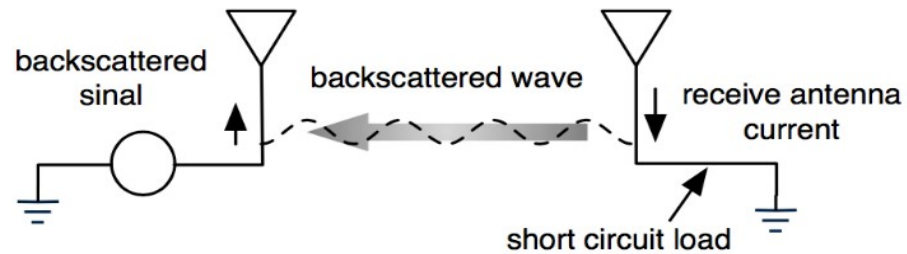
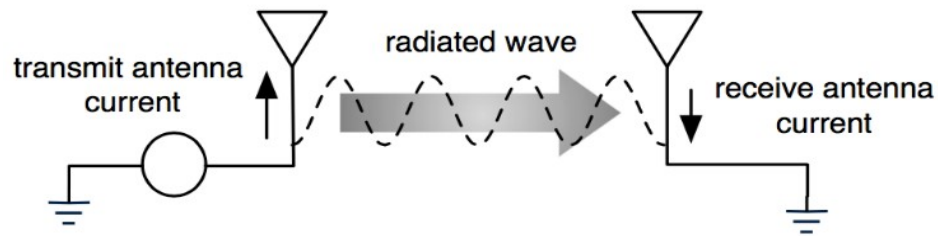


BACKSCATTER BATTERY-LESS PARADIGM



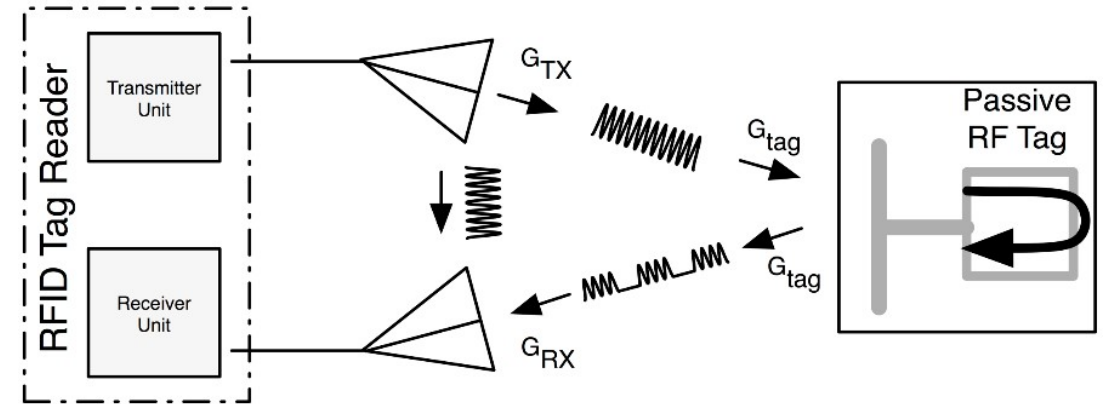
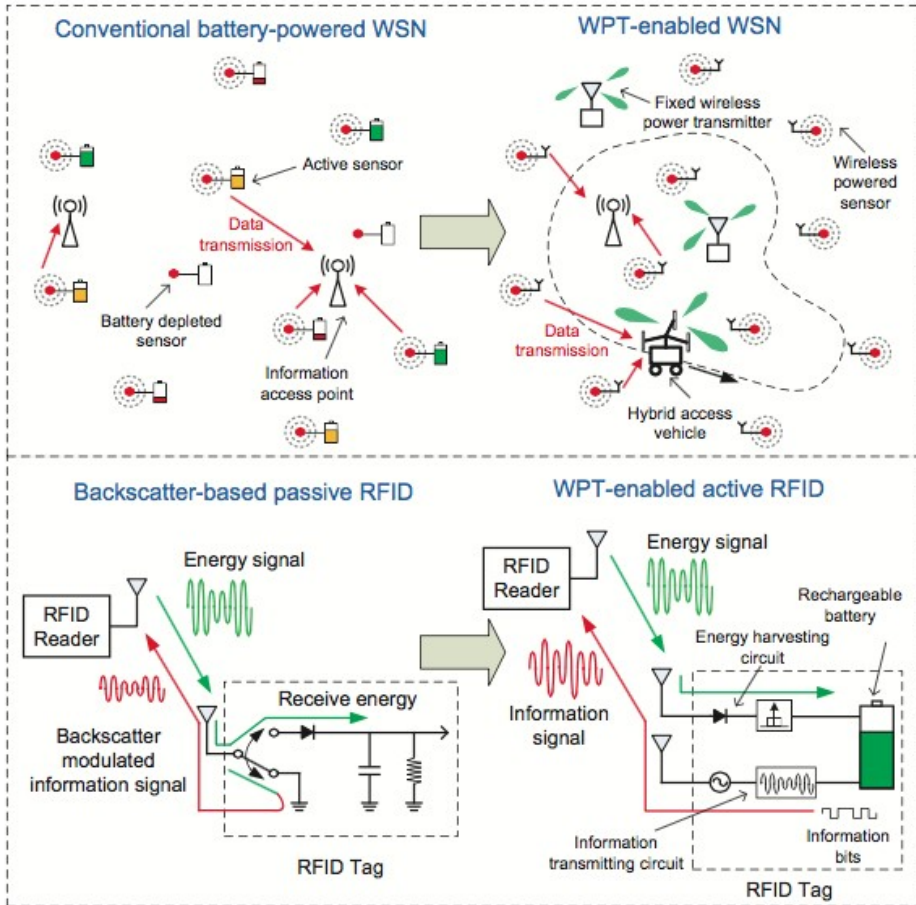
BACKSCATTER

BATTERY-LESS PARADIGM



- Backscatter communications has its roots in radars type approach
- The RF-front end reflects part of an incoming electromagnetic wave back to the reader based on the information pattern
- The backscatter reflection efficiency is maximized for antennas that are resonating with the incoming signal frequency

BACKSCATTER



RF tag communicates with a reader, by modulating the electromagnetic fields scattered from the RF tag's antenna.

Problems

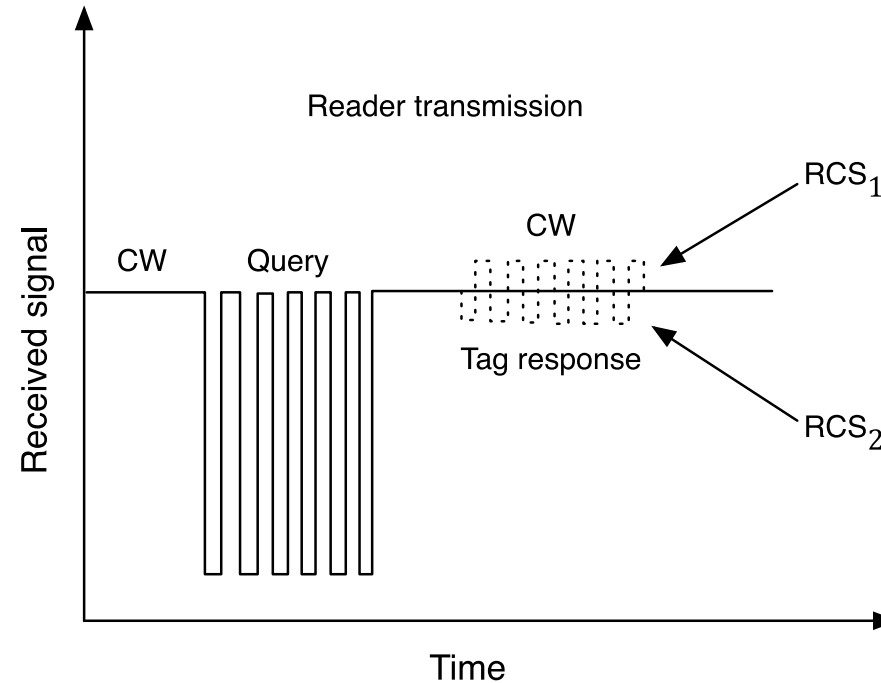
- Limited range communication
- ASK or PSK modulation schemes (one bit of data per symbol)

Advantages

- No need of active components in wireless transceivers
- Low-power implementation

Figure taken from S. Bi, C. K. Ho, and R. Zhang, "Wireless powered Communication Networks: An Overview,"

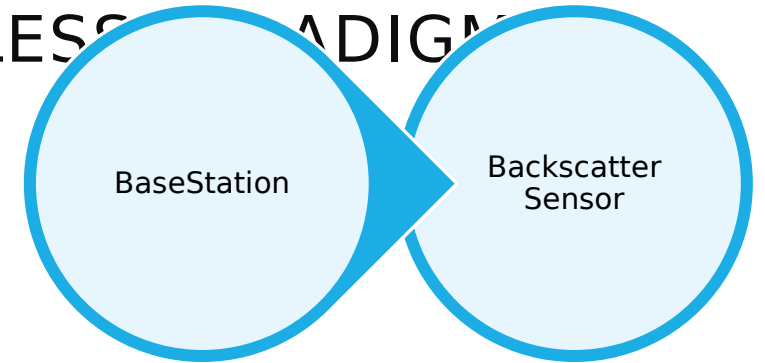
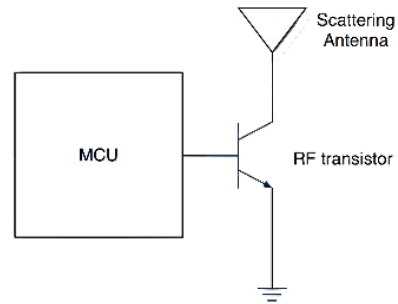
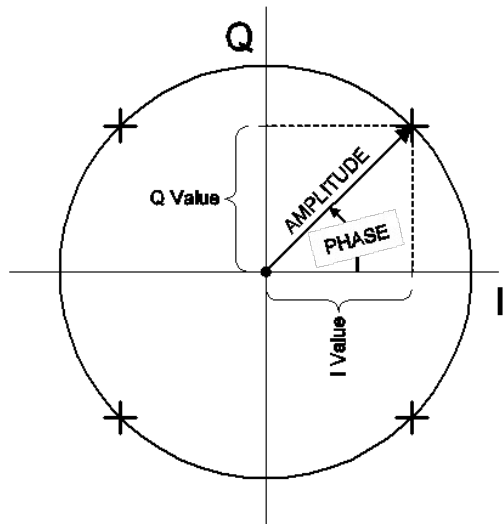
TRANSMIT AND RECEIVE SIGNALS



Signal transmitted on the forward link (reader to tag) contains both CW and modulated commands.

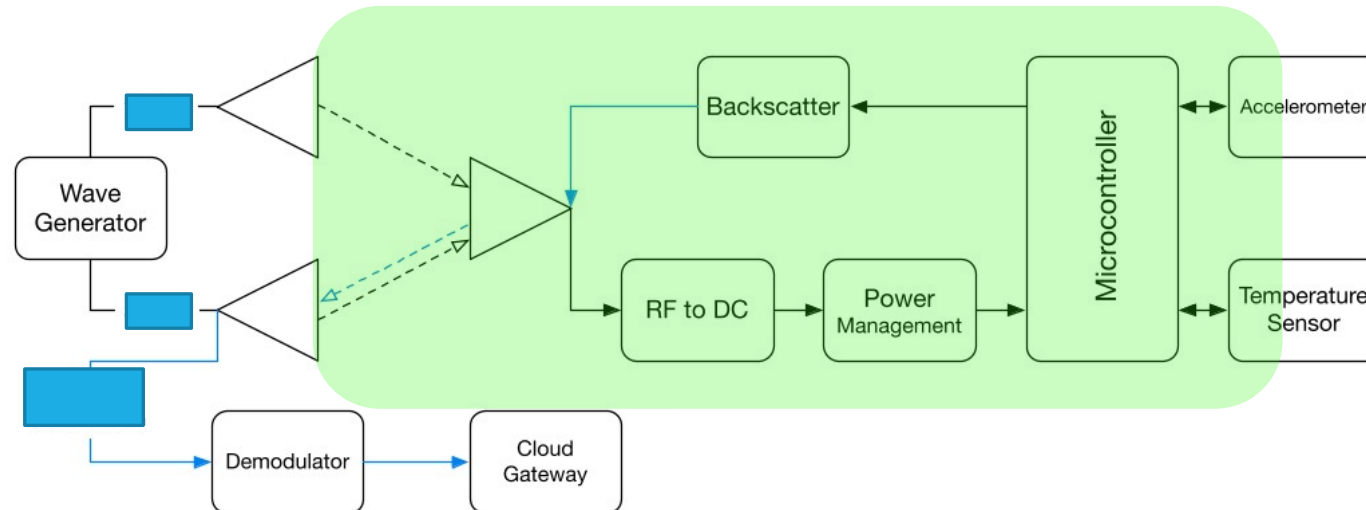
Each impedance state the tag presents a certain RCS high (RCS_1) and low (RCS_2) to provide significant difference in the backscattered signal.

BACKSCATTER BATTERY-LESS RADIGM

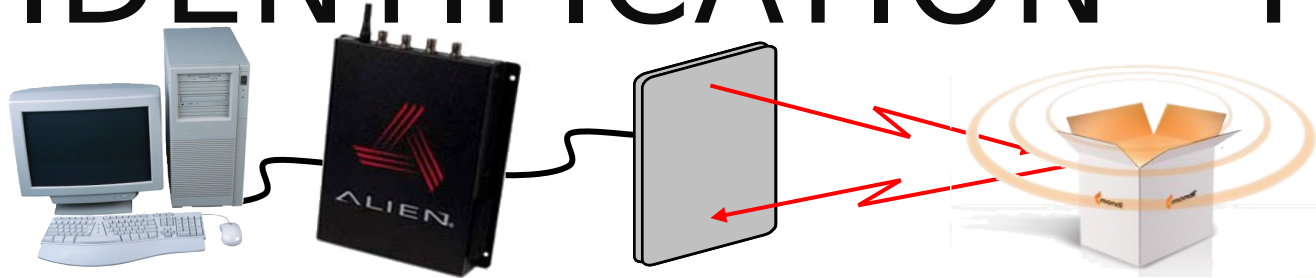


- RF Transmitter
- RF Receiver

- Voltage Harvester
- Signal Processor



RADIO FREQUENCY IDENTIFICATION - RFID



Host

Reader module

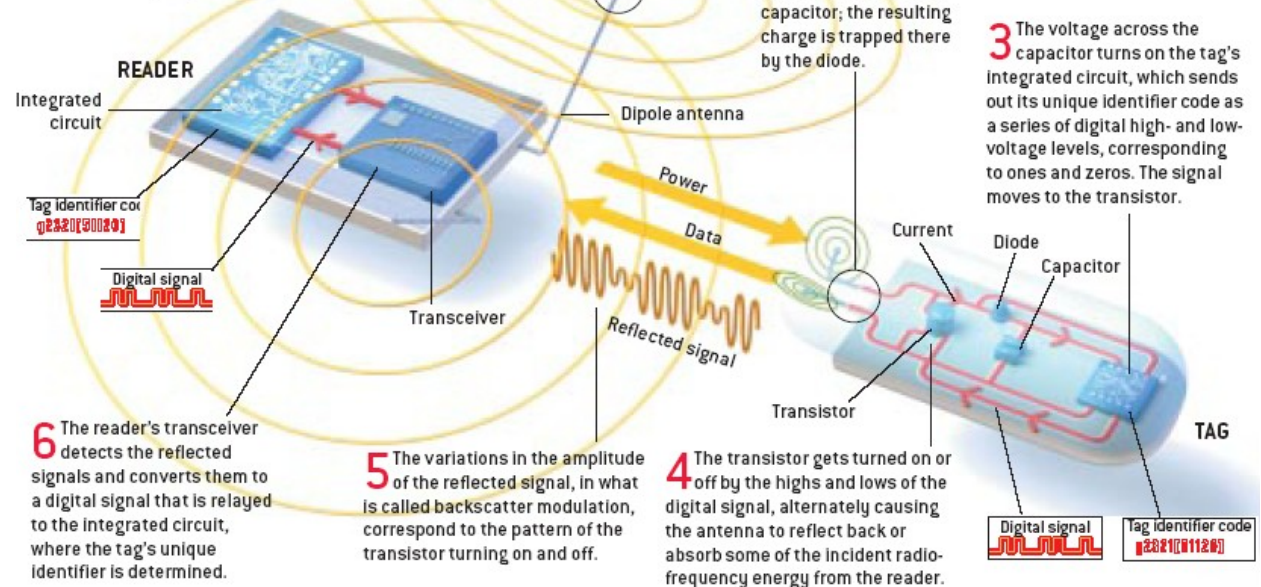
Antenna

HIGH-FREQUENCY SYSTEM

1 An integrated circuit sends a digital signal to a transceiver, which generates a radio-frequency signal that is transmitted by a dipole antenna.

2 The electric field of the propagating signal gives rise to a potential difference across the tag's dipole antenna, which causes current to flow into the capacitor; the resulting charge is trapped there by the diode.

3 The voltage across the capacitor turns on the tag's integrated circuit, which sends out its unique identifier code as a series of digital high- and low-voltage levels, corresponding to ones and zeros. The signal moves to the transistor.



6 The reader's transceiver detects the reflected signals and converts them to a digital signal that is relayed to the integrated circuit, where the tag's unique identifier is determined.

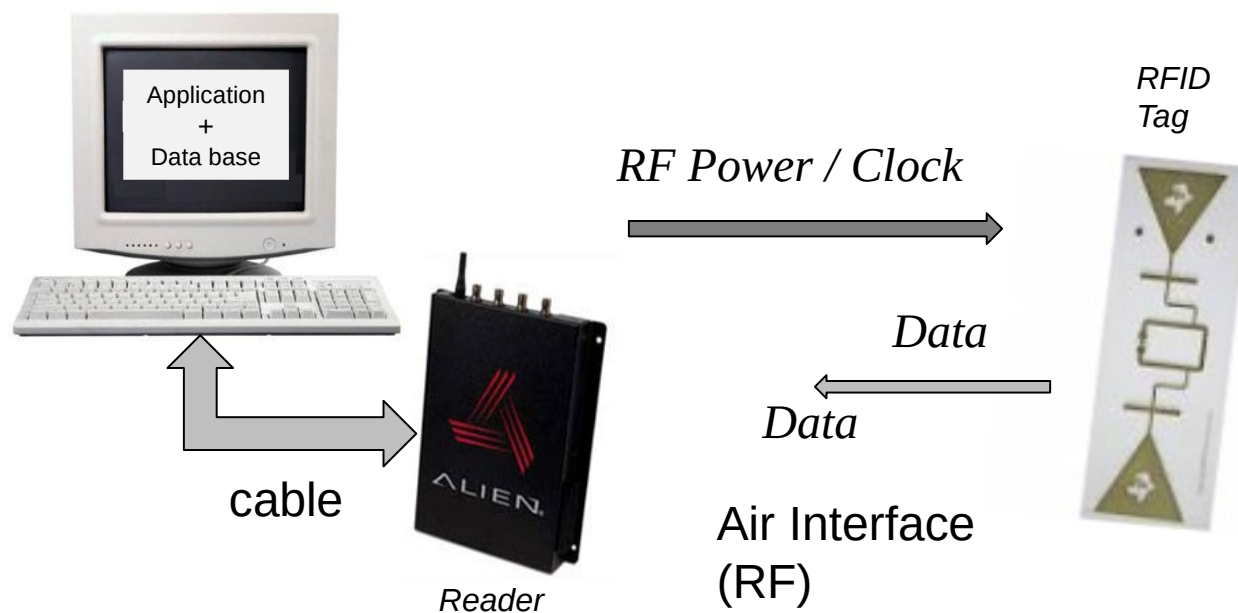
5 The variations in the amplitude of the reflected signal, in what is called backscatter modulation, correspond to the pattern of the transistor turning on and off.

4 The transistor gets turned on or off by the highs and lows of the digital signal, alternately causing the antenna to reflect back or absorb some of the incident radio-frequency energy from the reader.

PRINCIPLE OF OPERATION

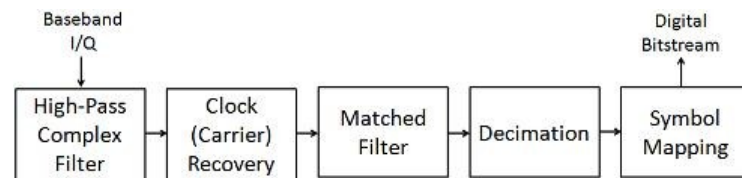
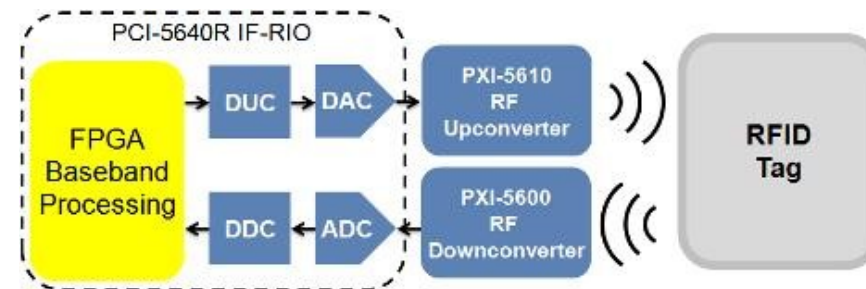
Basic components

- 1) **Reader/Interrogator** – Used for read and store information in Tag
- 2) **Tag/Transponder** – small device which carries data (e.g. Tag ID)
- 3) **Host** - computer running user application



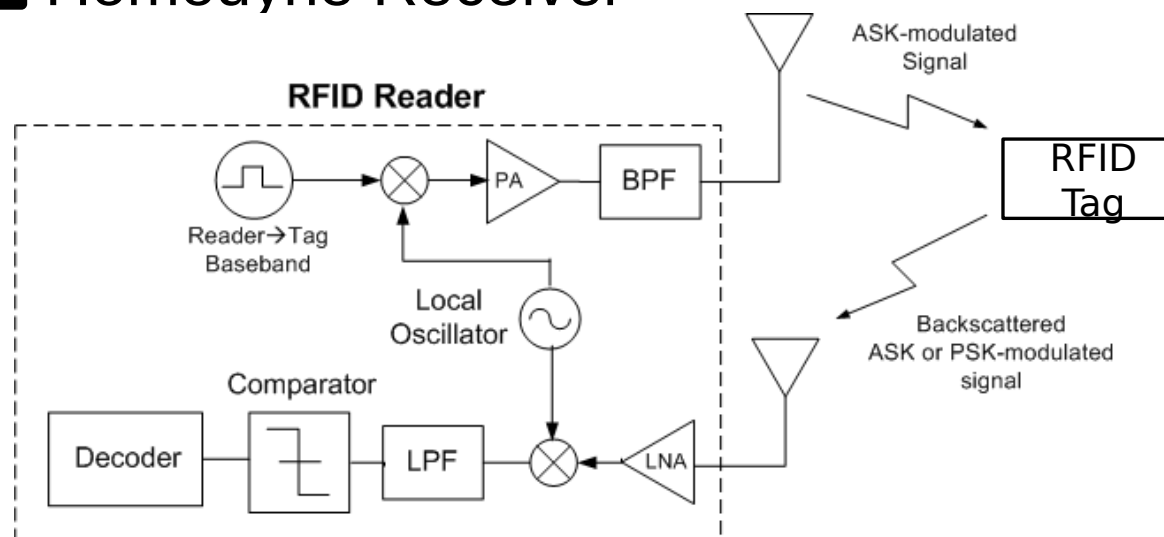
UHF RFID

Typical Readers are moving fast to a Software Defined Radio Solution, including a digital part and a RF up-converter and RF down-converter....



READER ARCHITECTURES

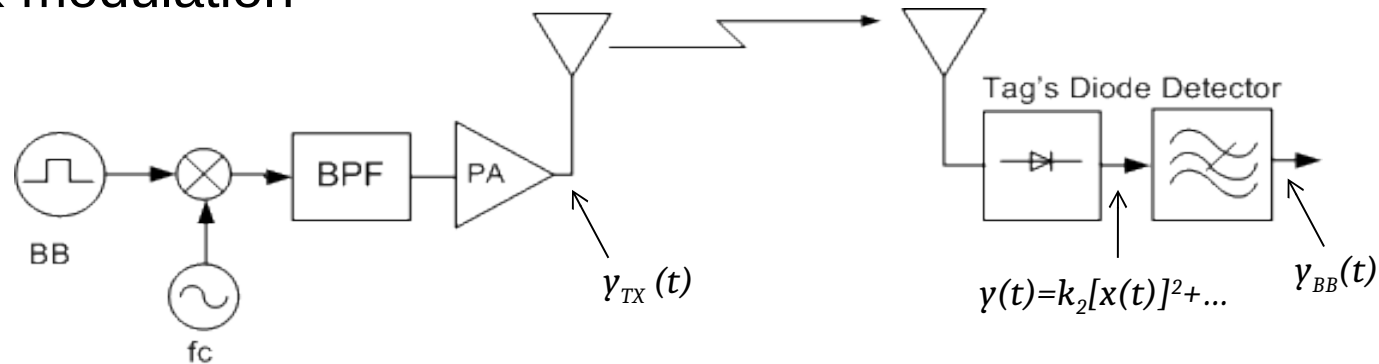
- ✓ Transmitter ✉ ASK Modulator
- ✓ Receiver ✉ Homodyne Receiver



DOWNLINK DATA COMMUNICATION

Downlink: Rader Tag communication

- ASK modulation



Baseband ASK signal:
$$m(t) = \sum_{k=-\infty}^{+\infty} a_k \delta(t - kT)$$

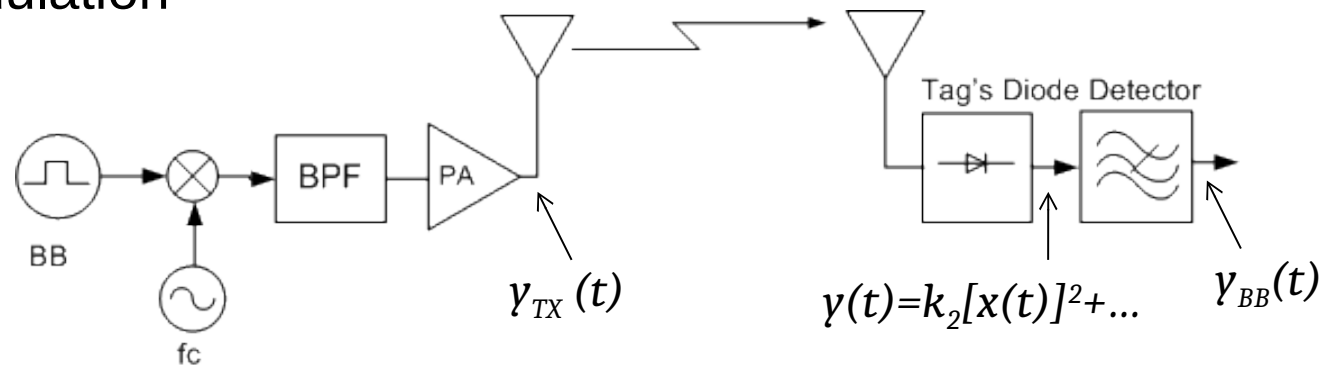
Transmitted Signal:
$$y_{TX}(t) = m(t) \left\{ A_C \frac{e^{j\omega_c t} + e^{-j\omega_c t}}{2} \right\}$$

RF Carrier

DOWNLINK DATA COMMUNICATION

Downlink: Rader ✉ Tag communication

- ASK modulation

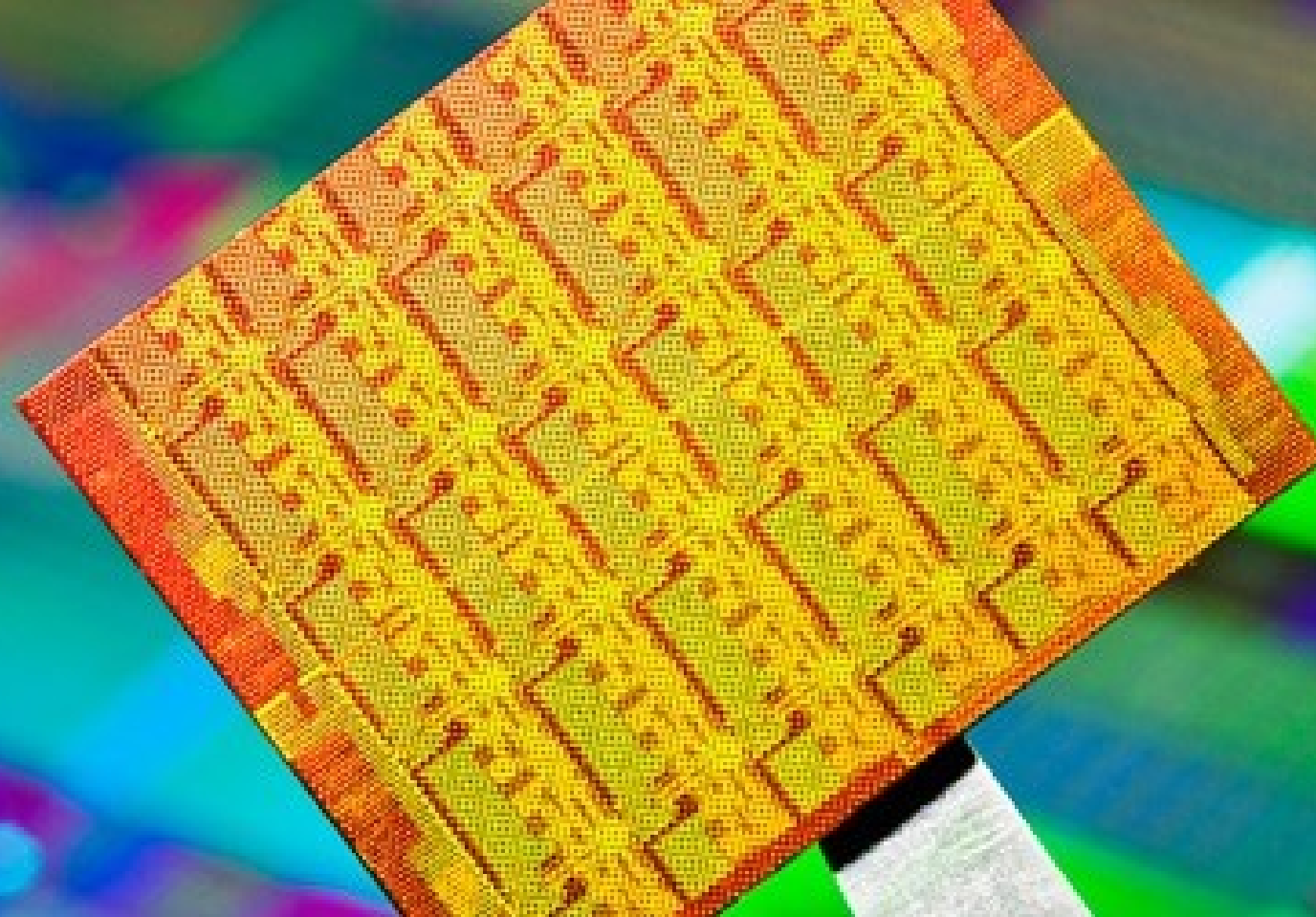


Envelope Demodulation:

$$y(t) = k_2 \left[m(t) \left\{ A_c \frac{e^{j\omega_c t} + e^{-j\omega_c t}}{2} \right\} \right]^2$$

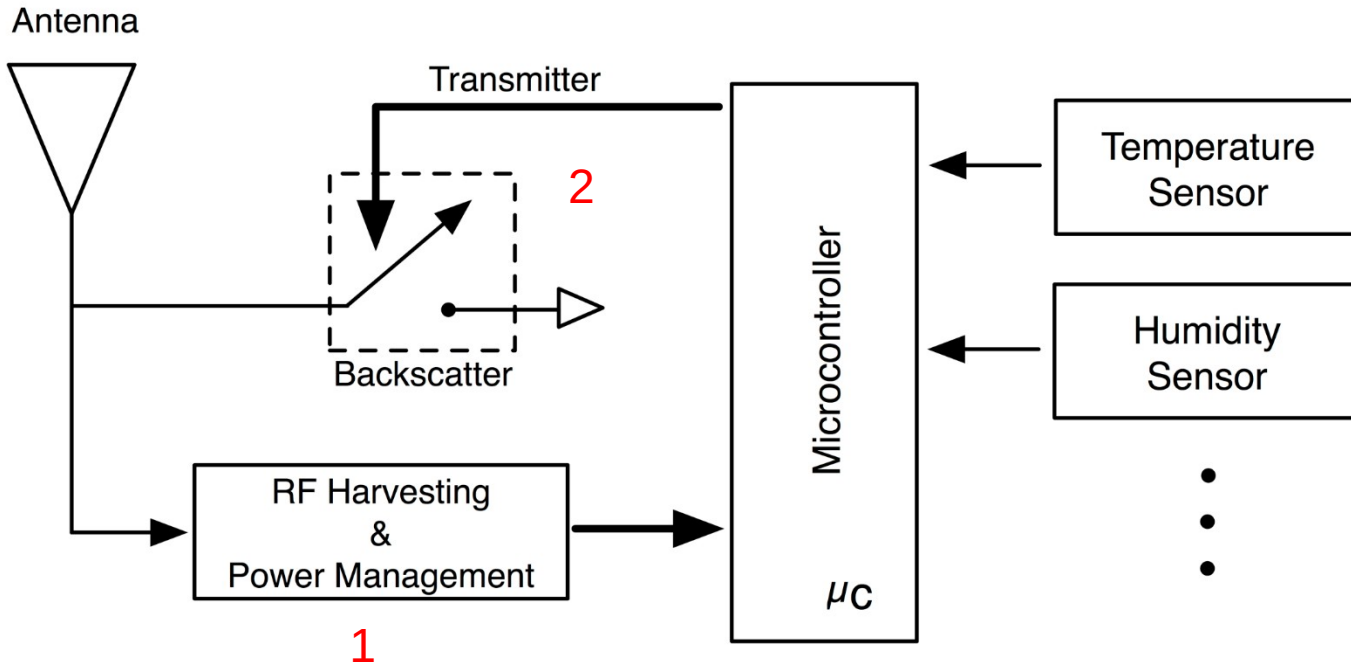
After low-pass filtering, the baseband signal sent by the reader is recovered by the Tag:

$$y_{BB}(t) = \frac{A_c^2 k_2}{2} [m(t)]^2$$



HOW TO DESIGN A MICROWAVE BACKSCATTER TAG

BACKSCATTER - STEPS FOR IMPLEMENTATION



Backscatter design:

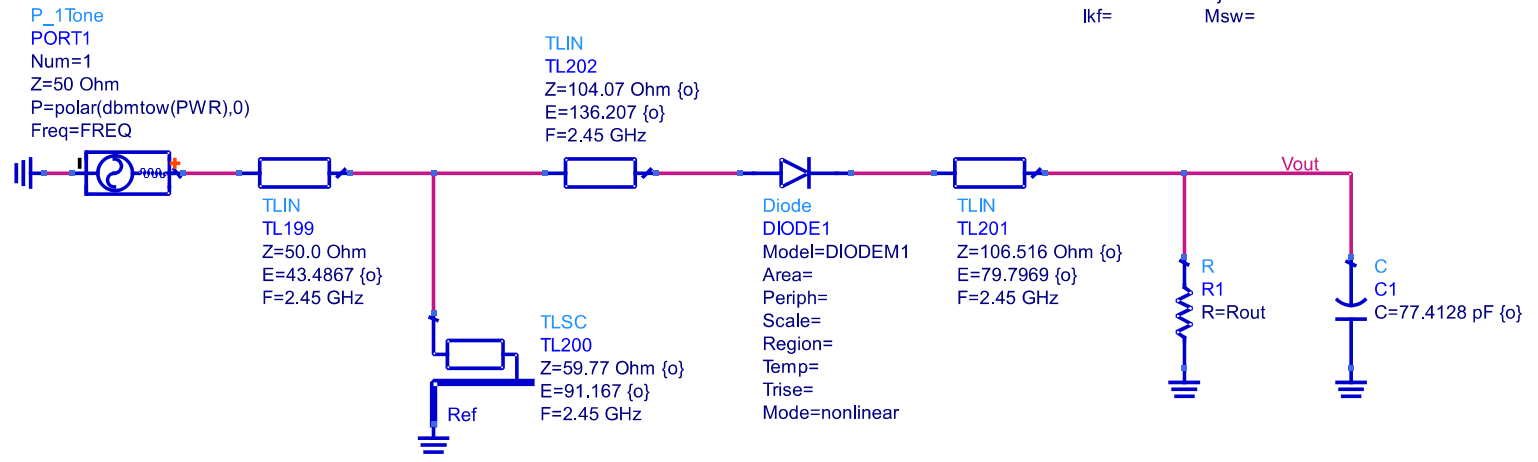
- Choose the frequency for the circuit; **2.45 GHz**
- Design the rectifier for the chosen frequency and optimize it for the highest efficiency! **1**
- Design the backscatter modulator; **2**
- Combine both solutions (Rectifier + Backscatter modulator).

DESIGN THE RF-DC FOR 2.45 GHz

MeasEqn
Meas1
Effn=100*(sqrt(real(Vout[:,:0]))/(Rout*(0.001*(10**(PWR/10))))))

Diode_Model
DIODEM1
Is=3e-6 Bv=3.8 V Vjsw=
Rs=25 Ibv=0.3 mA Fcsw=
Gleak= Nbv= AllowScaling=no
N=1.06 Ibvl= Tnom=
Tt= Nbv1= Trise=
Cd= Kf= Xti=2
Cjo=0.18 pF Af= Eg=0.69
Vj=0.35 V Ffe= AllParams=
M=0.5 Jsw=
Fc= Rsw=
Imax= Gleaksw =
Imelt= Ns=
Isr= Ikp=
Nr= Cjsw=
Ikf= Msw=

Set these values:
VAR
STIMULUS
volt=0 {-t}
Rout=4.45781 kOhm {o}
PWR=-10_dbm
FREQ=2450 MHz

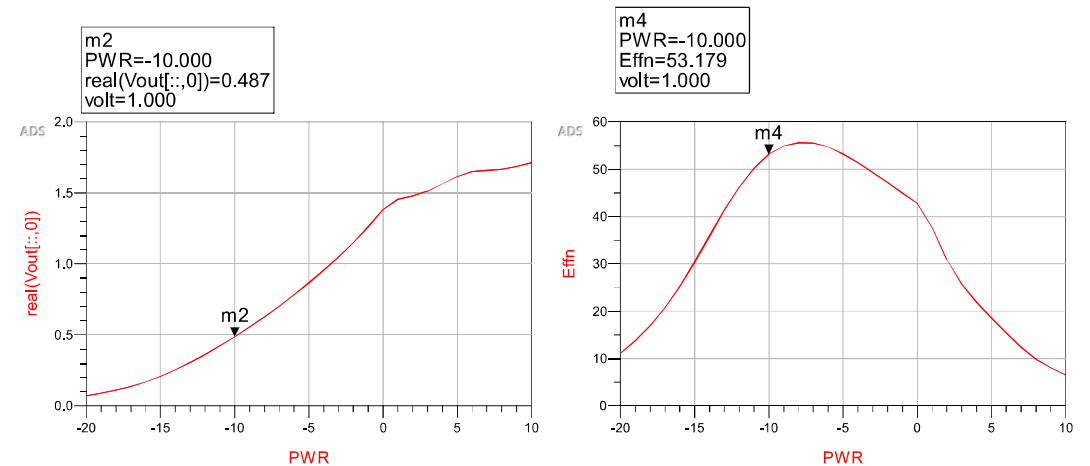
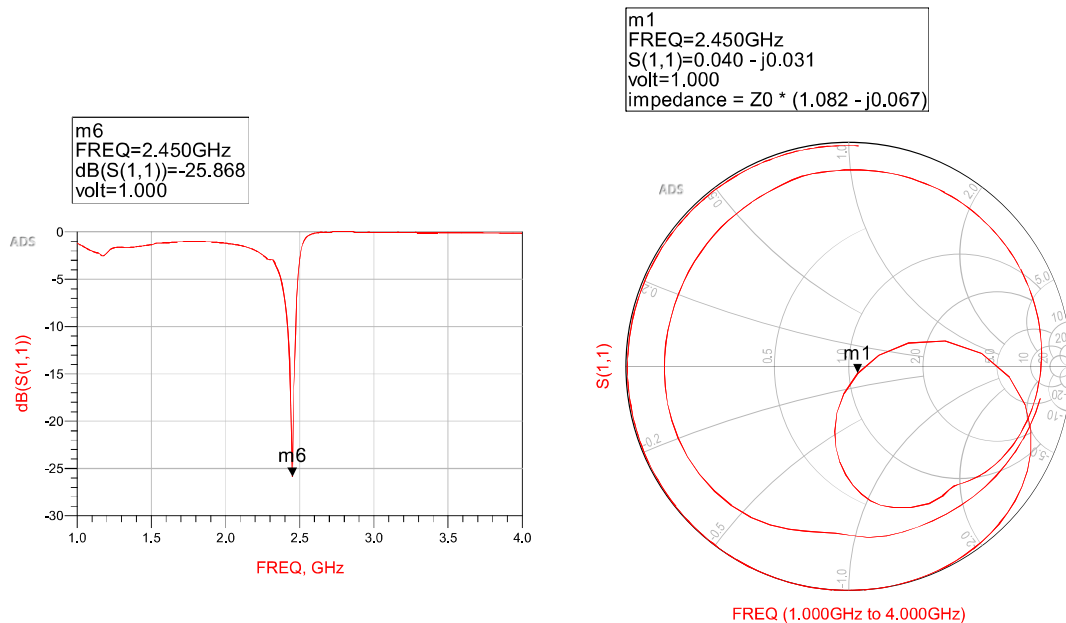


RESULTS OF RF-DC

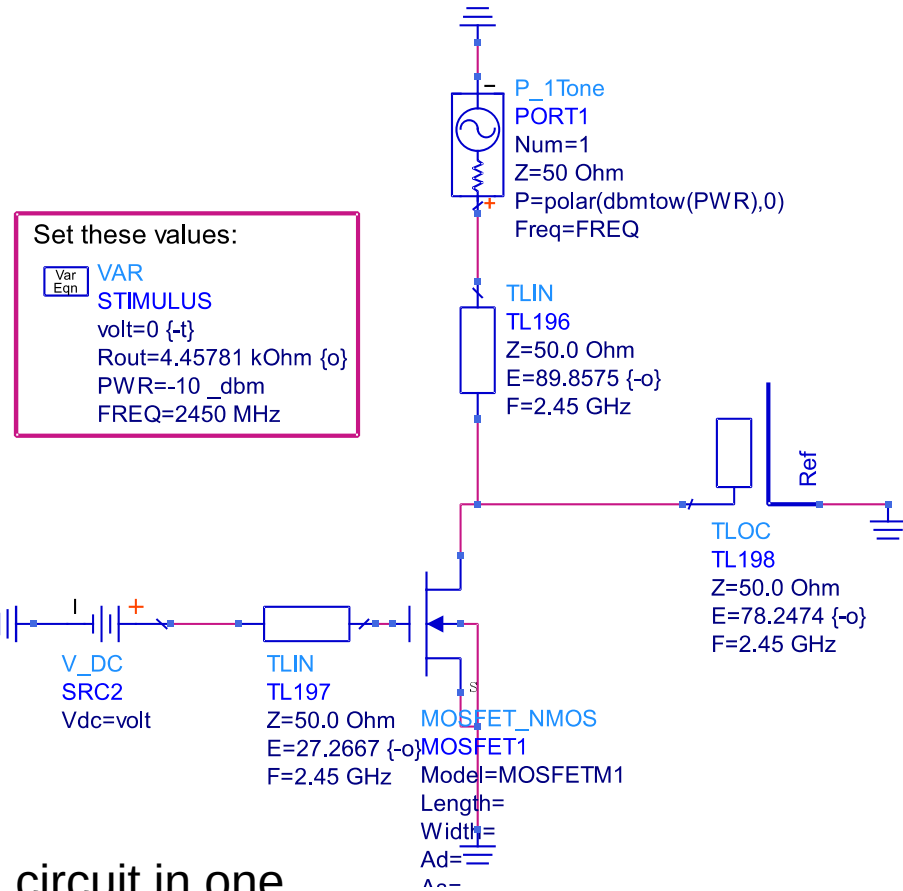
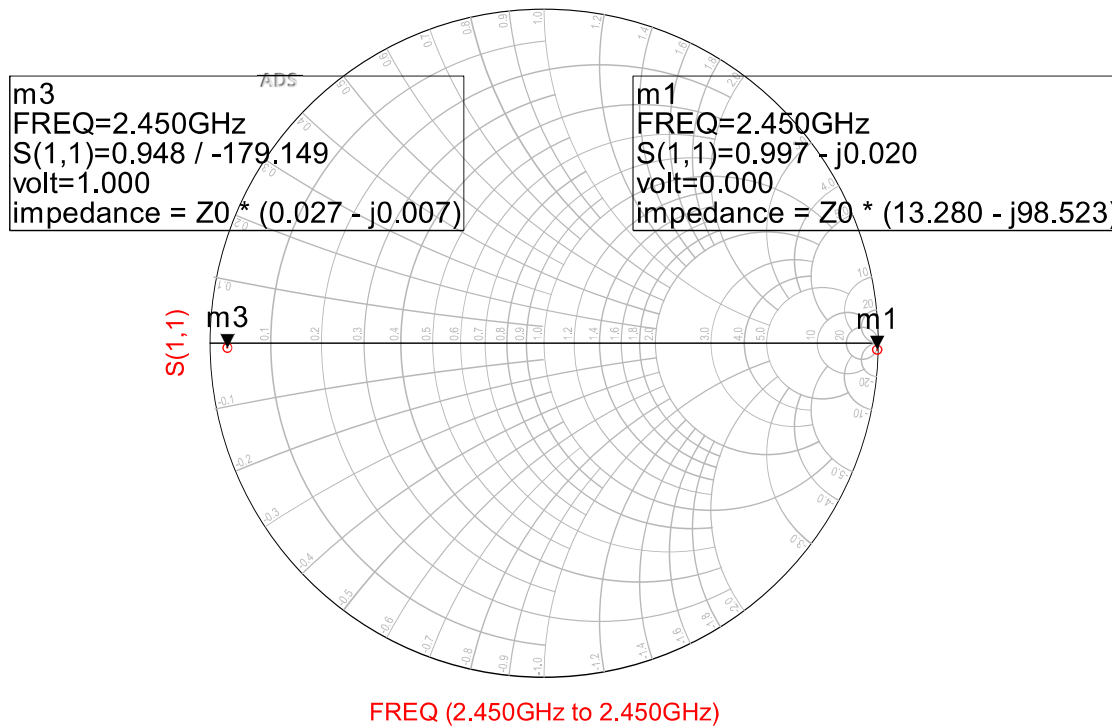
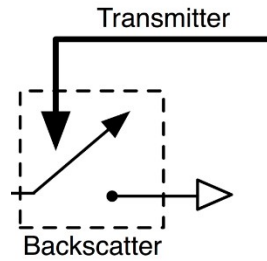
Using large signal S-parameters

Matching Frequency = 2.45 GHz

For -10 dBm (0.1 mW) of input power the output voltage generated is 0.487 V with 53.2 % of efficiency

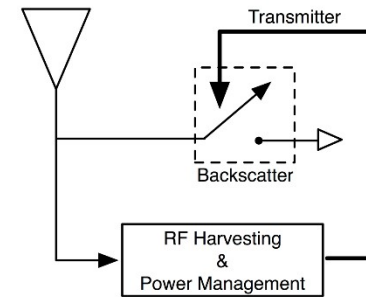
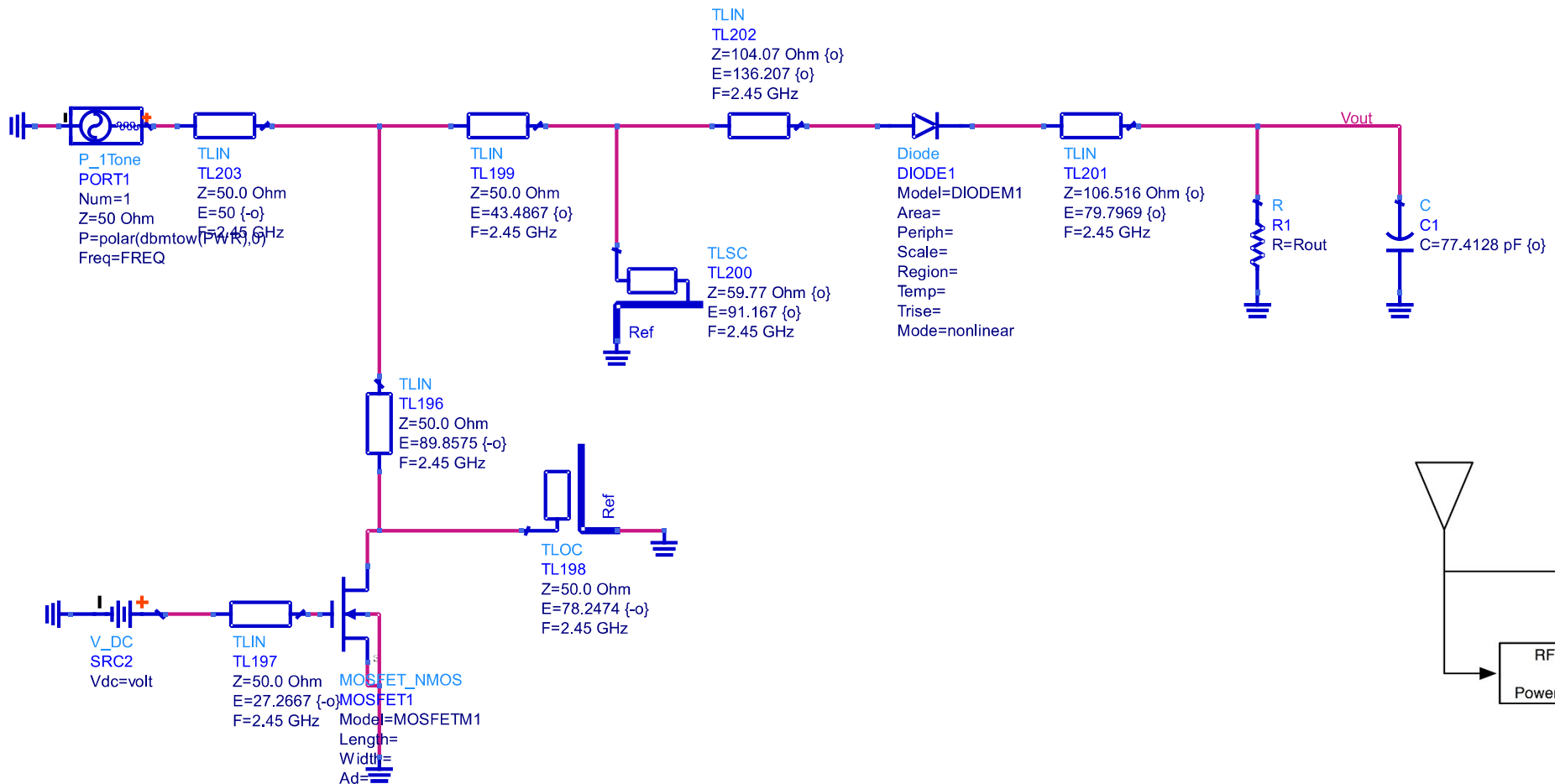


DESIGN OF BACKSCATTER MODULATOR

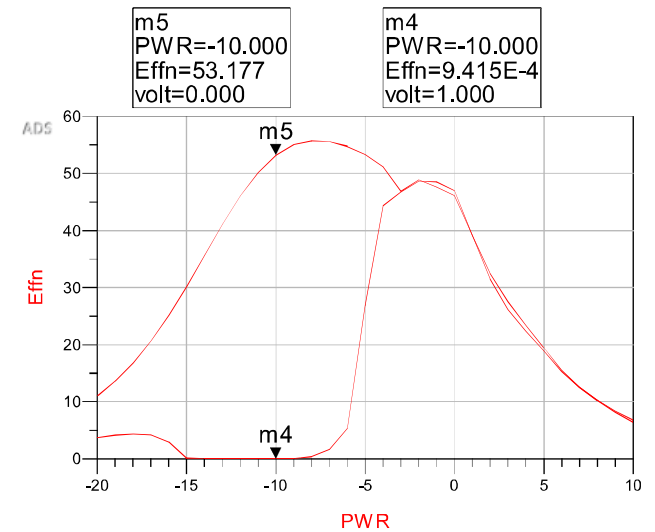
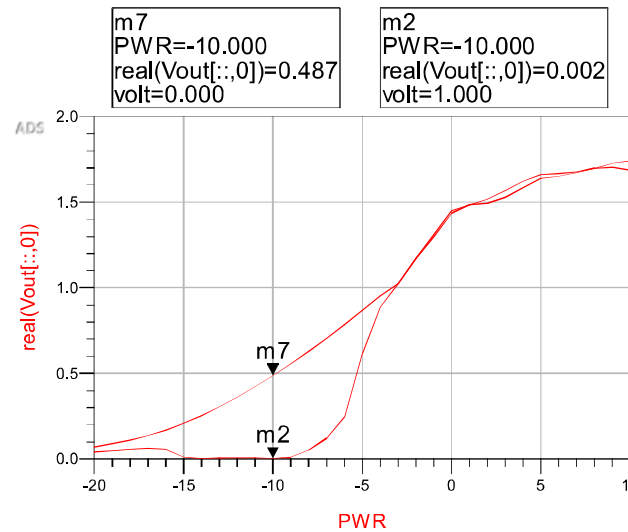
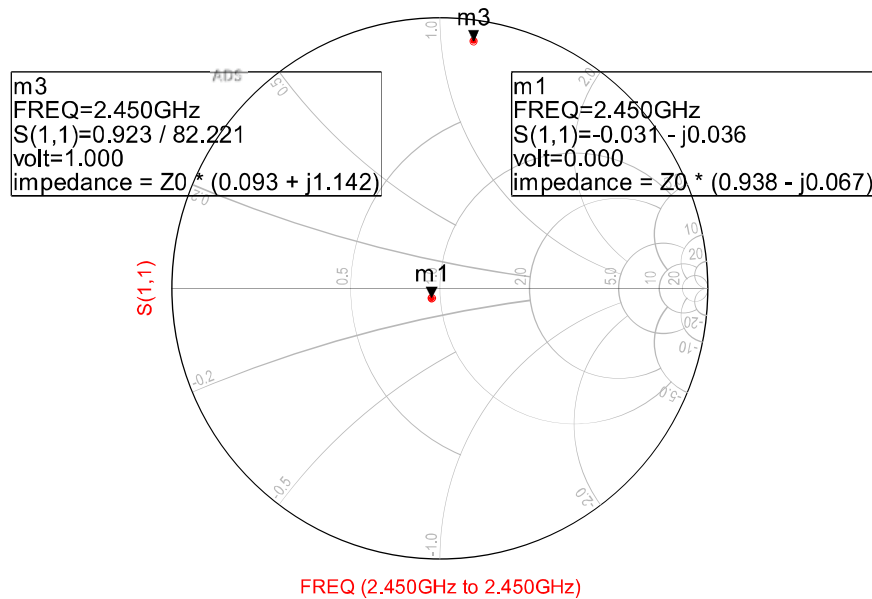


The goal is to optimize the modulator to behave as an open circuit in one state of transistor (0 V) and as a short circuit in other state (1 V).

COMBINING SOLUTIONS – BACKSCATTER DESIGN



RESULTS OF BACKSCATTER CIRCUIT



Two different loads

- The circuit is matched in one state (0 V) – **absorption state**;
- The circuit is mismatched in other state (1 V) – **reflection state**.

For -10 dBm (0.1 mW) of input power the output voltage generated is 0.487 V with 53.2 % of efficiency in the absorption state!

In the reflection state, the circuit can not generate any output voltage!

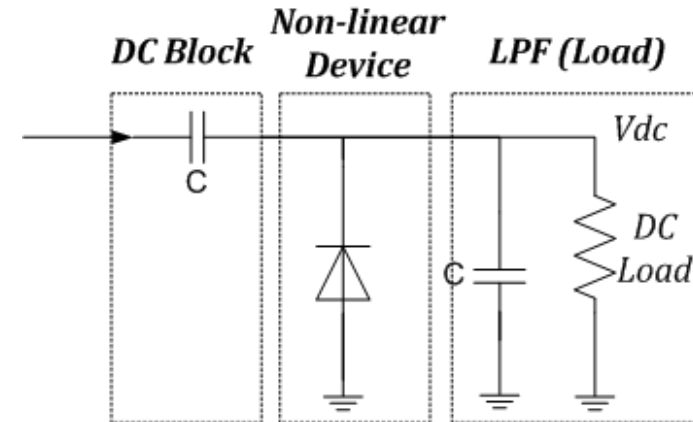
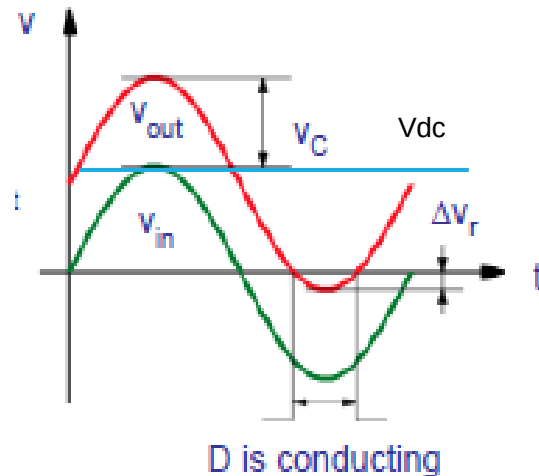
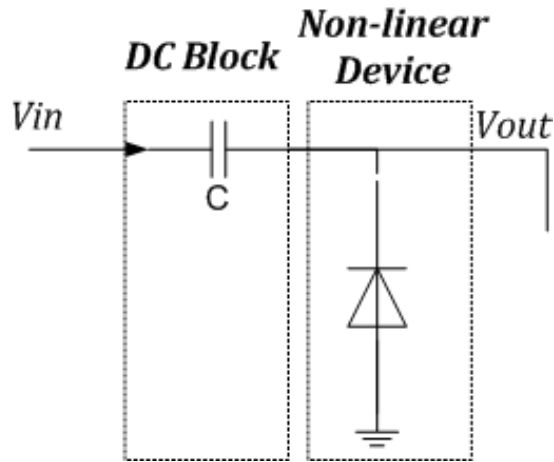


WIRELESS POWER TRANSMISSION

WIRELESS POWER TRANSMISSION

- ✓ Typically High Speed Schottky Diodes are used in RF-DC converters
- ✓ Commonly used configurations: single-diode detectors (**high RF-DC efficiency**), voltage multipliers (**high voltage**), full-wave rectifiers (current stability), ...

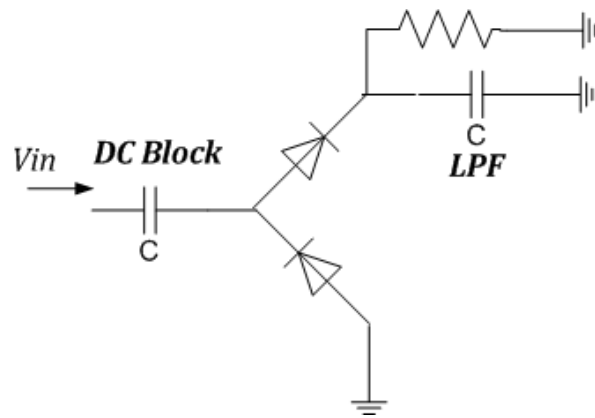
Single-diode Envelope Detector



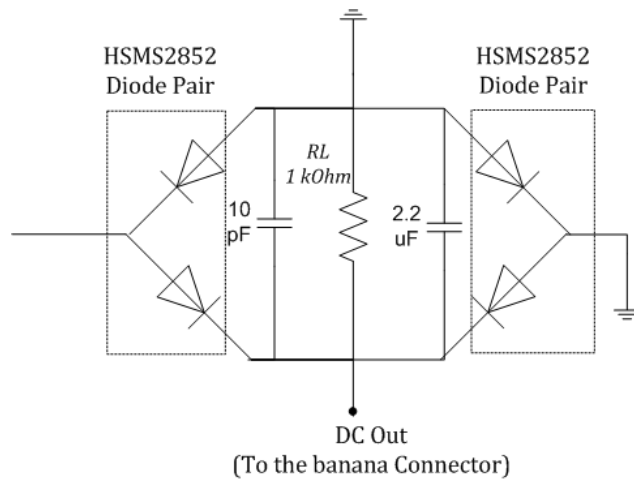
WIRELESS POWER TRANSMISSION

Half-wave Voltage Multipliers (Charge Pumps)

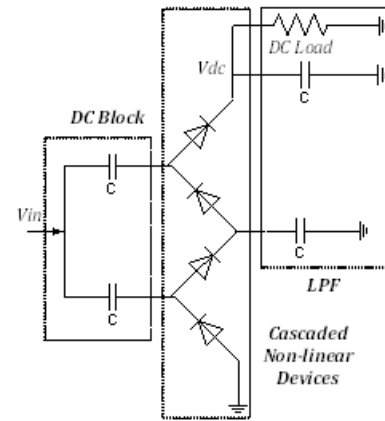
Voltage Doubler (1-stage)



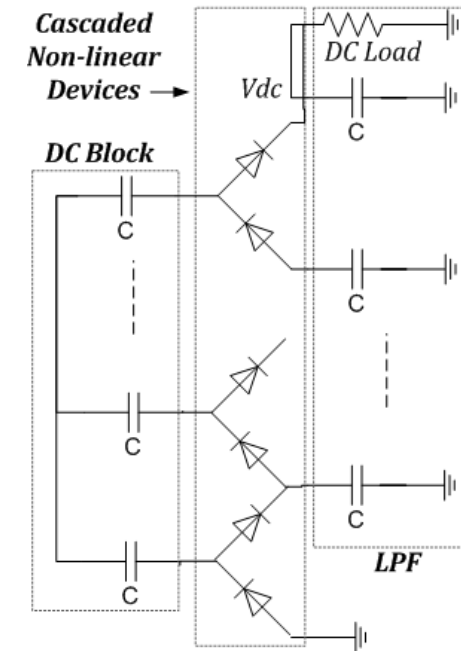
Full-wave Rectifiers



2-stages

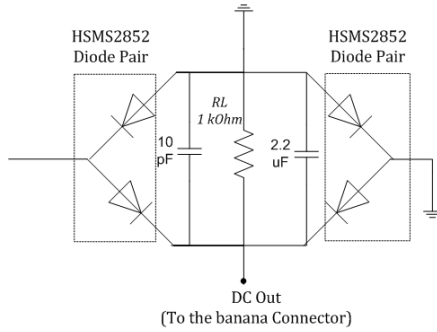
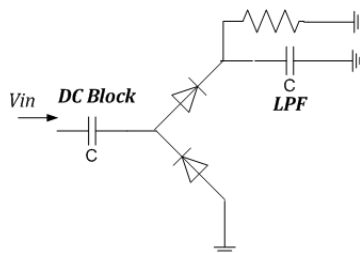
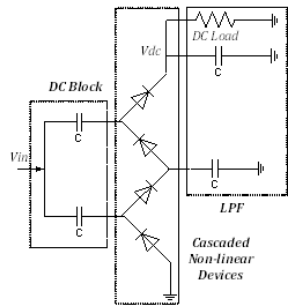


N-stages

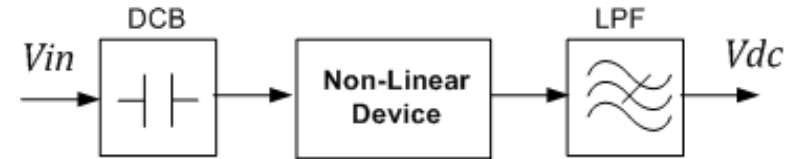
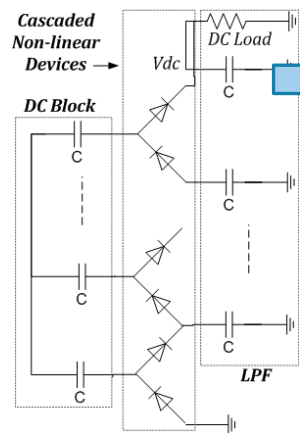


WIRELESS POWER TRANSMISSION

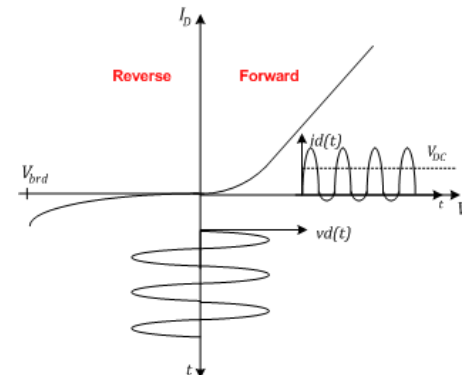
Non-linear Analysis of RF-DC Converters



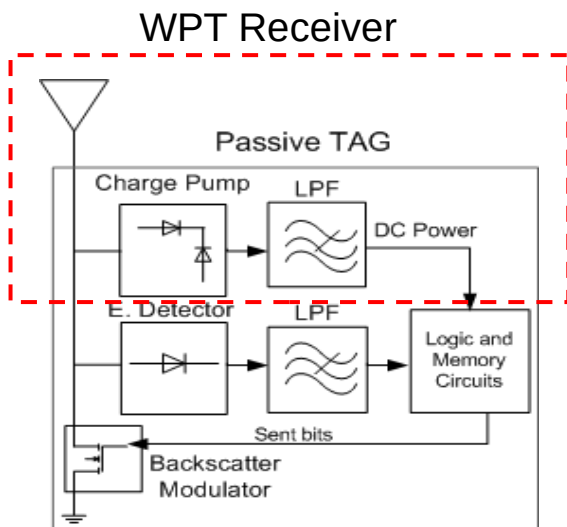
Any RF-DC Converter can be described by a non-linear Model:



Diode Non-linear Characteristic (I-V Curve)



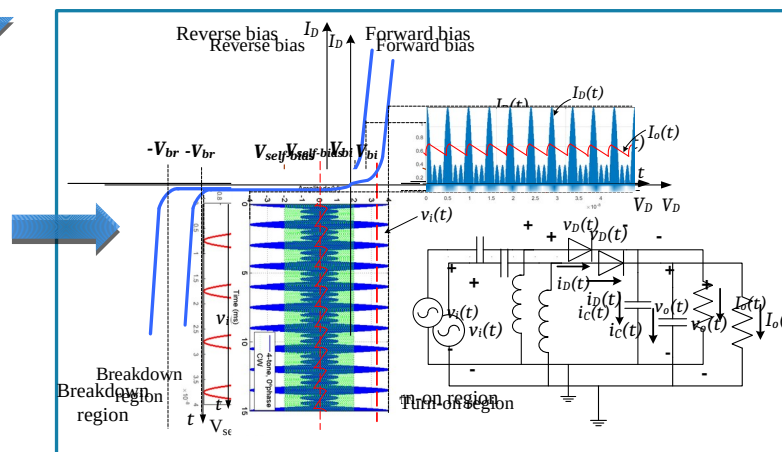
SENSOR RF-DC CONVERTER



Desired component

$$V_{out} = NL[x_{in}(f_0)] = Y(DC) + Y(f_0) + Y(2f_0) + Y(3f_0) + \dots + Y(nf_0)$$

RF-DC conversion process in a diode detector



$$I_D = I_S \left(e^{\frac{qV_j}{\eta kT}} - 1 \right) = I_S \left(e^{\frac{V_j}{\eta V_t}} - 1 \right) = I_S \left(e^{\frac{V_D - R_S I_D}{\eta V_t}} - 1 \right)$$

Rectifying devices exhibit a NON-ZERO turn-on voltage **☑ a certain amount of energy is needed to overcome the turn-on voltage** **☑ low power level efficiency is degraded**



WIRELESS POWER COMMUNICATIONS

BACKSCATTER BATTERY-LESS PARADIGM

Access Point-to-sensor link has two goals

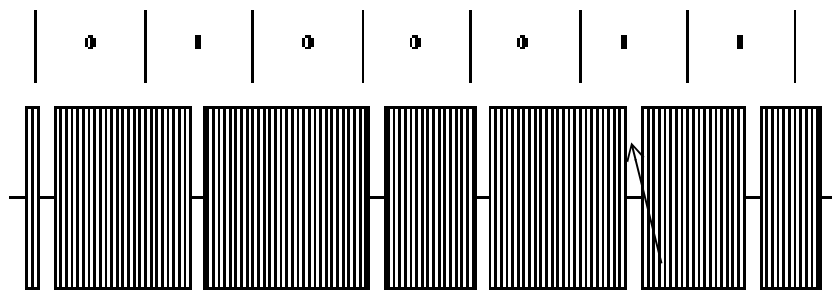
Communication

Energy transfer

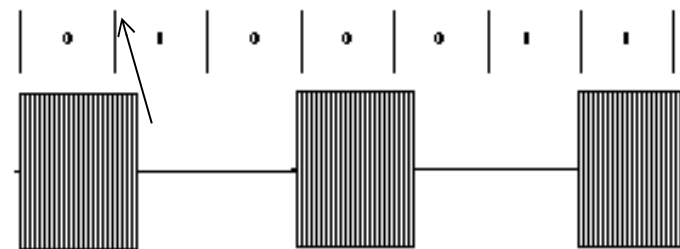
Data communication and Energy transfer can take place simultaneously

In such cases the Modulation and Codification must be carefully designed, otherwise the Energy transfer will be degraded

An inappropriate combination of Codification-Modulation with long dead periods (signal off) would lead the tag to fail



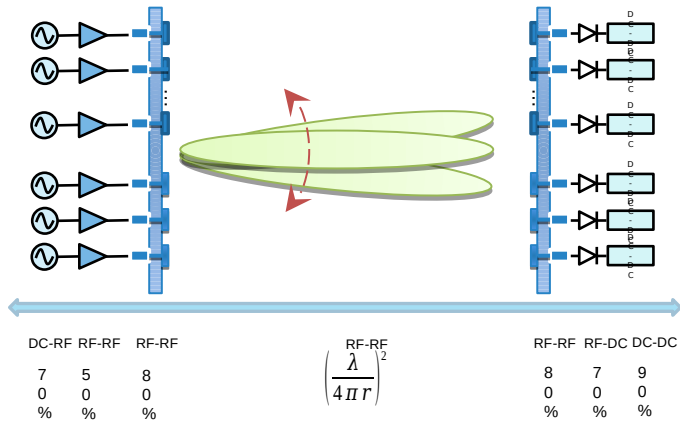
Good choice: signal is off for short periods of time



Bad choice: signal is off for long periods of time
✉ At those periods tag has no available energy to operate

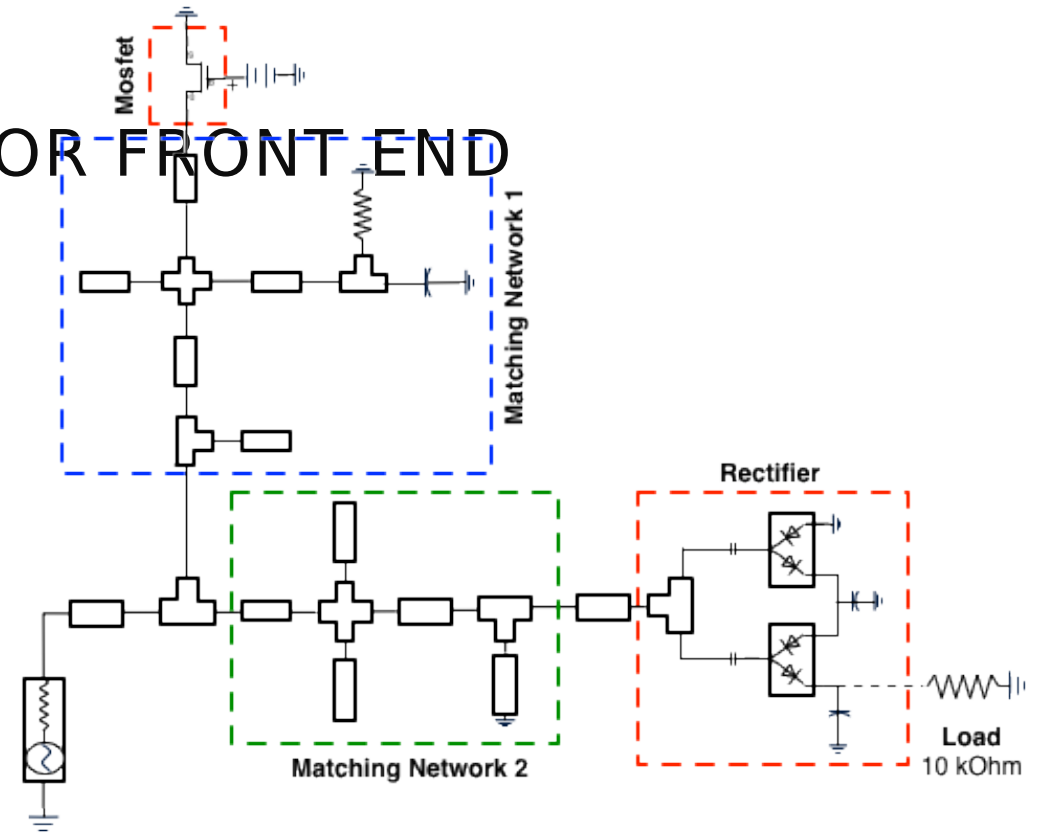
BACKSCATTER

Combining WPT and Backscatter using dual band sensors

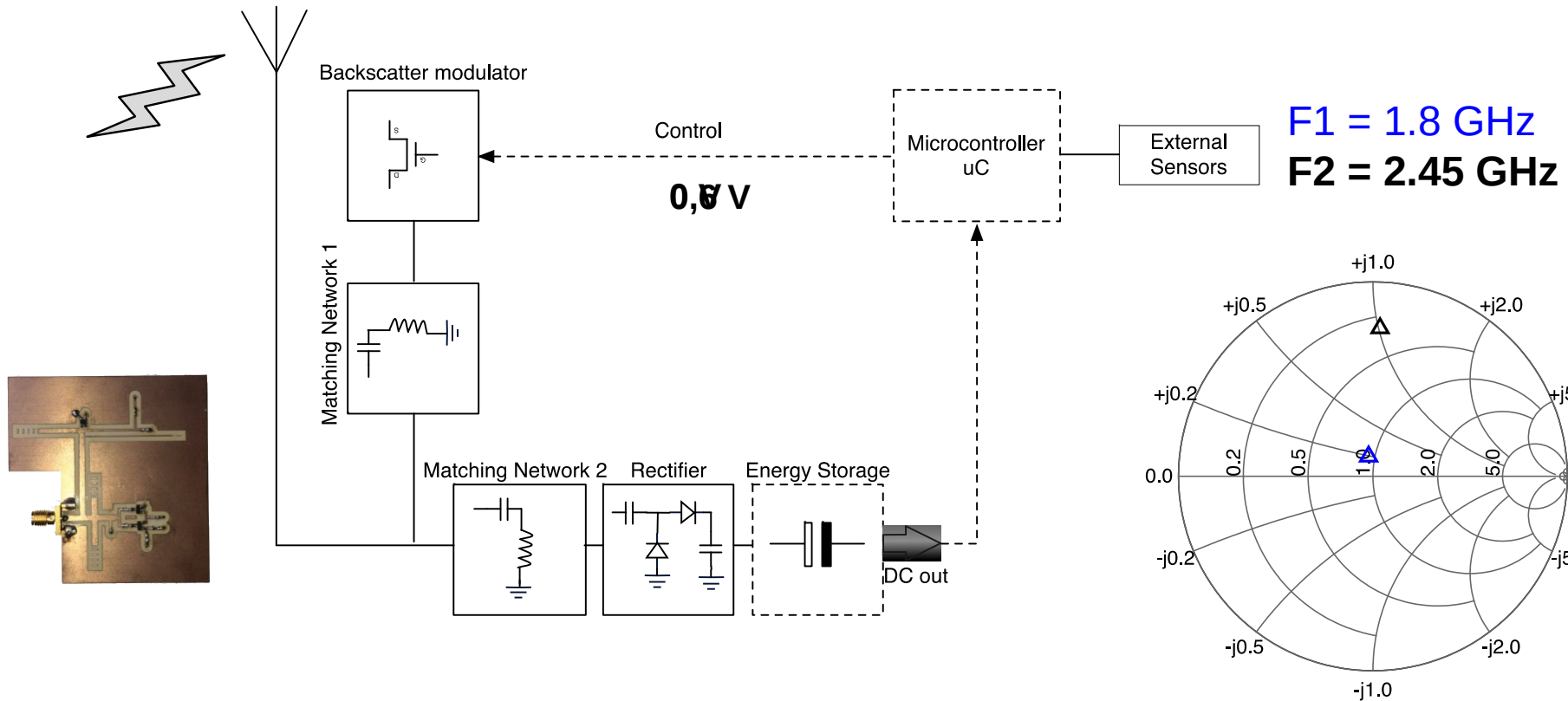
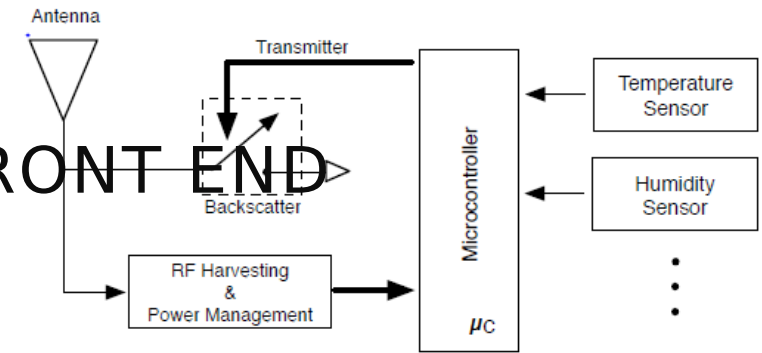


ONE frequency for WPT and OTHER for backscatter

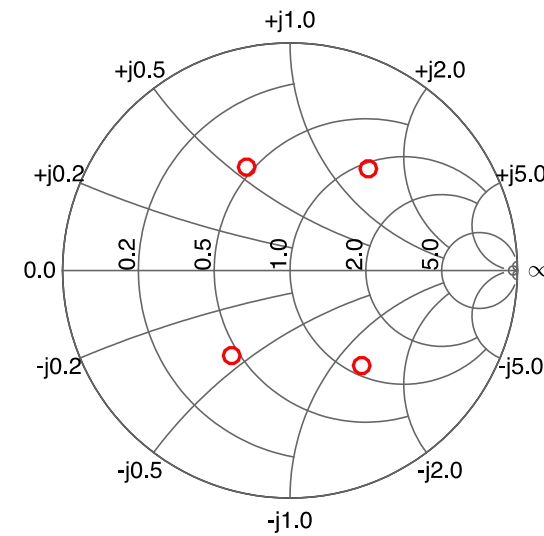
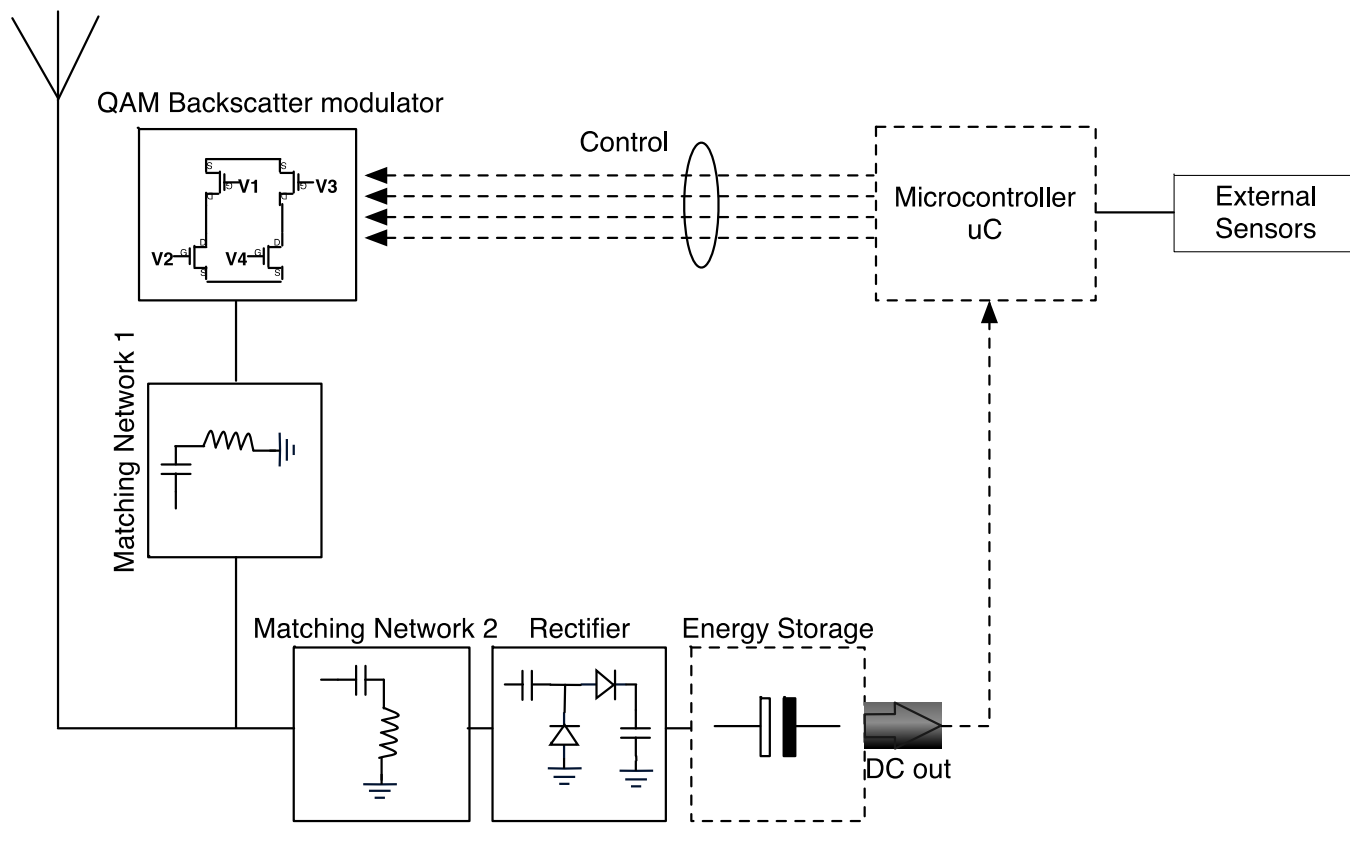
SENSOR FRONT END



BACKSCATTER SENSOR FRONT END

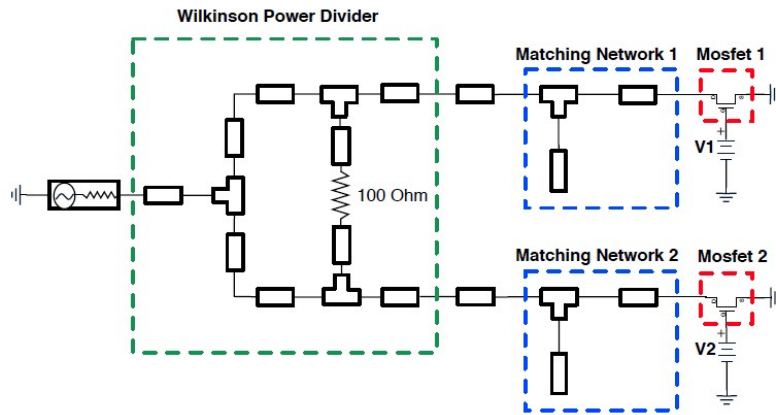


BACKSCATTER SENSOR FRONT END – HIGHER ORDER MODULATION

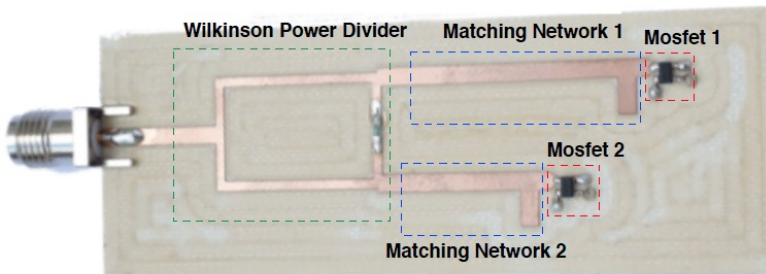


BACKSCATTER EXPERIMENTAL RESULTS

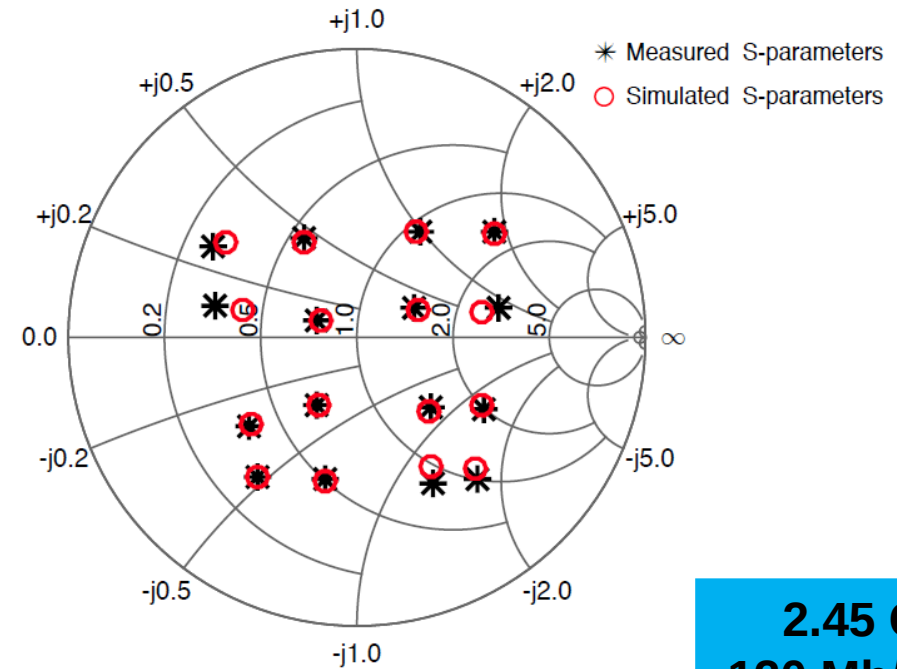
SENSOR FRONT END –



(a) IQ Backscatter Modulator

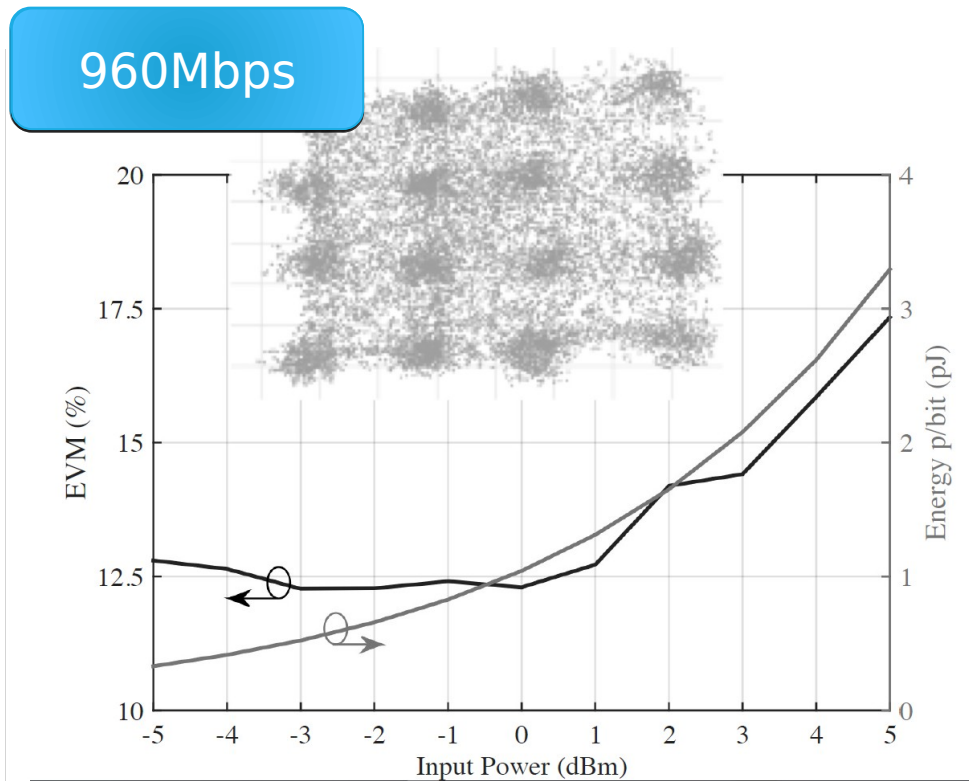
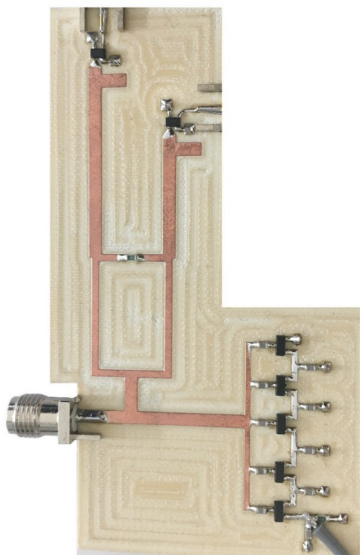


(b)

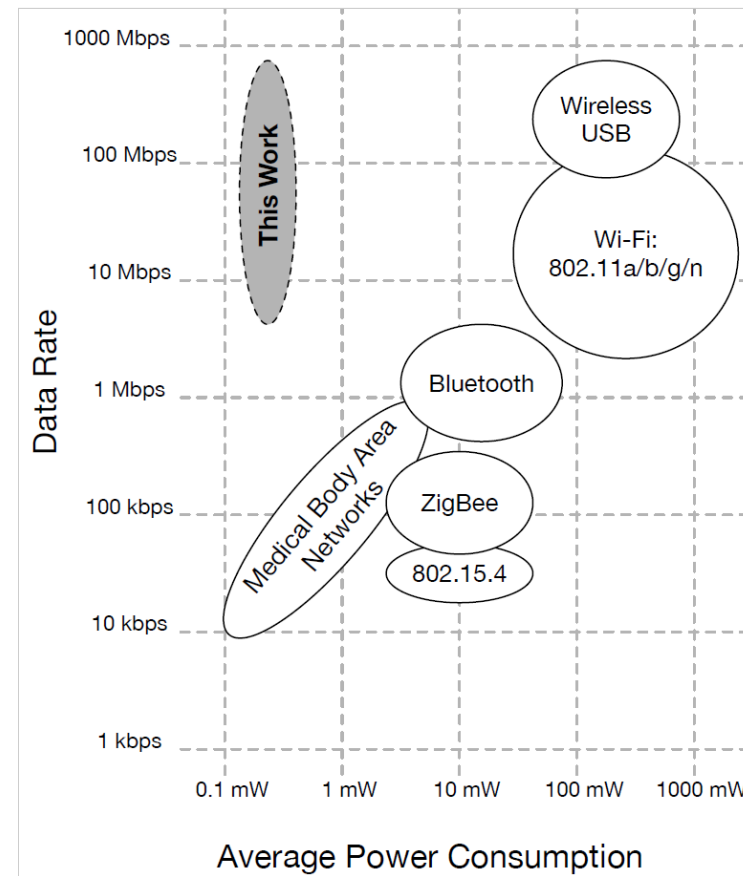


**2.45 GHz
120 Mb/s with
6.7pJ/bit**

BACKSCATTER SENSOR FRONT END – EXPERIMENTAL RESULTS



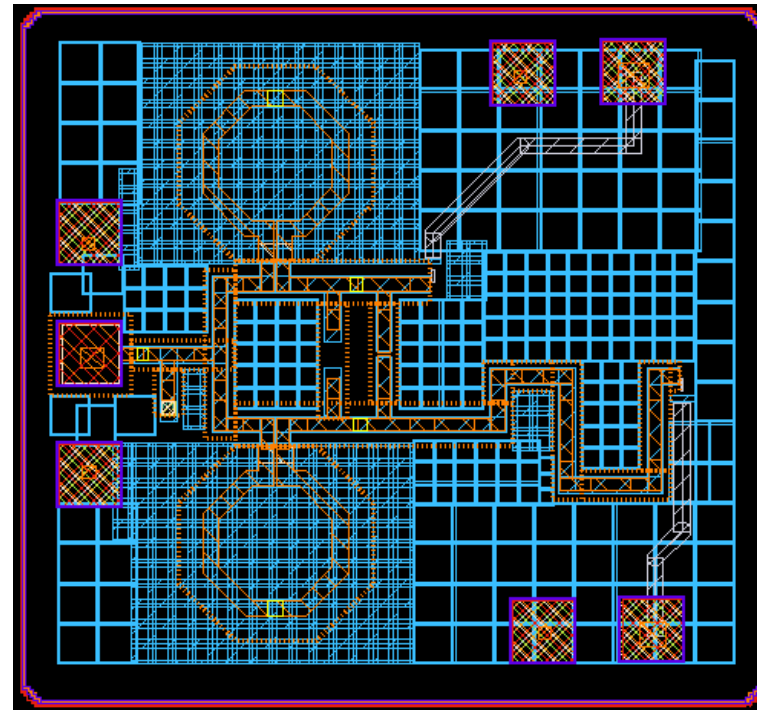
**2.45 GHz
960 Mb/s with
0,9 pJ/bit**



BACKSCATTER

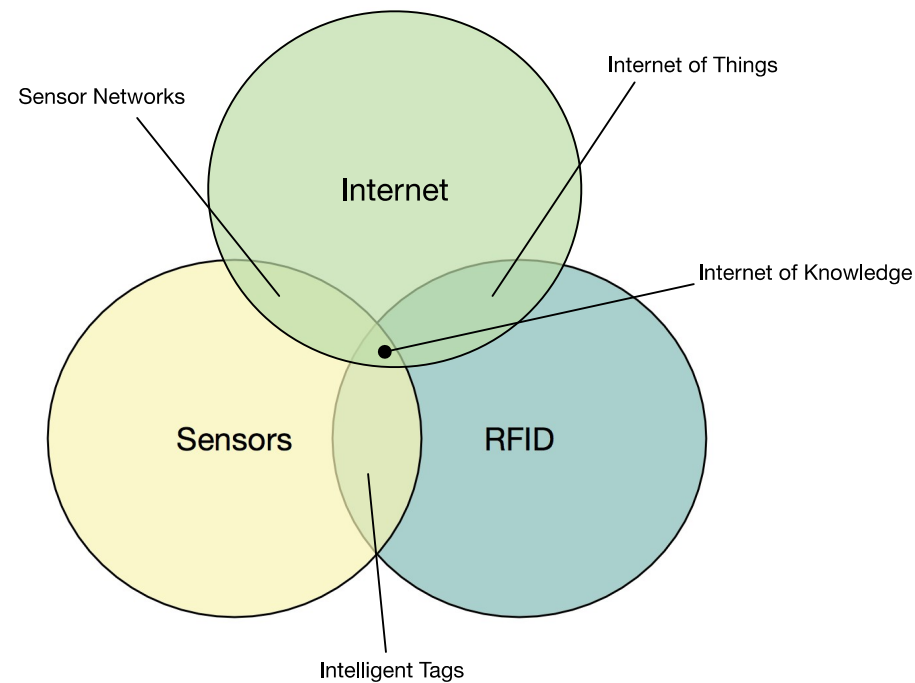
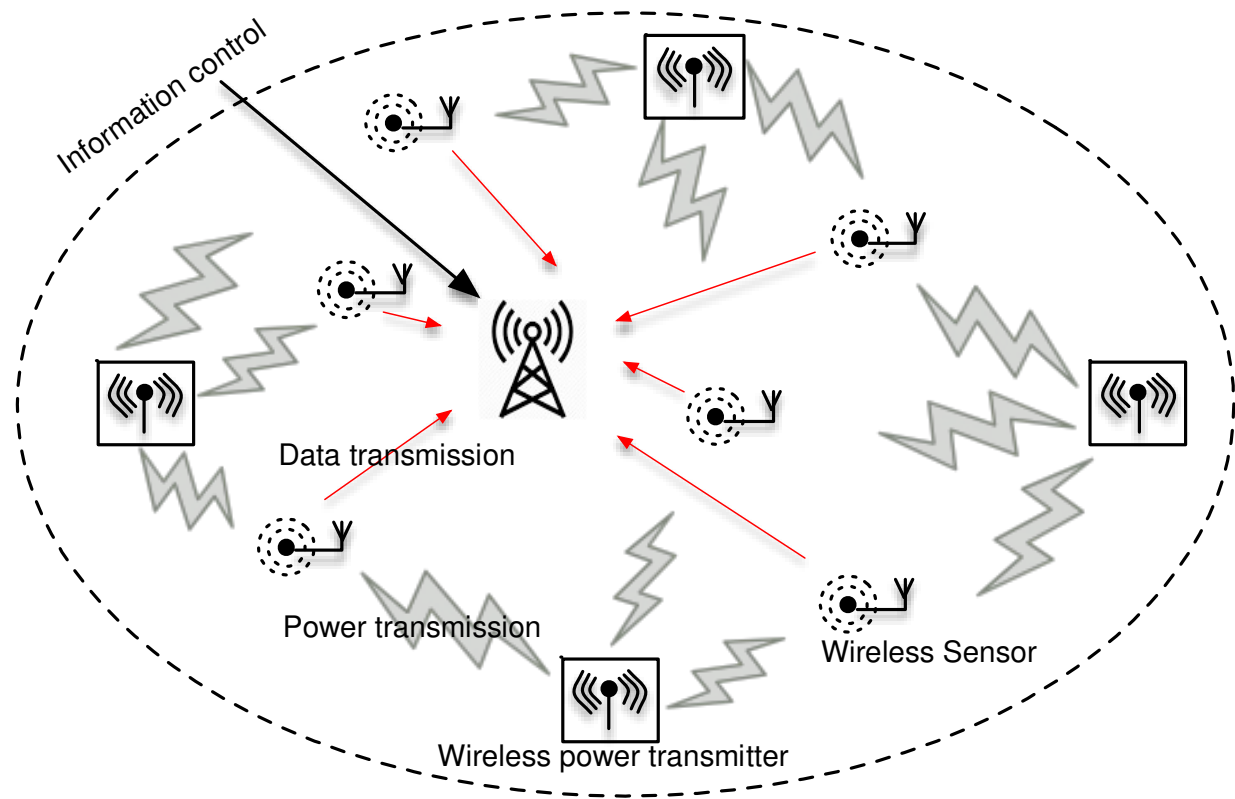
SENSOR FRONT END –
EXPERIMENTAL RESULTS

Design an MMIC chip for higher frequencies based on high order backscatter modulation.



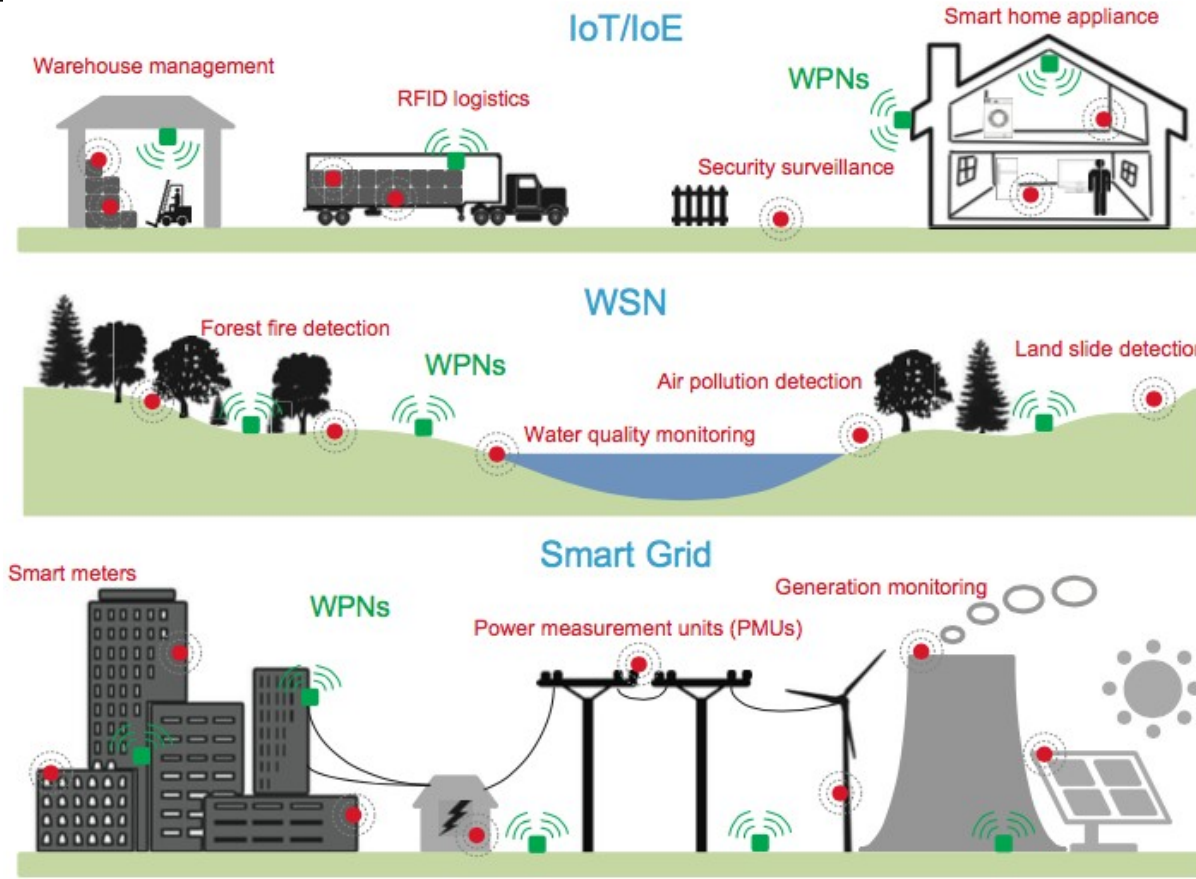
SiGe BICMOS
SG13S
technology
0.92 mm²

SWIPT – THE FUTURE



Solution for totally passive WSNs

BACKSCATTER - APPLICATIONS



WPC (Wireless Power Communication) will be an important building block of many popular commercial and industrial systems in the future, including the upcoming IoT/IoE systems consisting of billions of sensing/RFID devices as well as large-scale WSNs.

BACKSCATTER - APPLICATIONS

Wireless Sensor Node for Backscattering Electrical Signals generated by Plants

- Measures the electrical signal of the plant and backscatters the sensed information to a central reader
- Several WSN sensor nodes
- Carrier emitter
- Reader

Bistatic
Architecture

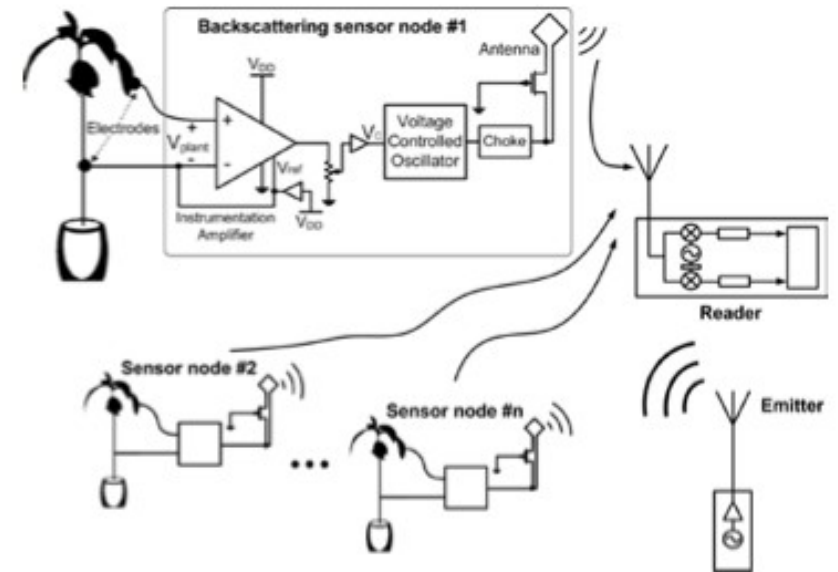
Backscattering Modulator

- Antenna and a switching transistor controlled by square-wave signal

ON OFF

Incoming carrier signal is reflected by antenna and scattered back to the reader (π phase-difference)

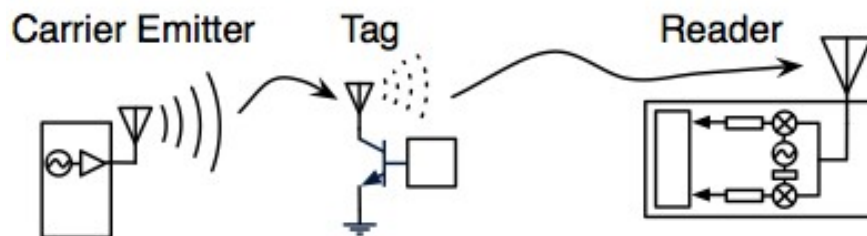
Received signal is reflected back to the reader (zero phase)



BACKSCATTER - APPLICATIONS

Backscatter Sensor Network for Extended Ranges and Low Cost with frequency modulation: Application on Wireless Humidity Sensing

- Low complexity
- Low power
- Low cost

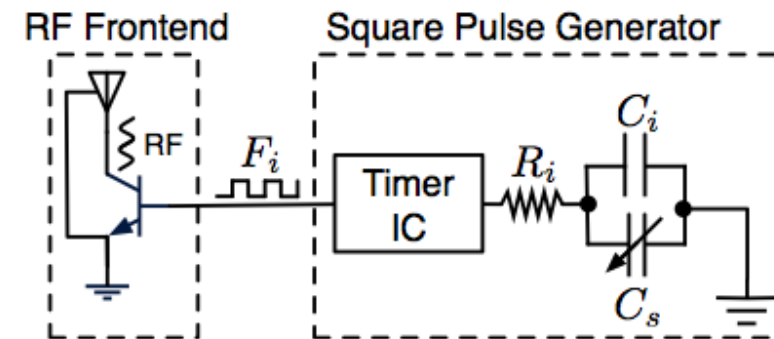


Based on low-cost and low-power relative humidity (%RH) to frequency modulator

$$F_i = \frac{1}{\ln 2 R_i (C_i + C_s)}$$



Each i-th tag can occupy bandwidth on the order of few kHz and is allocated a unique subcarrier frequency center, by selecting a specific pair of R_i, C_i



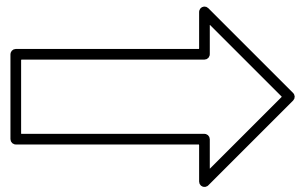
Timer acts as a generator of variable frequency pulser based on %RH

LORRa™

LORA BACKSCATTER

LONG RANGE (LORA) BACKSCATTER COMMUNICATIONS

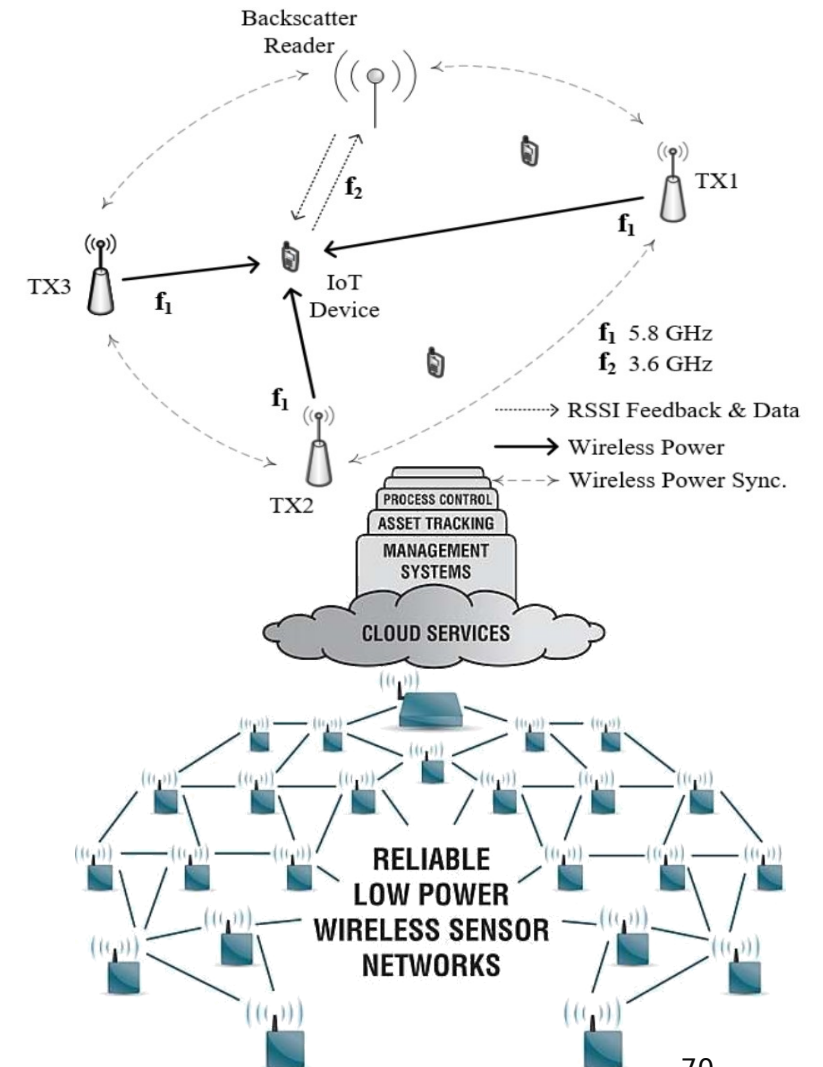
- Recall the distributed WPT scenario...
 - Several WPT beacons, high gain antennas
 - Fewer** backscatter readers with low gain antennas



In that situation, the operating range is no longer limited by the WPT link

- It is desirable to provide **long range** as well as **low power** communications

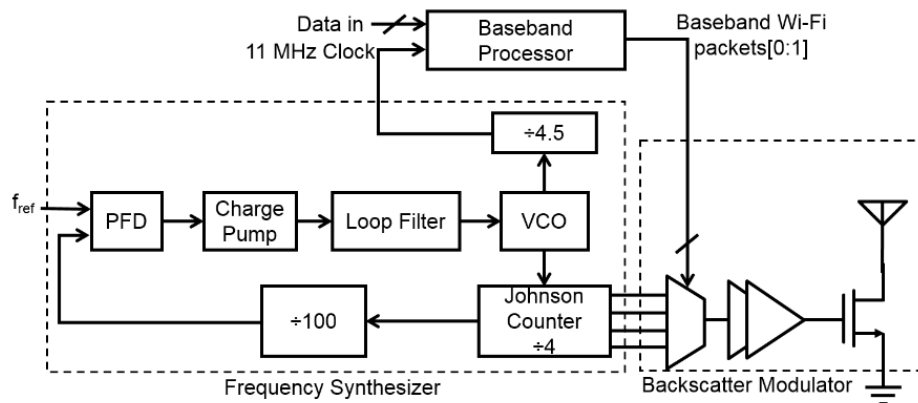
Not so simple to achieve both simultaneously!



LONG RANGE (LORA) BACKSCATTER COMMUNICATIONS

- Most common low power modulations used in IoT applications
- **Sensitivity - Range**
- Is it possible to generate those modulations with backscattering techniques?

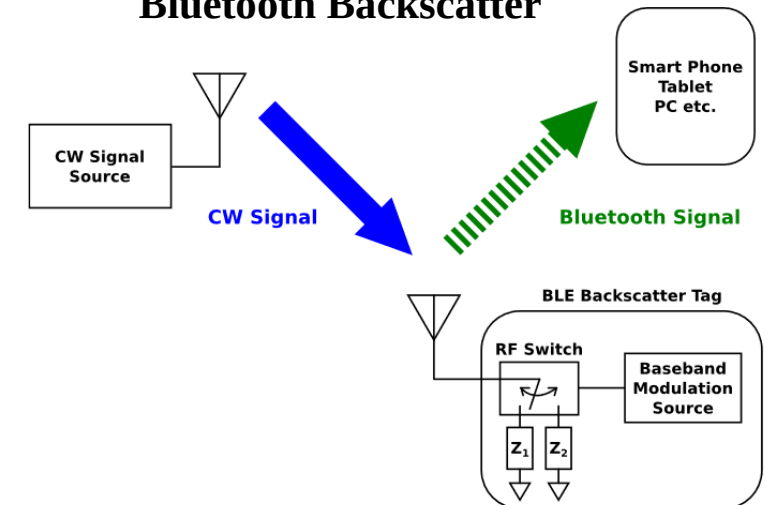
Technology	Sensitivity	Data Rate
LoRa	-149 dBm	18 bps - 37.5 kbps
Sigfox	-126 dBm	100 bps
NB-IoT	-129 dBm	0.6 - 250 kbps
Wi-Fi (802.11 b/g)	-95 dBm	1 - 54 Mbps
Bluetooth	-97 dBm	1 - 2 Mbps
ZigBee	-100 dBm	250 kbps
RFID	-85 dBm	40 - 640 kbps



Wi-Fi Backscatter

	1 Mbps	11 Mbps
Baseband Frequency Synthesizer	5.6 μW	5.6 μW
Baseband Processor	5.0 μW	48 μW
Backscatter Modulator	3.9 μW	5.6 μW
Total Power	14.5 μW	59.2 μW

Bluetooth Backscatter



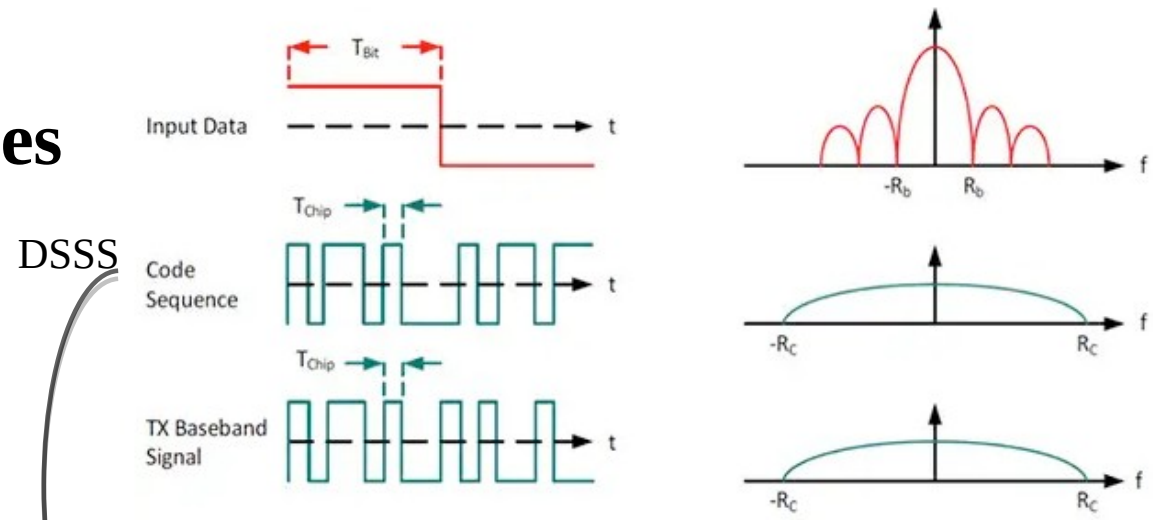
LONG RANGE (LORA) BACKSCATTER COMMUNICATIONS

- LoRa from Semtech provides the highest currently available sensitivity

Technology	Sensitivity	Data Rate
LoRa	-149 dBm	18 bps - 37.5 kbps
Sigfox	-126 dBm	100 bps
NB-IoT	-140 dBm	0.6 - 250 kbps

- Relies on Spread Spectrum techniques

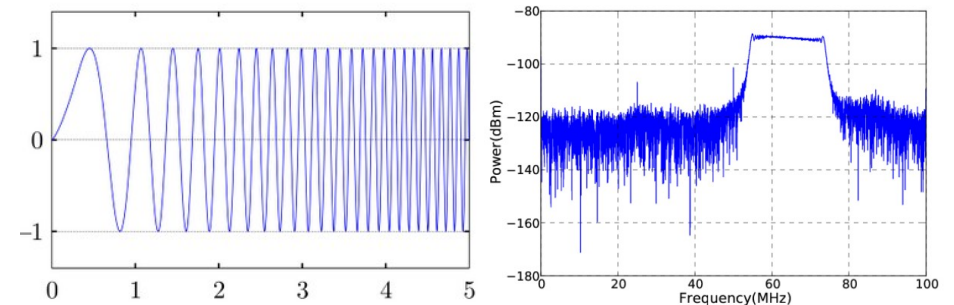
- Resilient to external interference
- Less sensitive to multipath fading
- Co-existence with other systems
- Hard to detect/demodulate



DSSS

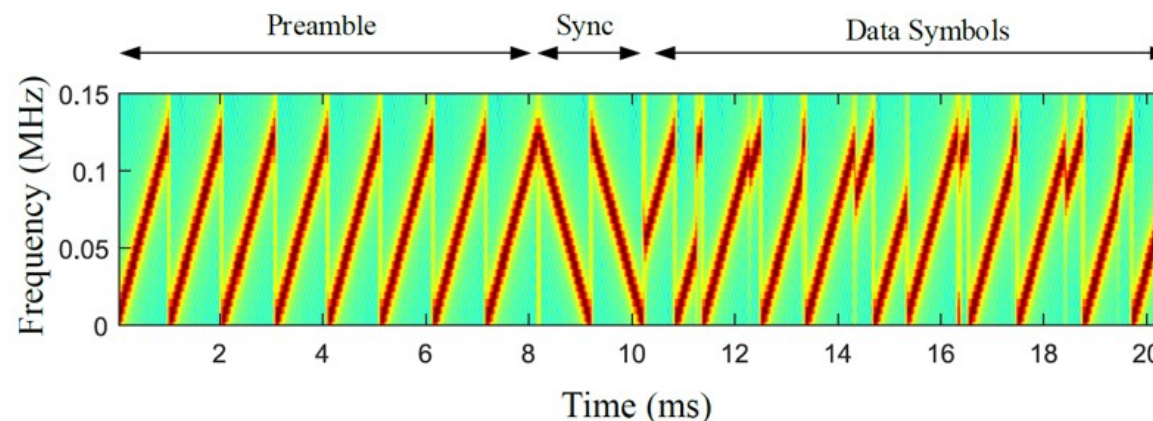
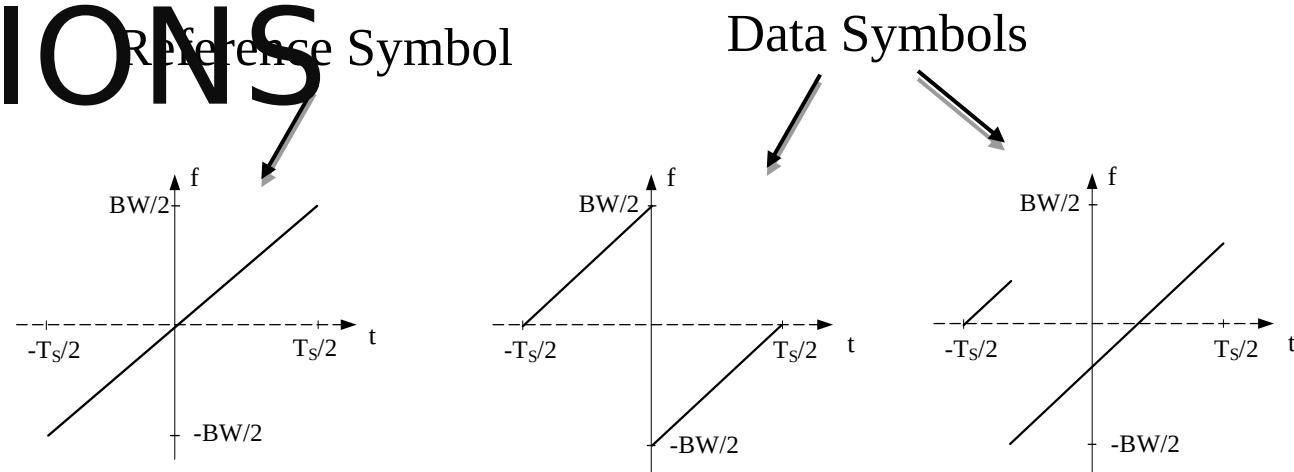
CSS

- Chirp Spread Spectrum

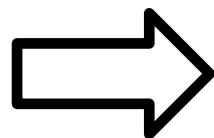


LONG RANGE (LORA) BACKSCATTER COMMUNICATIONS

- LoRa is a special case of chirp spread spectrum modulation
- Cyclic shifts** are introduced to encode data
- Bit rate:** $R_b = SF \frac{BW}{2SF}$
 - SF is the spreading factor (7...12)
 - BW is the bandwidth of the chirp (125, 250 or 500 kHz)



• **chips/symb**



possible cyclic shifts (each symbol carries SF bits)

LORA BACKSCATTER COMMUNICATIONS

- The baseband **reference symbol** is represented by

$$s(t) = e^{j\phi(t)}$$

- The **instantaneous frequency** of the signal, , must increase linearly with time...

$$f(t) = \frac{BW}{T_s} t, \text{ where } T_s = \frac{2^{SF}}{BW} \quad f(t) = \frac{1}{2\pi} \frac{d\phi(t)}{dt}$$

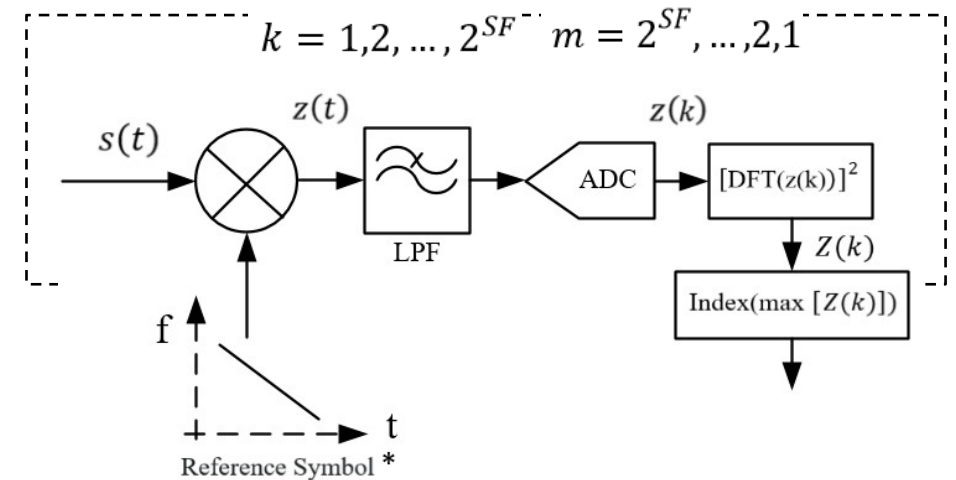
Thus, the **phase progression** required to generate a linear frequency modulated signal (chirp) is given by

$$\phi(t) = \phi_0 + 2\pi \int_0^t f(\tau) d\tau = \phi_0 + \pi \frac{BW}{T_s} t^2 \quad \Rightarrow$$

Demodulation

$$z(t) = s(t) e^{-j(\phi_0 + \pi \frac{BW}{T_s} t^2)}$$

$$z[k = T_s/m]$$

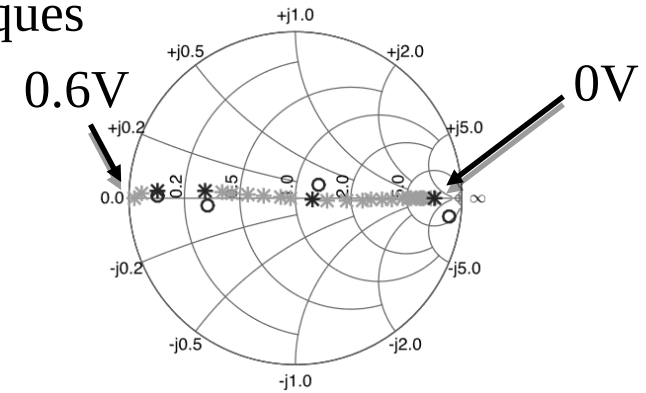
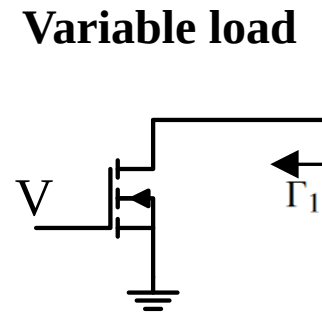
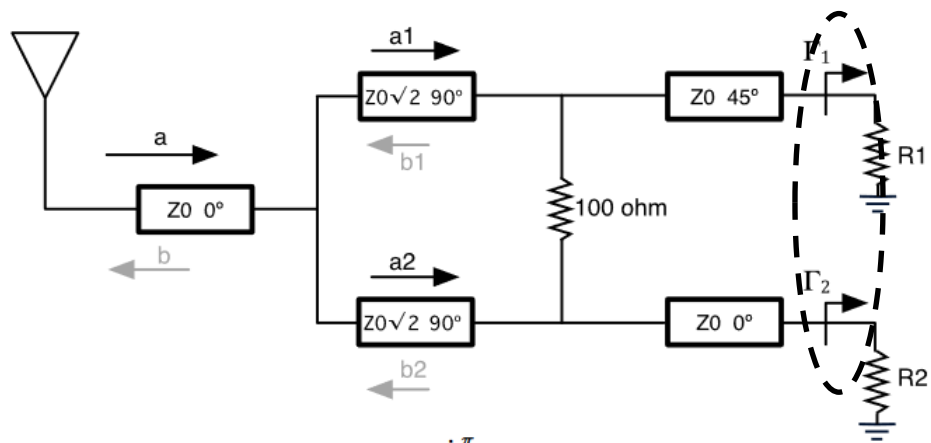


Index(max

Can be efficiently generated with backscattering techniques

LORA BACKSCATTER COMMUNICATIONS

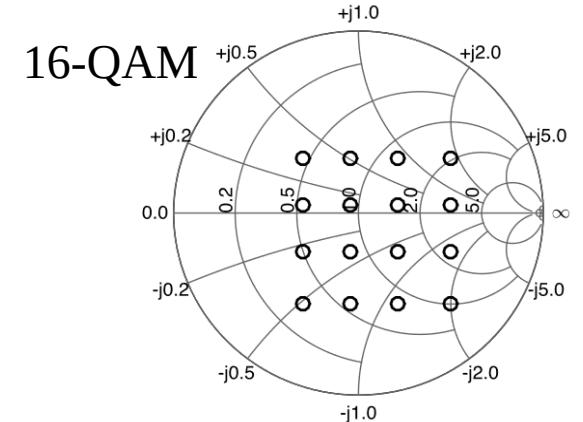
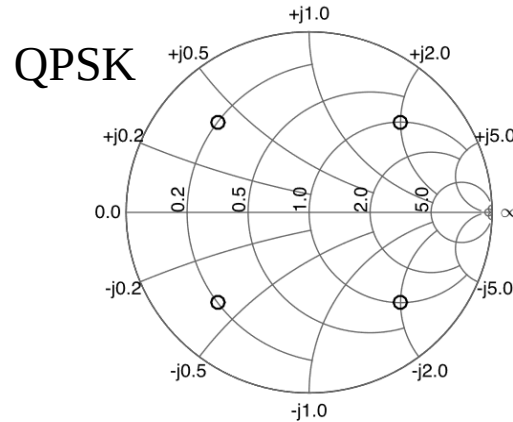
- It is required to **efficiently** generate **complex signals** with backscattering techniques



$$b = \frac{b_1}{\sqrt{2}} + \frac{b_2}{\sqrt{2}} = \frac{a_1 \Gamma_1 e^{j\frac{\pi}{2}}}{\sqrt{2}} + \frac{a_2 \Gamma_2}{\sqrt{2}}$$

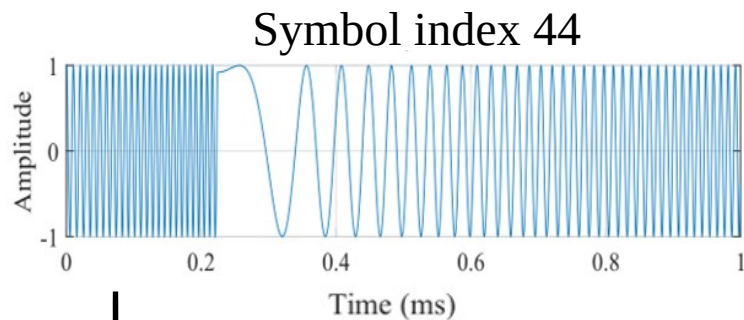
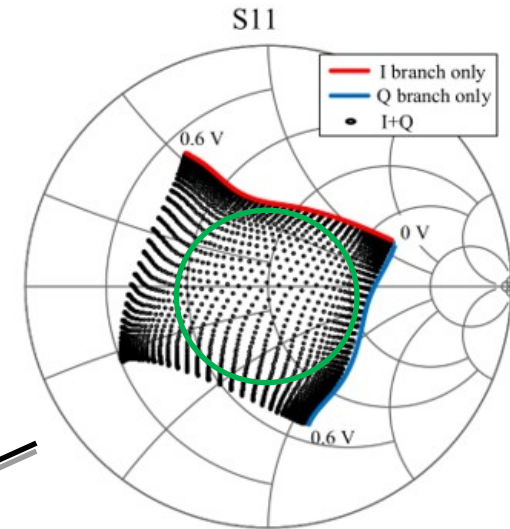
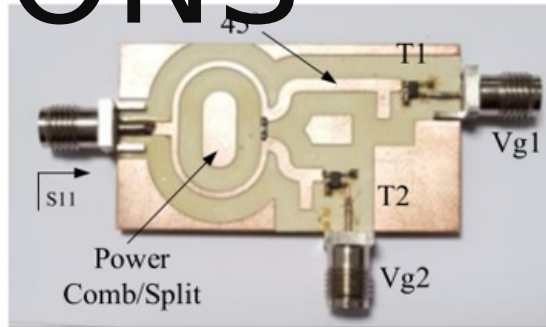
$$\Gamma_1 = \frac{R_1 - Z_0}{R_1 + Z_0} \quad \Gamma_2 = \frac{R_2 - Z_0}{R_2 + Z_0}$$

$$\Gamma = \frac{b}{a} = \frac{\Gamma_1 \frac{1}{\sqrt{2}} e^{j\frac{\pi}{2}}}{\sqrt{2}} + \frac{\Gamma_2 \frac{1}{\sqrt{2}}}{\sqrt{2}} \Rightarrow \Gamma = \frac{\Gamma_2}{2} + j \frac{\Gamma_1}{2}$$

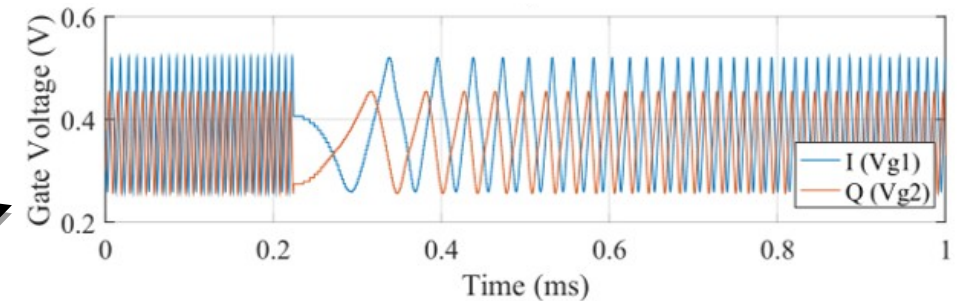
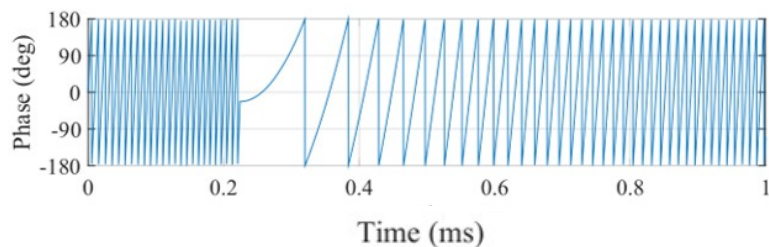
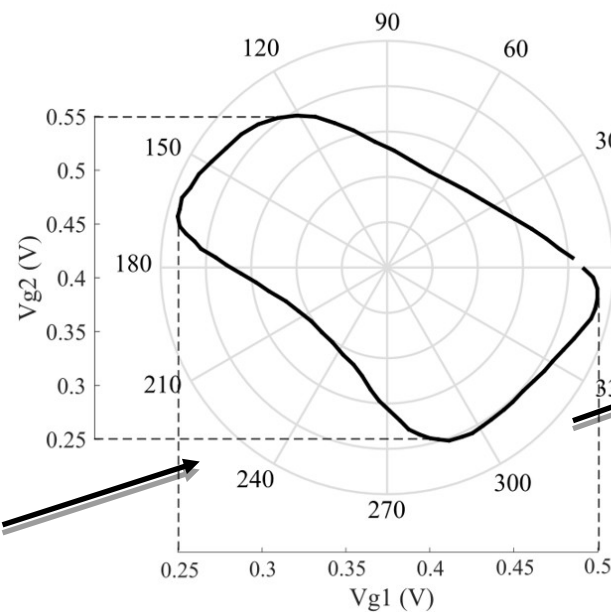


LORA BACKSCATTER COMMUNICATIONS

- **Characterization @ 2.45 GHz**
(mesh of 1x1 mV)
- **Non-Linear Transformation**



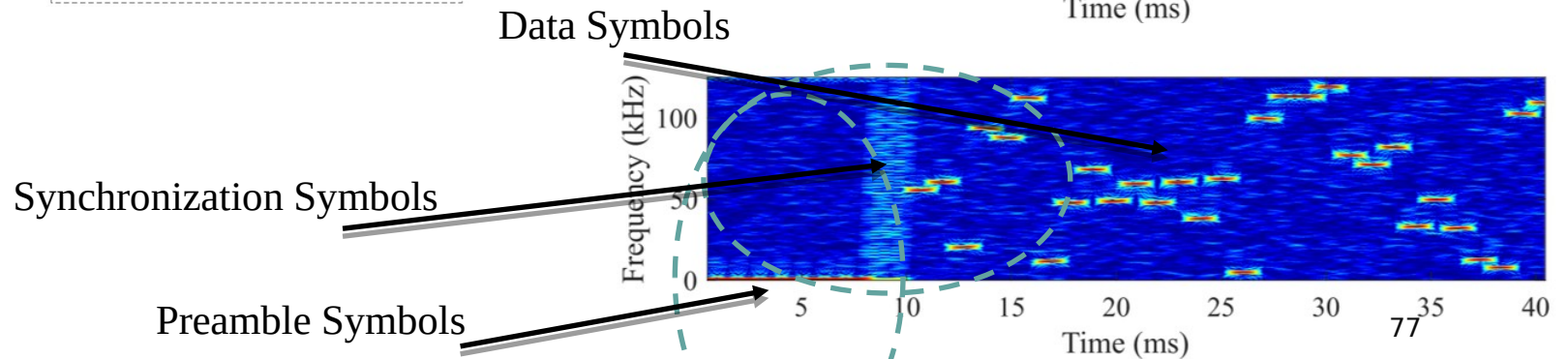
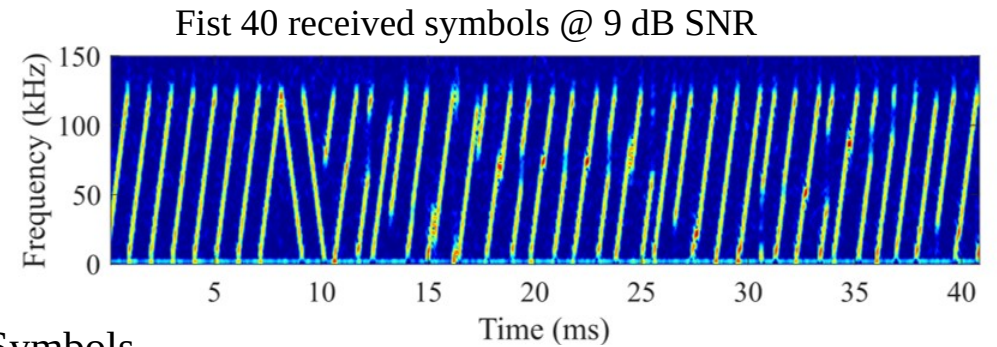
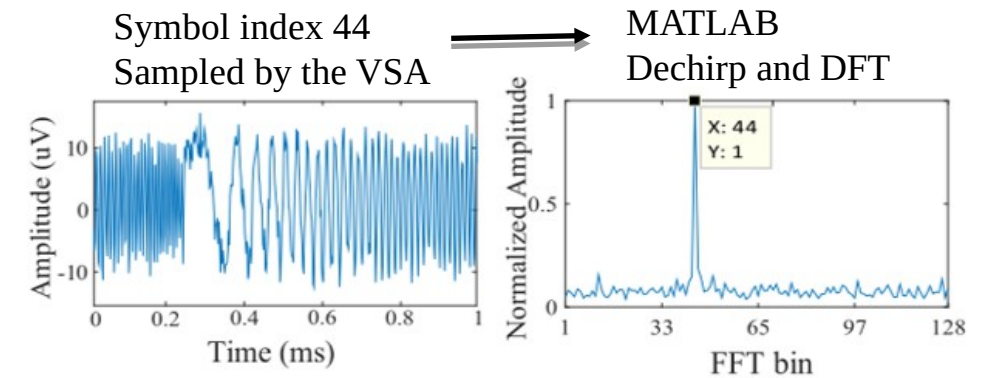
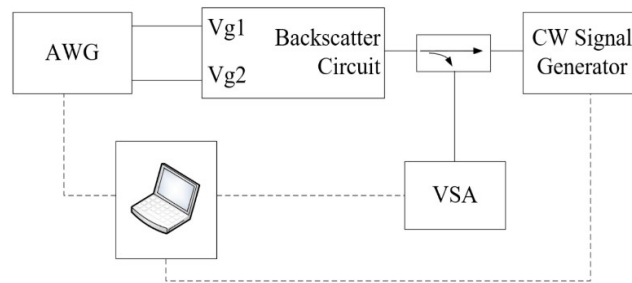
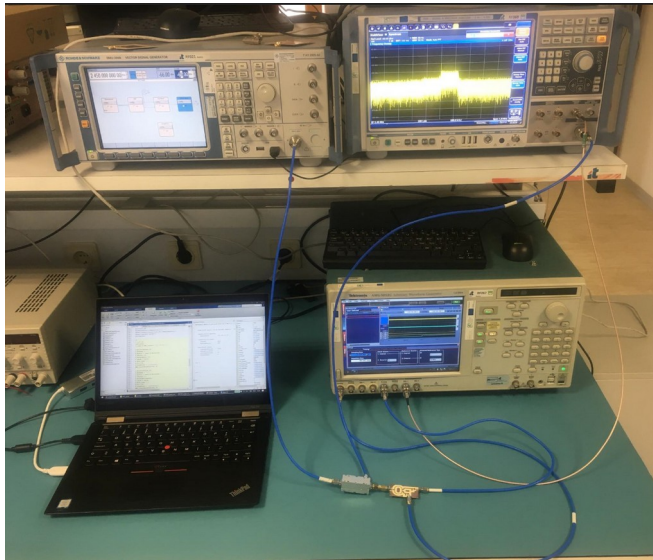
$$\phi(t) = \phi_0 + \pi \frac{BW}{T_s} t^2$$



LORA BACKSCATTER COMMUNICATIONS

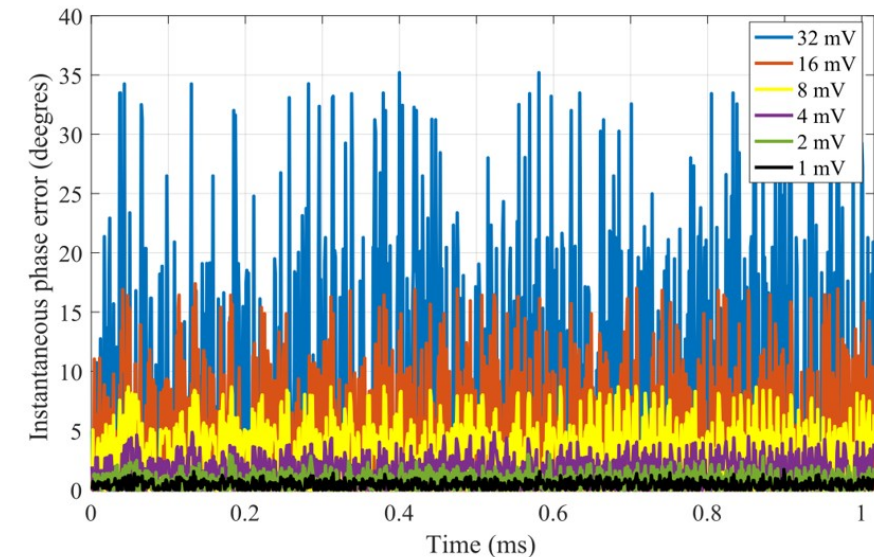
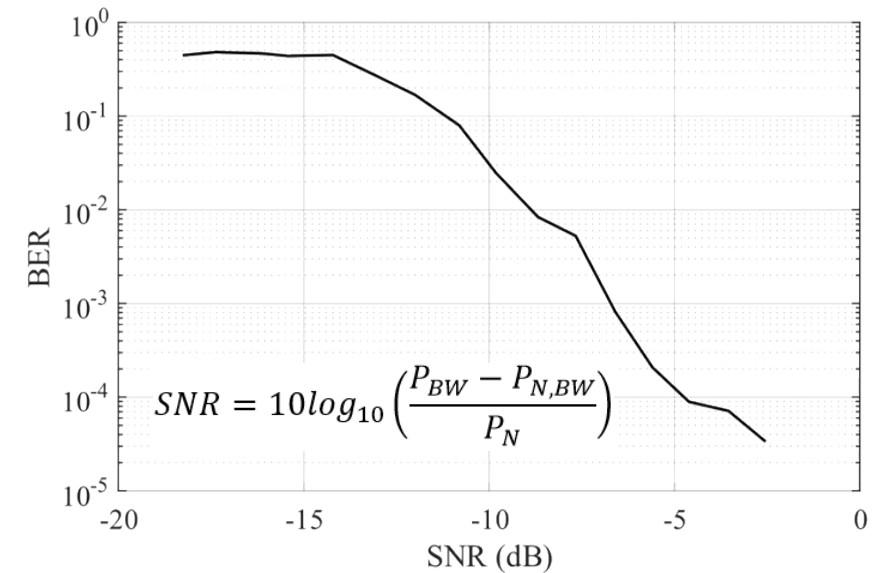
- **Cable measurements**

- 12,000 random symbols
- V_{G1} and V_{G2} generated with MATLAB and loaded to an AWG
- VSA down-converts to the complex baseband
- IQ loaded to MATLAB and decoded

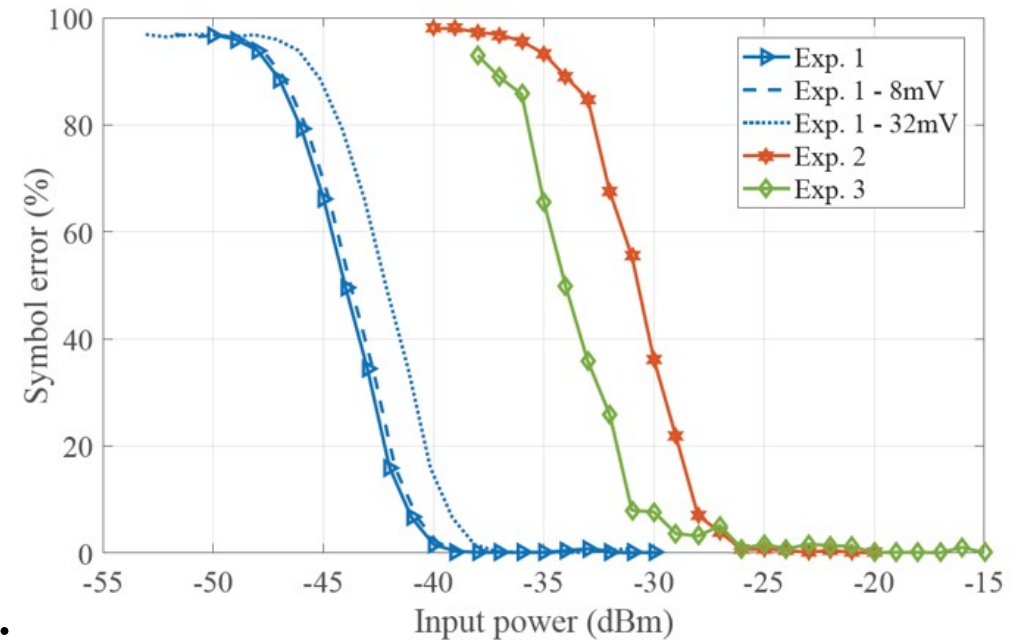
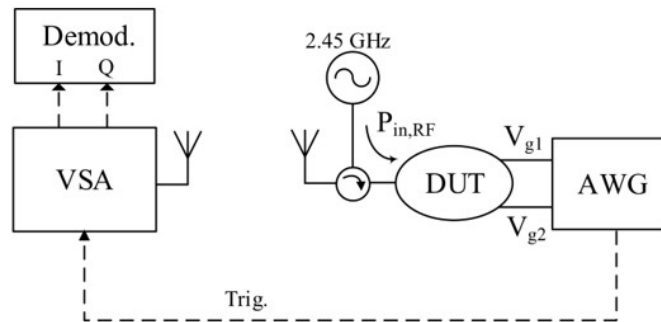


LORA BACKSCATTER COMMUNICATIONS

- **BER estimation**
 - 125 kHz LoRa BW
 - Sampled by the VSA at $4 \cdot BW$ in order to obtain an estimation of the noise power
 - Down-converted in MATLAB to exactly BW Hz
 - 12,000 Symbols transmitted several times for each value of SNR
- **Control voltage sensitivity**
 - Perturbations intentionally added to the control voltages, 1, 2, 4, 8, 16 and 32 mV



LOAD BACKSCATTER COMMUNICATIONS



• Over-the-Air Evaluation

- **Exp. 1** – Typical indoor scenario with LOS conditions, ~10 m
- **Exp. 2** – Desks and laboratorial hardware in between the antennas, ~7.5 m
- **Exp. 3** – A wall in between the antennas, ~10 m

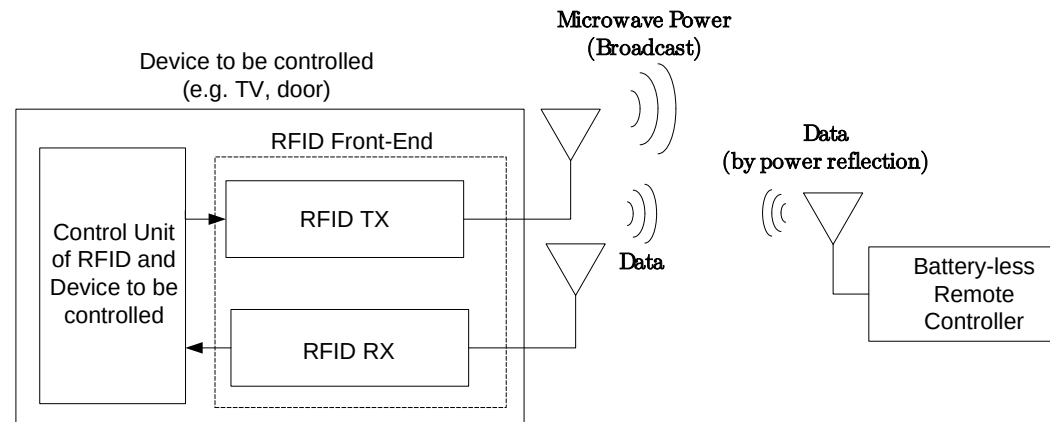


APPLICATIONS: PASSIVE REMOTE CONTROL

BATTERY-LESS REMOTE CONTROL

A battery-free **Remote Control System** is proposed:

- ❖ The Remote requires no battery, based on passive RFID technology
- ❖ Device to be Controlled wirelessly powers the remote control using radio waves
- ❖ The remote control send back information using Backscattering (Power reflection)

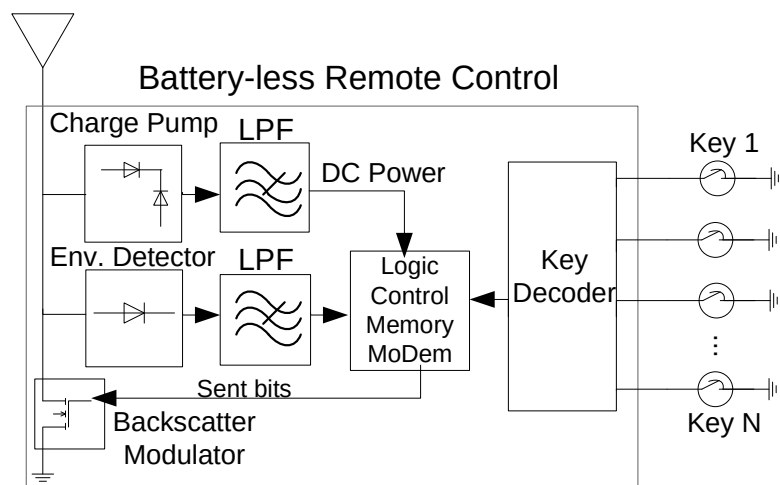


Advantages compared to conventional IR technology:

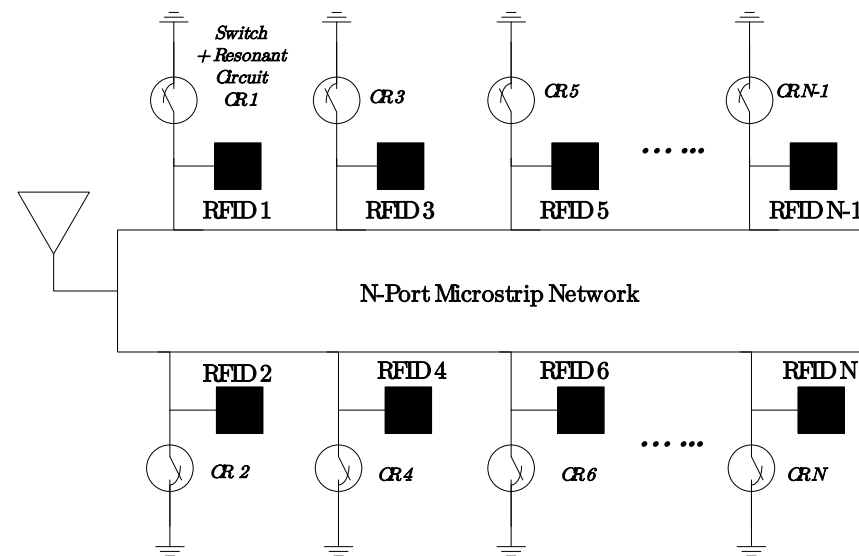
- ✓ **Elimination of costs** associated to battery maintenance and treatment of toxic waste
- ✓ **Long range and no line of sight communication** thanks to the use of radio waves
- ✓ **Cost-effective solution**, thanks to the use of a low-cost RFID technology (UHF EPC)

PROPOSED SOLUTIONS

Option I: Passive Wireless Sensor - alike



Option II: Multi-RFID scheme



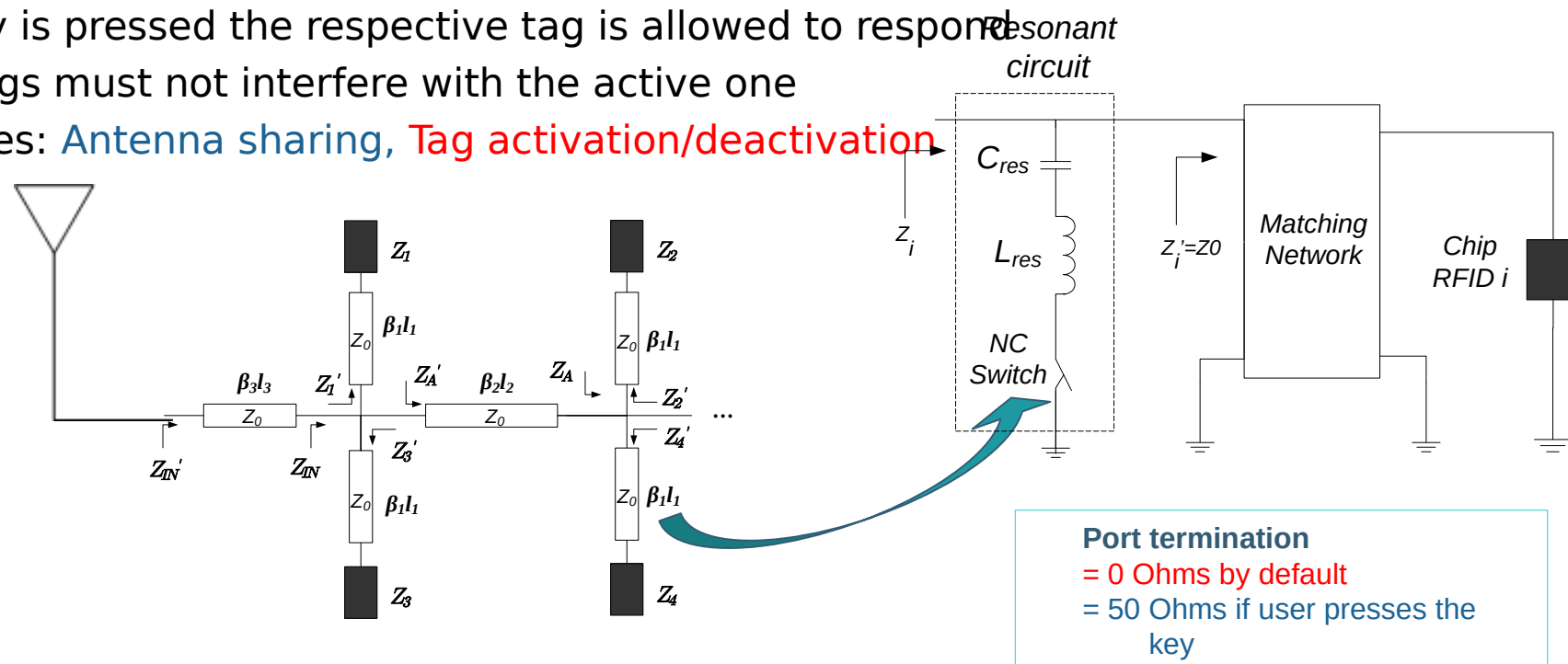
Multi-RFID scheme is implemented

- ❑ Several RFID chips are used, each one associated to a key
- ❑ Only the chip associated to the pressed key should be read by the RFID reader to identify the key

MULTI-RFID SCHEME

Operating principle:

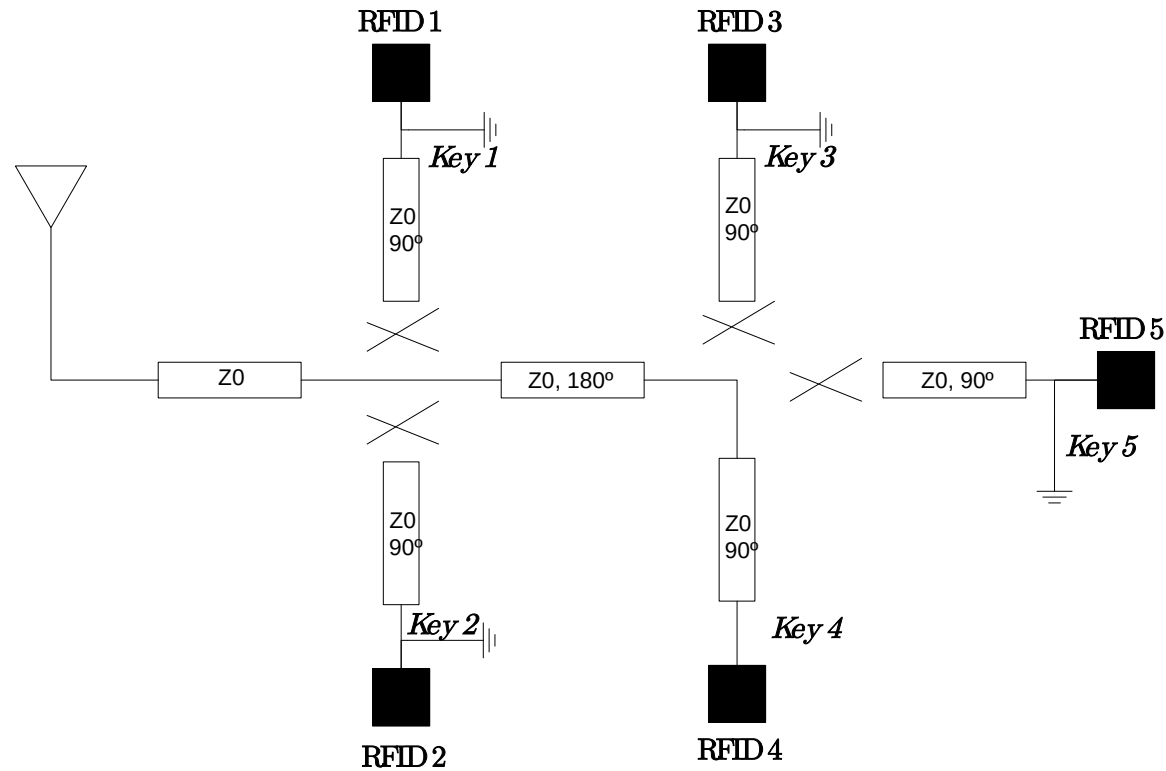
- N passive RFID tags associated to N keys/switchs
 - By default, no tag responds to reader (silent mode)
 - Once a key is pressed the respective tag is allowed to respond
 - Inactive tags must not interfere with the active one
- Two challenges: **Antenna sharing**, **Tag activation/deactivation**



Tags interconnection (N-port Network)

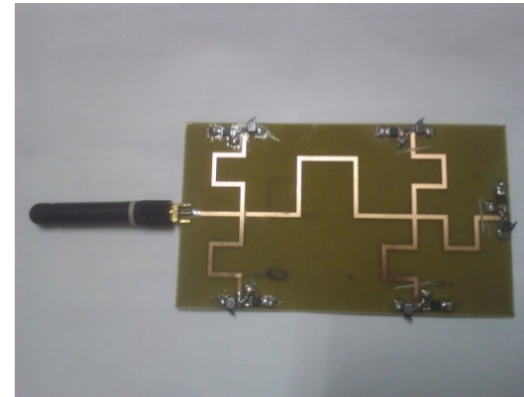
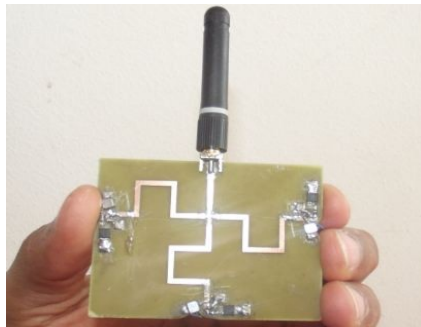
MULTI-RFID SCHEME

Example: key # 4 is pressed RFID4 is routed to the antenna port without interference of idle tags



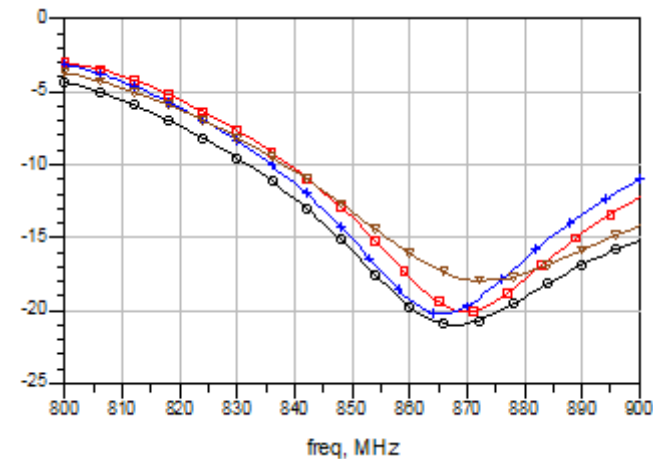
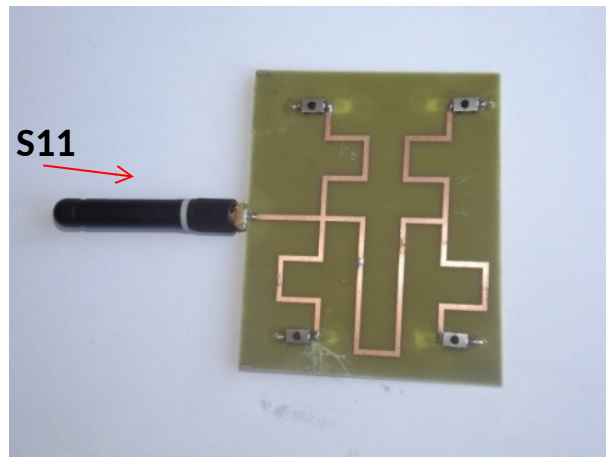
MULTI-RFID SCHEME - MEASUREMENTS

Remote control prototypes: 3, 4 and 5 keys



Return loss (S11) of 4-key prototype when each key is presses by the user

S11 - Return Loss (dB)



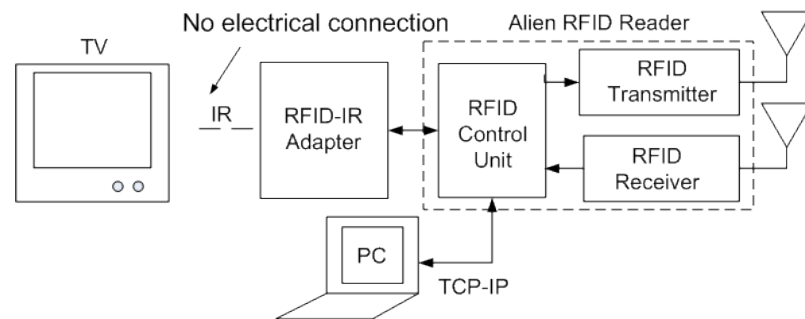
- o – key 1 pressed
- x – key 2 pressed
- – key 3 pressed
- < – key 4 pressed

MULTI-RFID SCHEME

MEASUREMENTS



- ✓ The complete system has been successfully tested and validated
- ✓ The remote control system has been integrated in a TV device
- ✓ **CH +**, **CH -**, **Vol +** and **Vol -** functions were implemented



The prototype is composed by:

- 1) TV
- 2) RFID reader and Computer
- 3) RFID-IR adapter



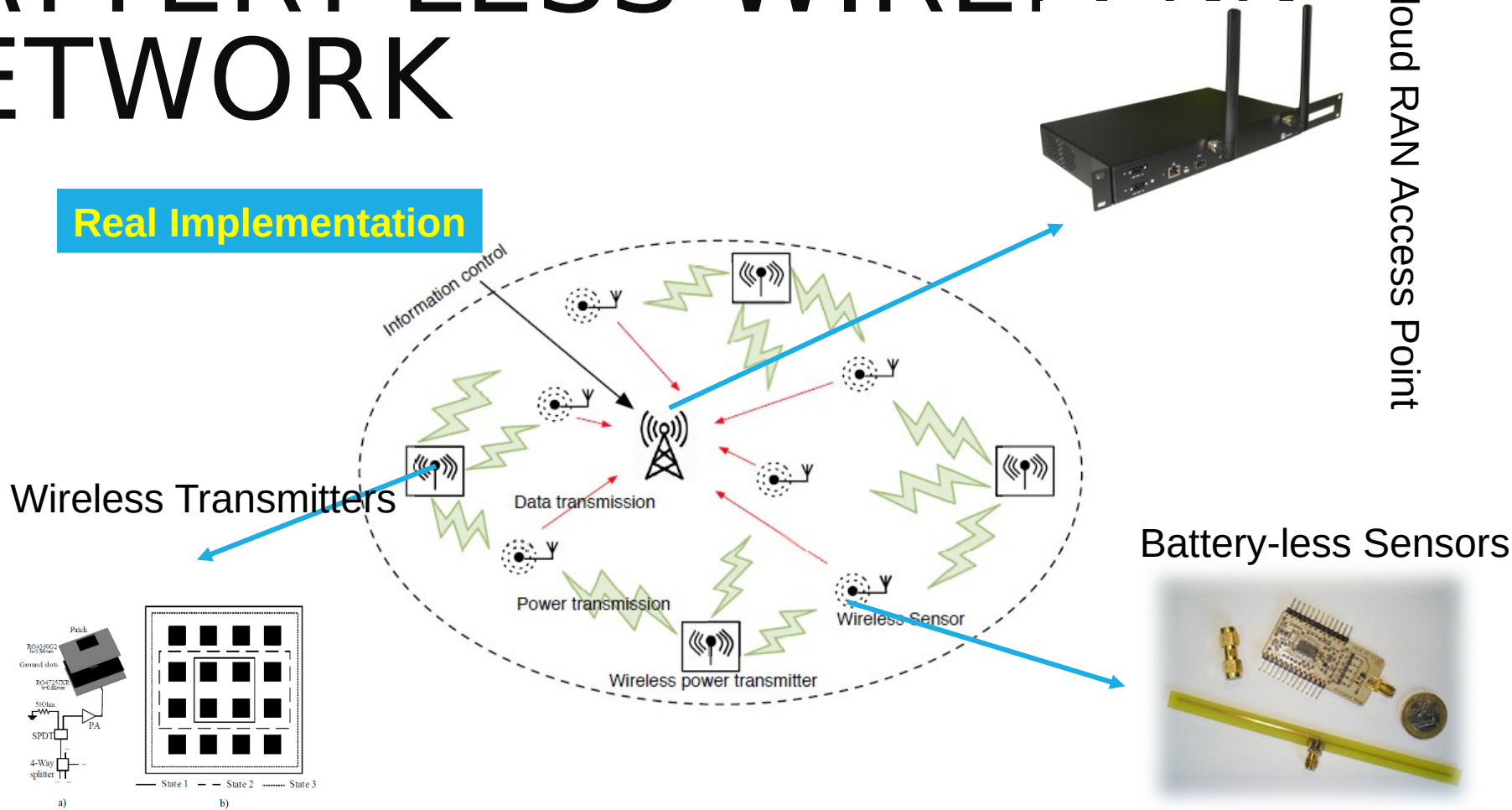
BATTERY-LESS WIRELESS NETWORK

Cloud RAN Access Point

Real Implementation

Wireless Transmitters

Battery-less Sensors



WPT.IEEE.ORG



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